

# **CYNSE70256 Network Search Engine**



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## 1.0 Overview

Cypress Semiconductor Corporation's (Cypress's) CYNSE70256 network search engine (NSE) incorporates patent-pending Associative Processing Technology™ (APT) and is designed to be a high-performance, pipelined, synchronous, 256K-entry NSE. The CYNSE70256 database entry size can be 72 bits, 144 bits, or 288 bits. In the 72-bit entry mode, the size of the database is 128K entries. In the 144-bit mode, the size of the database is 64K entries, and in the 288-bit mode, the size of the database is 32K entries. The CYNSE70256 is configurable to support multiple databases with different entry sizes. The 36-bit entry table can be implemented using the global mask registers (GMRs), building database size of 256K entries with a single device.

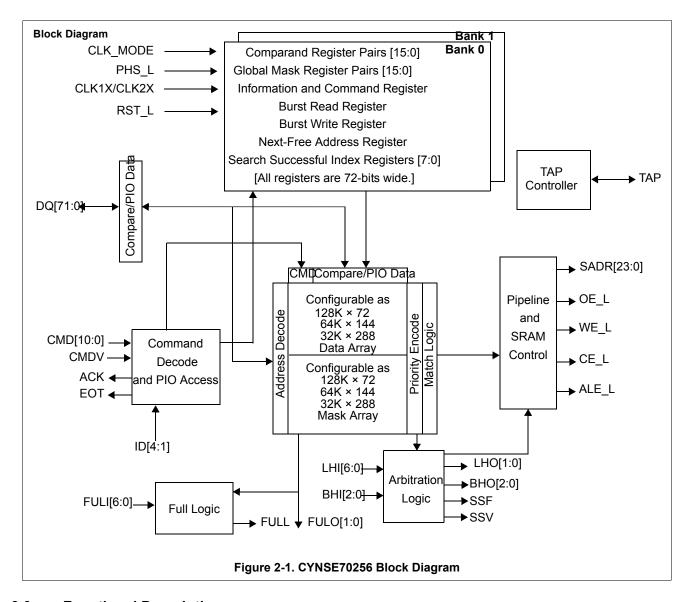
The NSE can sustain 83 million transactions per second when the database is programmed or configured as 72 or 144 bits. When the database is programmed to have an entry size of 36 or 288 bits, the NSE will perform at 41.5 million transactions per second. Cypress's CYNSE70256 can be used to accelerate network protocols such as Longest-Prefix Match (CIDR), ARP, MPLS, and other layer 2, 3, and 4 protocols.

This high-speed, high-capacity NSE can be deployed in a variety of networking and communications applications. The performance and features of the CYNSE70256 make it attractive in applications such as Enterprise LAN switches and routers, and broadband switching and/or routing equipment that supports multiple data rates at OC–48 and beyond. The NSE is designed to be scalable in order to support network database sizes of up to 3840K entries specifically for environments that require large network policy databases. *Figure 2-1* on page 9 shows the block diagram for the CYNSE70256 device.

### 2.0 Features

- 256K 36-bit entries in a single device
- 128K entries in 72-bit mode, 64K entries in 144-bit mode, 32K entries in 288-bit mode
- 83 million transactions per second in 72- and 144-bit configurations (CFGs)
- · 41.5 million transactions in 36- and 288-bit configurations
- · Searches any subfield in a single cycle
- Synchronous pipelined operation
- Up to fifteen NSEs can be cascaded
- When cascaded, database entries can range to 3840K 36-bit entries
- · Multiple width tables in a single database bank
- Glueless interface to industry-standard SRAMs and/or SSRAMs
- · Simple hardware instruction interface
- IEEE 1149.1 test access port
- 1.5V core voltage supply
- 2.5V/3.3V I/O voltage supply
- 388-pin BGA package.





# 3.0 Functional Description

The following subsections contain command (CMD) and DQ bus (command and databus), database entry, arbitration logic, pipeline and SRAM control, and full logic descriptions.

# 3.1 Command Bus and DQ Bus

CMD[10:0] carries the command and its associated parameters. DQ[71:0] is used for data transfer to and from the database entries, which comprise data and mask fields that are organized as data and mask arrays. The DQ bus carries the Search data (of the data and mask arrays and internal registers) during the Search command as well as the address and data during Read and/or Write operations. The DQ bus can also carry address information for the transparent accesses to the external SRAMs and/or SSRAMs.

# 3.2 Database Entry (Data and Mask Arrays)

Each database entry comprises data and mask fields. The resultant value of the entry is "1," "0," or "X (do not care)," depending on the value in the data mask bit. The on-chip priority encoder selects the first matching entry in the database that is nearest to location 0.



# 3.3 Arbitration Logic

When multiple NSEs are cascaded to create large databases, the data being searched is presented to all NSEs simultaneously in the cascaded system. If multiple matches occur, arbitration logic on the NSEs will enable the winning device (the one with a matching entry closest to address 0 of the cascaded database) to drive the SRAM bus.

# 3.4 Pipeline and SRAM Control

Pipeline latency is added to give enough time to a cascaded system's arbitration logic to determine the device that will drive the index of the matching entry on the SRAM bus. Pipeline logic adds latency to both the SRAM access cycles and the search successful flag (SSF) and search successful flag valid (SSV) signals to align them to the host ASIC receiving the associated data.

# 3.5 Full Logic

Bit[0] in each of the 72-bit entries has a special purpose for the Learn command (0 = empty, 1 = full). When all the data entries have bit[0] set to 1, the database asserts the FULL flag, indicating that all the NSEs in the depth-cascaded array are full.

# 4.0 Signal Descriptions

Table 4-1 lists and describes all CYNSE70256 signals.

Table 4-1. CYNSE70256 Signal Description

Pin Name	Pin Type <sup>[1]</sup>	Pin Description			
Clocks and Reset	•				
CLK_MODE	ı	Clock Mode. This signal allows the selection of clock input to the CLK1X/CLK2X pin. If the CLK_MODE pin is LOW, CLK2X must be supplied on that pin. PHS_L must also be supplied. If the CLK_MODE pin is HIGH, CLK1X must be supplied on the CLK2X/CLK1X pin, and the PHS_L signal is not required. When the CLK_mode is HIGH, PHS_L is unused and should be externally grounded.			
CLK2X/CLK1X	I	Master Clock. Depending on the CLK_MODE pin, either the CLK2X or the CLK1X must be supplied. CYNSE70256 samples control and data signals on both edges of CLK1X (if CLK1X is supplied). CYNSE70256 samples all data and control pins on the positive edge of CLK2X (if the CLK2X and PHS_L signals are supplied). All signals are driven out of the device on the rising edge of CLK1X (if CLK1X is supplied), and are driven on the rising edge of CLK2X when PHS_L is LOW (if CLK2X is supplied).			
PHS_L	I	<b>Phase</b> . This signal runs at half the frequency of CLK2X and generates an internal CLK from CLK2X. See Section 5.0, "Clocks," on page 12.			
RST_L	I	Reset. Driving RST_L LOW initializes the device to a known state.			
CFG_L	I	Configuration. When CFG_L is LOW, CYNSE70256 will operate in backward compatibil mode with CYNSE70032 and CYNSE70064. When CFG_L is LOW, the CMD[10:9] should be externally grounded. With CFG_L LOW, the device will behave identically w CYNSE70032 and CYNSE70064, and the new feature added to CYNSE70256 will be disabled. When CFG_L is HIGH, the additional CMD[10:9] can be used and the following additional features will be supported: 1. sixteen pairs of global masks are supported instead of eight; 2. parallel Write to the data and mask arrays is supported (see Subsection 10.5, "Parallel Write," on page 24); and 3. configuring tables of up to three different widths does not require table identification bits in the data array, thus saving to bits from each 72-bit entry.			
Command and DQ Bus					
CMD[10:0]	I	Command Bus. [1:0] specifies the command; [10:2] contains the command parameter. The descriptions of individual CMDs explains the details of the parameters. The encode of CMDs based on the [1:0] field are: 00: PIO Read 01: PIO Write 10: Search 11: Learn.			
CMDV	I	Command Valid. This signal qualifies the command bus: 0: No command 1: Command.			



Table 4-1. CYNSE70256 Signal Description (continued)

Pin Name	Pin Type <sup>[1]</sup>	Pin Description	
DQ[71:0]	I/O	Address/Data Bus. This signal carries the Read and Write address and data during register, data, and mask array operations. It carries the compare data during Search operations. It also carries the SRAM address during SRAM PIO accesses.	
ACK <sup>[2]</sup>	Т	<b>Read Acknowledge</b> . This signal indicates that valid data is available on the DQ bus during register, data, and mask array Read operations, or that the data is available on the SRAN data bus during SRAM Read operations.	
EOT <sup>[2]</sup>	Т	<b>End of Transfer</b> . This signal indicates the end of burst transfer to the data or mask arra during Read or Write burst operations.	
SSF	Т	Search Successful Flag. When asserted, this signal indicates that the device is the globa winner in a Search operation.	
SSV	Т	Search Successful Flag Valid. When asserted, this signal qualifies the SSF signal.	
HIGH_SPEED	I	<b>High Speed</b> . This pin must be connected to ground (V <sub>SS</sub> ). It is provided for backward as well as forward device compatibility.	
RAM Interface		·	
SADR[23:0]	Т	<b>SRAM Address</b> . This bus contains address lines to access off-chip SRAMs that contain associative data. See <i>Table 12-1</i> for the details of the generated SRAM address. In a database of multiple CYNSE70256s, each corresponding SADR bit from all cascaded devices must be connected.	
CE_L	Т	<b>SRAM Chip Enable</b> . This is the chip-enable (CE) control for external SRAMs. In a database of multiple CYNSE70256s, CE_L of all cascaded devices must be connected. This signal is then driven by only one of the devices.	
WE_L	Т	<b>SRAM Write Enable</b> . This is the Write-enable control for external SRAMs. In a database of multiple CYNSE70256s, WE_L of all cascaded devices must be connected. This signal is then driven by only one of the devices.	
OE_L	Т	<b>SRAM Output Enable</b> . This is the output-enable (OE) control for external SRAMs. Only the last device drives this signal (with the LRAM bit set).	
ALE_L	Т	Address Latch Enable. When this signal is LOW, the addresses are valid on the SRAM address bus. In a database of multiple CYNSE70256s, the ALE_L of all cascaded devices must be connected. This signal is then driven by only one of the devices.	
ascade Interface			
LHI[6:0]	I	Local Hit In. These pins depth-cascade the device to form a larger table. One signal of this bus is connected to the LHO[1] or LHO[0] of each of the upstream devices in a blocal unused LHI pins are connected to a logic 0. For more information, see Section 11 "Depth Cascading," on page 84. LHI[0] stays unconnected.	
LHO[1:0]	0	<b>Local Hit Out.</b> The LHO[1] and the LHO[0] are connected to one input on the LHI bus (from up to four downstream devices in a block totalling up to four). For more information, section 11.0, "Depth Cascading," on page 84.	
BHI[2:0]	1	<b>Block Hit In.</b> Inputs from the previous block BHO[2:0] are tied to BHI[2:0] of the current device. In a four-block system, the last block can contain only seven devices because the identification code 11111 is used for broadcast access.	
BHO[2:0]	0	<b>Block Hit Out</b> . These outputs from the last device in a block are connected to the BHI[2:0] inputs of the devices in the downstream blocks.	
FULI[6:0]	I	Full In. Each signal in this bus is connected to FULO[0] or FULO[1] of an upstream device to generate the FULL flag for the depth-cascaded block. FULI[0] stays unconnected.	
FULO[1:0]	0	Full Out. Both of these signals must be connected to the FULI of up to four downstream devices in a depth-cascaded table. Bit[0] in the data array indicates whether the entry is full (1) or empty (0). This signal is asserted if all bits in the data array are 1s. (Refer to Section 11.0, "Depth Cascading," on page 84, for information on how to generate the FULL flag.)	
FULL	0	Full Flag. When asserted, this signal indicates that the table of multiple depth-cascaded devices is full.	



Table 4-1. CYNSE70256 Signal Description (continued)

	Din		
Pin Name	Pin Type <sup>[1]</sup>	Pin Description	
Device Identification			
ID[4:0]	I	Device Identification. The binary-encoded device identification (ID[4:1]) for a depth-cascaded system starts at 0000 and goes up to 1110. 1111 is reserved for a special broadcast address that selects all cascaded NSEs in the system. On a broadcas Read-only, the device with the LDEV bit set to 1 responds. ID[0] stays unconnected.	
Supplies	•		
$V_{DD}$	n/a	Chip Core Supply: 1.5V.	
$V_{DDQ}$	n/a	Chip I/O Supply: 2.5V/3.3V.	
Test Access Port			
TDI	I	Test access port's test data in.	
TCK	Ĺ	Test access port's test clock.	
TDO	Т	Test access port's test data out.	
TMS	I	Test access port's test mode select.	
TRST_L	I	Test access port's reset.	

#### 5.0 Clocks

If the CLK\_MODE pin is LOW, CYNSE70256 receives the CLK2X and PHS\_L signals. It uses the PHS\_L signal to divide CLK2X and generate an internal clock ( $CLK^{[3,\ 4]}$ ), as shown in *Figure 5-1*. The CYNSE70256 uses CLK2X and CLK for internal operations.

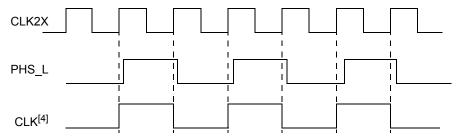


Figure 5-1. CYNSE70256 Clocks (CLK2X and PHS\_L)

If the CLK\_MODE pin is HIGH, CYNSE70256 receives CLK1X only. CYNSE70256 uses an internal phase-lock loop (PLL) to double the frequency of CLK1X and then divides that clock by two to generate a CLK for internal operations, as shown in Figure 5-2.

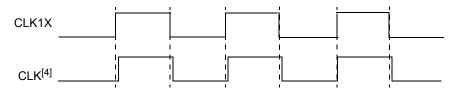


Figure 5-2. CYNSE70256 Clocks (CLK1X)<sup>[5]</sup>

### Notes:

- I = Input only, I/O = Input or Output, O = Output only, T = three-state output. ACK and EOT require a weak external pulldown such as  $47K\Omega$  or  $100K\Omega$ . Any reference to "CLK" cycles means one cycle of CLK. "CLK" is an internal clock signal.

- For the purpose of showing timing diagrams, all such diagrams in this document will be shown in CLK2X mode. For a timing diagram in CLK1X mode, the following substitution can be made (see *Figure 5-3*).



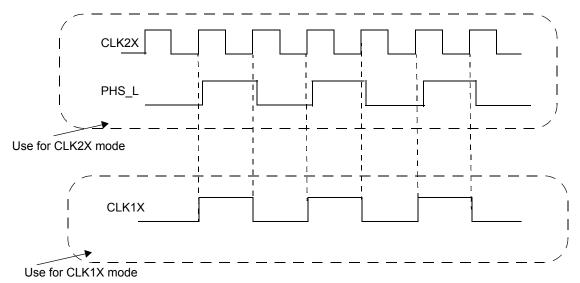


Figure 5-3. CYNSE70256 Clocks for All Timing Diagrams

# 6.0 Phase-Lock Loop Usage

When the device first powers up, it takes 0.5 ms to lock the internal PLL. During this PLL, the RST\_L must be held LOW for proper initialization of the device. It also takes 32 extra CLK1X cycles in CLK1X mode and 64 extra cycles in CLK2X mode. Setup and hold requirements will change in CLK1X mode if the duty cycle of the CLK1X is varied. All signals to the device in CLK1X mode are sampled by a clock that is generated by multiplying CLK1X by two. Since the PLL has a locking range, the device will only work between the range of frequencies specified in the timing specification section of this datasheet.

# 7.0 Registers

All registers in the CYNSE70256 device are 72 bits wide. The CYNSE70256 contains two banks of sixteen pairs of comparand storage registers, sixteen pairs of GMRs, eight search successful index registers, and one each of command, information, burst Read, burst Write, and next-free address registers. *Table 7-1* provides an overview of all the CYNSE70256 registers. The registers are in ascending address order; each register group is described in the following subsections.

Address	Abbreviation	Type	Name
0–31	COMP0-31	R	Sixteen pairs of comparand registers that store comparands from the DQ bus for learning later.
32–47 96–111	MASKS	RW	Sixteen GMR pairs.
48–55	SSR0-7	R	Eight search successful index registers.
56	COMMAND	RW	Command register.
57	INFO	R	Information register.
58	RBURREG	RW	Burst Read register.
59	WBURREG	RW	Burst Write register.
60	NFA	R	Next-free address register.
61–63	_	_	Reserved.

Table 7-1. Register Overview (Bank0 and Bank1)

# 7.1 Comparand Registers

The device contains two banks of 32 72-bit comparand registers (sixteen pairs) dynamically selected in every Search operation to store the comparand presented on the DQ bus. The Learn command will later use these registers when it is executed. The CYNSE70256 device stores the Search command's cycle A comparand in the even-numbered register and the cycle B comparand in the odd-numbered register, as shown in *Figure 7-1* for each of the two banks of registers.



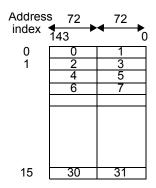


Figure 7-1. Comparand Register Selection during Search and Learn Instructions

# 7.2 Mask Registers

The device contains two banks of 32 72-bit GMRs (sixteen pairs) dynamically selected in every Search operation to select the Search subfield. The addressing of these registers is explained in *Figure 7-2*. The 4-bit GMR index supplied on the command bus can apply sixteen pairs of global masks during Search and Write operations, as shown below.<sup>[6]</sup>

	72	72
Index	143	Ó
0	0	1
1	2	3
2	4	5
3	6	7
4	8	9
5	10	11
6	12	13
7	14	15
8	16	17
9	18	19
10	20	21
11	22	23
12	24	25
13	26	27
14	28	29
15	30	31

Search and Write Command Global Mask Selection

Figure 7-2. Addressing the GMR Array

Each mask bit in the GMRs is used during Search and Write operations. In Search operations, setting the mask bit to 1 enables compares; setting the mask bit to 0 disables compares at the corresponding bit position (forced match). In Write operations to the data or mask array, setting the mask bit to 1 enables Writes; setting the mask bit to 0 disables Writes at the corresponding bit position.

# 7.3 Search Successful Registers (SSR[0:7])

The device contains two banks of eight search successful registers (SSRs) to hold the index of the location at which a successful search occurred. The format of each register is described in *Table 7-2*. The Search command specifies which SSR stores the index of a specific Search command in cycle B of the Search instruction. Subsequently, the host ASIC can use this register to access that data array, mask array, or external SRAM using the index as part of the indirect access address (see *Table 10-3* and *Table 10-6*).

The device with a valid bit set performs a Read or Write operation. All other devices suppress the operation.

#### Note

6. In 72-bit Search and Write operations, the host ASIC must program both the even and odd mask registers with the same values for each of the banks.



Table 7-2. Search Successful Register Description

Field	Range	Initial Value	Description
INDEX	[15:0]	Х	Index. This is the address of the 72-bit entry where a successful search occurs. The device updates this field only when the search is successful. If a hit occurs in a 144-bit entry-size quadrant, the least-significant bit (LSB) is 0. If a hit occurs in a 288-bit entry-size quadrant, the two LSBs are 00. This index updates if the device is either a local or global winner in a Search operation.
_	[30:16]	0	Reserved.
VALID	[31]	0	<i>Valid.</i> During Search operation in a depth-cascaded configuration, this ban of the device that is a global winner in a match sets this bit to 1. This bit updates only when the device is a global winner in a Search operation.
1	[71:32]	0	Reserved.

# 7.4 Command Register

Table 7-3 describes the command registers' fields for each of the two banks; Bank 0 and Bank 1.

Table 7-3. Command Register Description

Field	Range	Initial Value	Description
SRST	[0]	0	<b>Software Reset</b> . If 1, this bit resets the bank with the same effect as a hardware reset. Internally, it generates a reset pulse lasting for eight CLK cycles. This bit automatically resets to a 0 after the reset has completed.
DEVE	[1]	0	<b>Device Enable</b> . If 0, it keeps the SRAM bus (SADR, WE_L, CE_L, OE_L and ALE_L), SSF, and SSV signals in a three-state condition and forces the cascade interface output signals LHO[1:0] and BHO[2:0] to 0. It also keeps the DQ bus in input mode. The purpose of this bit is to make sure that there are no bus contentions when the devices power up in the system.
TLSZ	[3:2]	01	Table Size. The host ASIC must program this field to configure each bank into a table of a certain size. This field affects the pipeline latency of the Search and Learn operations as well as the Read and Write accesses to the SRAM (SADR[23:0], CE_L, OE_L, WE_L, ALE_L, SSV, SSF, and ACK). Once programmed, the Search latency stays constant.  Latency in number of CLK cycles with HIGH_SPEED LOW:  01: Up to four devices  5  10: Up to fifteen devices  6  11: Reserved.
HLAT	[6:4]	000	Latency of Hit Signals. This field adds further latency to the SSF and SSV signals during Search, and ACK signal during SRAM Read access by the following number of CLK cycles.  000: 0 100: 4  001: 1 101: 5  010: 2 110: 6  011: 3 111: 7.
LDEV	[7]	0	Last Device in the Cascade. When set, this is the last bank of the last device in the depth-cascaded table and is the default driver for the SSF and SSV signals. In the event of a Search failure, the bank with this device with this bit set drives the hit signals as follows: SSF = 0, SSV = 1. During non-Search cycles, the device with this bit set drives the signals as follows: SSF = 0, SSV = 0.
LRAM	[8]	0	Last Device on the SRAM Bus. When set, this is the last bank of the last device on the SRAM bus in the depth-cascaded table and is the default driver for the SADR, CE_L, WE_L, and ALE_L signals. In cycles where no CYNSE70256 device in a depth-cascaded table drives these signals, this drives the signals as follows: SADR = 23'hFFFFFF, CE_L = 1, WE_L = 1, and ALE_L = 1. OE_L is always driven by the device for which this bit is set.



Table 7-3. Command Register Description (continued)

Field	Range	Initial Value	Description
CFG	[24:9]	000000000	Database Configuration. The device is divided internally into two banks each consisting of sixteen partitions of 8K × 72, each of which can be configured as 8K × 72, 4K × 144, or 2K × 288, as follows.  00: 8K × 72 01: 4K × 144 10: 2K × 288 11: LOW power, partition not used for Search.  Bits[10:9] apply to configuring the first partition in the address space.  Bits[12:11] apply to configuring the second partition in the address space.  Bits[14:13] apply to configuring the third partition in the address space.  Bits[16:15] apply to configuring the fourth partition in the address space.  Bits[18:17] apply to configuring the fifth partition in the address space.  Bits[20:19] apply to configuring the sixth partition in the address space.  Bits[22:21] apply to configuring the seventh partition in the address space.  Bits[24:23] apply to configuring the eighth partition in the address space.
	[71:25]	0	Reserved.

# 7.5 Information Register

Table 7-4 describes the information register fields for both banks.

Table 7-4. Information Register Description

Field	Range	Initial Value	Description
Revision	[3:0]	0001	Revision Number. This is the current device revision number. Numbers start at one and increment by one for each revision of the device.
Implementation	[6:4]	001	This is the CYNSE70256 implementation number.
Reserved	[7]	0	Reserved.
Device ID	[15:8]	00000100	This is the device identification number.
MFID	[31:16]	1101_1100_0111_1111	<b>Manufacturer ID</b> . This field is the same as the manufacturer identification number and continuation bits in the TAP controller.
Reserved	[71:32]		Reserved.

# 7.6 Read Burst Address Register

Table 7-5 shows the Read burst address register (RBURREG) fields that must be programmed before a burst Read when a Read burst transfer is done from any bank.

Table 7-5. Read Burst Register Description

Field	Range	Initial Value	Description
ADR	[15:0]	0	Address. This is the starting address of the data or mask array during a burst-Read operation from a bank. It automatically increments by one for each successive Read of the data or mask array. Once the operation is complete, the contents of this field must be reinitialized for the next operation.
	[18:16]		Reserved.
BLEN	[27:19]	0	Length of Burst Access. The device provides the capability to Read from 4–511 locations in a single burst from each bank. The BLEN decrements automatically. Once the operation is complete, the contents of this field must be reinitialized for the next operation.
	[71:28]		Reserved.

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# 7.7 Write Burst Address Register Description

Table 7-6 describes the Write burst address register (WBURREG) fields that must be programmed before a burst Write.

Table 7-6. Write Burst Register Description

Field	Range	Initial Value	Description
ADR	[15:0]	0	Address. This is the starting address of the data or mask array during a burst-Write operation from a bank. It automatically increments by one for each successive Write of the data or mask array. Once the operation is complete, the contents of this field must be reinitialized for the next operation.
	[18:16]		Reserved.
BLEN	[27:19]	0	Length of Burst Access. The device provides the capability to Write from 4–511 locations in a single burst. The BLEN decrements automatically. Once the operation is complete, the contents of this field must be reinitialized for the next operation.
	[71:28]		Reserved.

# 7.8 NFA Register

Bit[0] of each 72-bit data entry is specially designated for use in the operation of the Learn command in each of the banks. For 72-bit-configured quadrants, this bit indicates whether a location is full (bit set to 1) or empty (bit set to 0). Every Write and/or Learn command loads the address of the first 72-bit location that contains a 0 in the entry's bit[0]. This is stored in the NFA register(see *Table 7-7*). If all the bits[0] in a device for both the banks within the device are set to 1, the CYNSE70256 asserts FULO[1:0] to 1.

For 144-bit-configured quadrants, the LSB of the NFA register is always set to 0. The host ASIC must set both bit[0] and bit[72] in a 144-bit word to either 0 or 1 to indicate full or empty status. Both bit[0] and bit[72] must be set to either 0 or 1, (that is, the 10 or 01 settings are invalid).

Table 7-7. NFA Register

Address	71–16	15–0
60	Reserved	Index

# 8.0 NSE Architecture and Operation Overview

The CYNSE70256 device consists of two banks of 64K × 72-bit storage cells referred to as data bits. There is a mask cell corresponding to each data cell. *Figure 8-1* shows the three organizations of the device based on the value of the CFG bits in the command register.

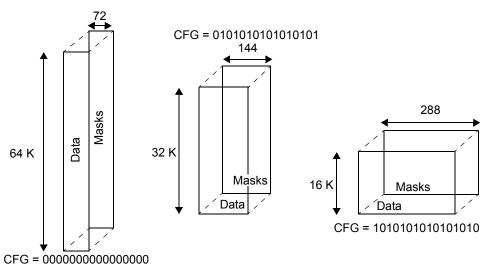


Figure 8-1. CYNSE70256 Database Width Configuration for Each of the Two Banks

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During a Search operation, the search data bit (S), data array bit (D), mask array bit (M), and global mask bit (G) are used in the following manner to generate a match at that bit position (see *Table 8-1*). The entry with a match on every bit position results in a successful Search.

Table 8-1. Bit Position Match

G	М	D	S	Match
0	Х	X	X	1
1	0	X	X	1
1	1	0	0	1
1	1	1	0	0
1	1	0	1	0
1	1	1	1	1

In order for a successful Search within a device to make the device the local winner, all 72-bit positions must generate a match for a 72-bit entry in 72-bit-configured quadrants, or all 144-bit positions must generate a match for two consecutive even and odd 72-bit entries in quadrants configured as 144 bits, or all 288-bit positions must generate a match for four consecutive entries aligned to four entry-page boundaries of 72-bit entries in quadrants configured as 288 bits.

An arbitration mechanism using a cascade bus determines the global winning device among the local winning devices in a Search cycle. The global winning device drives the SRAM bus, the SSV, and the SSF signals. In case of a Search failure, the device(s) with a bank with the LDEV and LRAM bits set drives the SRAM bus, SSF, and SSV signals.

The CYNSE70256 device can be configured to contain tables of different widths, even within the same chip. *Figure 8-2* shows a sample configuration of different widths.

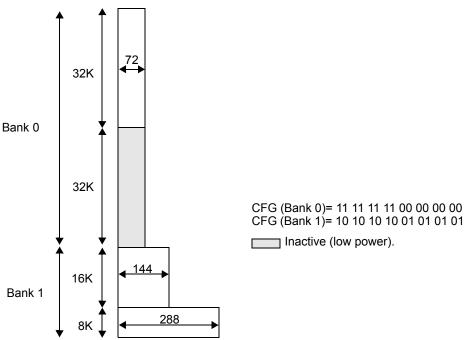
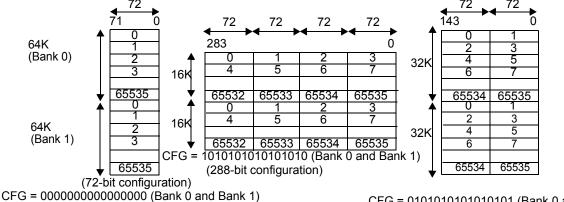


Figure 8-2. Multiwidth Database Configurations Example



#### 9.0 **Data and Mask Addressing**

Figure 9-1 shows CYNSE70256 data and mask array addressing for both Bank 0 and Bank 1.



CFG = 0101010101010101 (Bank 0 and Bank 1) (144-bit configuration)

Figure 9-1. Addressing the CYNSE70256 Data and Mask Arrays

#### 10.0 Commands

A master device such as an ASIC controller issues commands to the CYNSE70256 device using the CMDV signal and the command bus. The following subsections describe the operation of these commands.

#### 10.1 **Command Codes**

The CYNSE70256 device implements four basic commands, as shown in Table 10-1. The command code must be presented to CMD[1:0] while keeping the CMDV signal HIGH for two CLK2X cycles (cycles A and B) when the CLK MODE pin is LOW. In CLK2X mode, the controller ASIC must align the instructions using the PHS L signal. The command code must be presented to CMD[1:0] while keeping the CMDV signal HIGH for one CLK1X cycle when the CLK\_MODE pin is HIGH. In CLK1X mode the HIGH phase is cycle A and the LOW phase is cycle B. The CMD[10:2] field passes command parameters in cycles A and B.

Table 10-1. Command Codes

Command Code	Command	Description
00	Read	Reads one of the following: data array, mask array, device registers, or external SRAM.
01	Write	Writes one of the following: data array, mask array, device registers, or external SRAM.
10	Search	Searches the data array for a desired pattern using the specified register from the GMR array and local mask associated with each data cell.
11	Learn	The device has internal storage for up to sixteen comparands that it can learn. The device controller can insert these entries at the next-free address (as specified by the NFA register) using the Learn instruction.

#### 10.2 **Commands and Command Parameters**

Table 10-2 lists the command bus fields that contain the CYNSE70256 command parameters and their respective cycles. Each command is described separately in the subsections that follow.



**Table 10-2. Command Parameters** 

CMD <sup>[7,8]</sup>	CYC	10	9	8	7	6	5	4	3	2	1	0
Read	Α	Х	Х	SADR[23]	SADR[22]	SADR[21]	0	0	0	0 = Single 1 = Burst	0	0
	В	Х	Х	0	0	0	0	0	0	0 = Single 1 = Burst	0	0
Write	Α	GMR Index <sup>[9]</sup>	0: Normal Write 1: Parallel Write	SADR[23]	SADR[22]	SADR[21]	_	R Ind [2:0]	lex	0 = Single 1 = Burst	0	1
	В	GMR Index <sup>[9]</sup>	0: Normal Write 1: Parallel Write	0	0	0	_	R Ind [2:0]	lex	0 = Single 1 = Burst	0	1
Search	Α	GMR Index <sup>[9]</sup>	72 bits: 0 144 bits: 1 288 bits: X	SADR[23]	SADR[22]	SADR[21]	_	R Ind [2:0]	lex	72 bits or 144 bits: 0 288 bits: 1 in first cycle 0 in second cycle	1	0
	В	Х		S	SR Index[2:	0]	Comparand Register Index		1	0		
Learn <sup>[9]</sup>	Α	Х	Х	SADR[23]	SADR[22]	SADR[21]	Comparand Register Index		1	1		
	В	Х	Х	0	0	Mode 0: 72 bits 1: 144 bits	Comparand Register Index		1	1		

#### 10.3 **Read Command**

The Read can be a single Read of a data array, a mask array, an SRAM, or a register location (CMD[2] = 0). It can be a burst Read of the data (CMD[2] = 1) or mask array locations using an internal auto-incrementing address register (RBURADR). A description of each type is provided in Table 10-3. A single-location Read operation lasts six cycles, as shown in Figure 10-1. The burst Read adds two cycles for each successive Read. The SADR[23:21] bits supplied in Read instruction cycle A drives SADR[23:21] signals during a Read of an SRAM location.

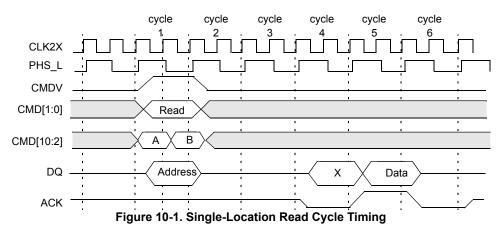
Table 10-3. Read Command Parameters

CMD Parameter CMD[2]	Read Command	Description
0		Reads a single location of the data array, mask array, external SRAM, or device registers. All access information is applied on the DQ bus.
1		Reads a block of locations from the data or mask array as a burst. RBURADR specifies the starting address and the length of the data transfer from the data or mask array; it also auto-increments the address for each access. All other access information is applied on the DQ bus. <sup>[10]</sup>

#### Notes:

- Use CMD[8:0] only and connect CMD[10:9] to ground with CFG\_L LOW.
   For a description of CMD[9] and CMD[2], see Search 288-bit-configured tables and mixed-size searches with CFG\_L HIGH.
   The 288-bit-configured devices or 288-bit-configured quadrants within devices do not support the Learn instruction.
   The device registers and external SRAM can only be read in single-Read mode.





The single Read operation takes six CLK cycles that operate in the following sequence.

- Cycle 1: The host ASIC applies the Read instruction on CMD[1:0] (CMD[2] = 0) using CMDV = 1, and the DQ bus supplies the address, as shown in Table 10-4 and Table 10-5. The host ASIC selects the CYNSE70256 device for which ID[4:1] matches the DQ[25:22] lines. The DQ[21] specifies the bank of the device. If DQ[25:21] = 11111, the host ASIC selects the CYNSE70256 with the LDEV bit set. The host ASIC also supplies SADR[23:21] on CMD[8:6] in cycle A of the Read instruction if the Read is directed to the external SRAM.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive the DQ[71:0] bus and drives the ACK signal from Z to LOW.
- Cycle 5: The selected device drives the Read data from the addressed location on the DQ[71:0] bus, and drives the ACK signal HIGH.
- Cycle 6: The selected device floats the DQ[71:0] to a three-state condition and drives the ACK signal LOW.

At the termination of cycle 6, the selected device releases the ACK line to a three-state condition. The Read instruction is complete, and a new operation can begin [11]

Table 10-5 describes the Read address format for the internal registers. Figure 10-2 illustrates the timing diagram for the burst Read of the data or mask array.

Table 10-4. Read Address Format for Data Array, Mask Array, or SRAM

DQ [71:30]	DQ [29]	DQ [28:26]	DQ [25:22]	DQ [21]	DQ [20:19]	DQ [18:16]	DQ [15:0]
Reserved	0: Direct 1: Indirect	SSR Index (applicable if DQ[29] is indirect)	ID	Bank 0 or 1	00: Data Array		If DQ[29] is 0, this field carries the address of the data array location. If DQ[29] is 1, the SSR index specified on DQ[28:26] is used to generate the address of the data array location: {SSR[15:2], SSR[1]   DQ[1], SSR[0]   DQ[0]}. [12]
Reserved	0: Direct 1: Indirect	SSR Index (applicable if DQ[29] is indirect)	ID	Bank 0 or 1	01: Mask Array	Reserved	If DQ[29] is 0, this field carries the address of the mask array location. If DQ[29] is 1, the SSR index specified on DQ[28:26] is used to generate the address of the mask array location: {SSR[15:2], SSR[1]   DQ[1], SSR[0]   DQ[0]}. [12]
Reserved	0: Direct 1: Indirect	SSR Index (applicable if DQ[29] is indirect)	ID	Bank 0 or 1	10: External SRAM		If DQ[29] is 0, this field carries the address of the SRAM location. If DQ[29] is 1, the SSR index specified on DQ[28:26] is used to generate the address of the SRAM location: {SSR[15:2], SSR[1]   DQ[1], SSR[0]   DQ[0]}.

#### Notes:

The latency of the SRAM Read will be different than the one described above (see Subsection 12.1, "SRAM PIO Access," on page 86). *Table 10-4* lists and describes the format of the Read address for a data array, mask array, or SRAM. "|" stands for logical OR operation. "{}" stands for concatenation operator.



Table 10-5. Read Address Format for Internal Registers

DQ[71:26]	DQ[25:22]	DQ[21]	DQ[20:19]	DQ[18:7]	DQ[6:0]
Reserved	ID	Bank 0 or 1	11: Register	Reserved	Register Address

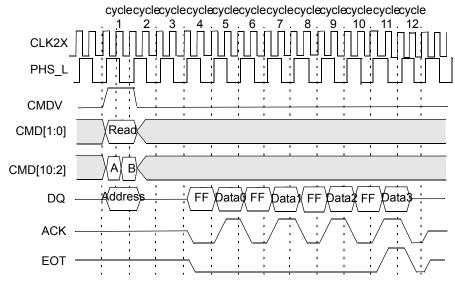


Figure 10-2. Burst Read of the Data and Mask Arrays (BLEN = 4)

The Read operation lasts 4 + 2n CLK cycles (where n is the number of accesses in the burst specified by the BLEN field of the RBURREG) in the sequence shown below. This operation assumes that the host ASIC has programmed the RBURREG of the appropriate bank of the device with the starting ADR and the BLEN before initiating the burst Read command for the appropriate bank in the appropriate device.

- Cycle 1: The host ASIC applies the Read instruction on CMD[1:0] (CMD[2] = 1) using CMDV = 1, and the address supplied on the DQ bus as shown in *Table 10-6*. The host ASIC selects the bank 0 or 1 (Based on Bank Bit) of the CYNSE70256 device where ID[4:1] matches the DQ[25:22] lines. DQ[21] specfies the bank of the device that is written. If DQ[25:21] = 11111 the host ASIC selects the bank of the CYNSE70256 device with the LDEV bit set.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive the DQ[71:0] bus and drives ACK and EOT from Z to LOW.
- Cycle 5: The selected device drives the Read data from the address location on the DQ[71:0] bus and drives the ACK signal HIGH.

Cycles 4 and 5 repeat for each additional access until all the accesses specified in the BLEN field of RBURREG are complete. On the last transfer, the CYNSE70256 device drives the EOT signal HIGH.

Cycle (4 + 2n): The selected device drives the DQ[71:0] to a three-state condition, and drives the ACK and EOT signals LOW.

At the termination of cycle (4 + 2n), the selected device floats the ACK line to a three-state condition. The burst Read instruction is complete, and a new operation can begin. *Table 10-6* describes the Read address format for data and mask arrays for burst Read operations.

Table 10-6. Read Address Format for Data and Mask Arrays

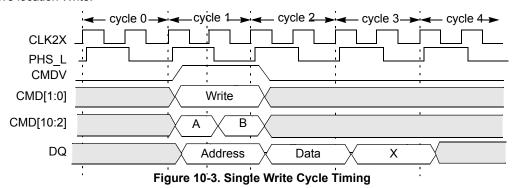
DQ[71:26]	DQ[25:22]	DQ[21]	DQ[20:19]	DQ[18:16]	DQ[15:0]
Reserved	ID	Bank 0 or 1	00: Data Array		<b>Do not care</b> . These seventeen bits come from the RBURADR, which increments for each access.
Reserved	ID	Bank 0 or 1	01: Mask Array		<b>Do not care.</b> These seventeen bits come from the RBURADR, which increments for each access.

# 10.4 Write Command

The Write command can be a single Write of a data array, mask array, register, or external SRAM location (CMD[2] = 0). It can be a burst Write (CMD[2] = 1) using an internal auto-incrementing address registers (WBURADR) of the data or mask array



locations. A single-location Write is a three-cycle operation, as shown in Figure 10-3. The burst Write adds one extra cycle for each successive location Write.



The following is the Write operation sequence. Table 10-7 shows the Write address format for the data array, the mask array, or single-Write SRAM. Table 10-8 shows the Write address format for the internal registers.

- Cycle 1A: The host ASIC applies the Write instruction to CMD[1:0] (CMD[2] = 0) using CMDV = 1, and the address supplied on the DQ bus. The host ASIC also supplies the GMR index to mask the Write to the data or mask array location on {CMD[10], CMD[5:3]). For SRAM Writes, the host ASIC must supply the SADR[23:21] on CMD[8:6]. The host ASIC sets CMD[9] to 0 for a normal Write.
- Cycle 1B:The host ASIC continues to apply the Write instruction to CMD[1:0] (CMD[2] = 0) using CMDV = 1, and the address supplied on the DQ bus. The host ASIC continues to supply the GMR index to mask the Write to the data or mask array locations in {CMD[10], CMD[5:3]}. The host ASIC selects the device where ID[4:1] matches the DQ[25:22] lines and the bank within the device using value on DQ[21], or it selects both banks of all the devices when DQ[25:21] = 11111.
- Cycle 2: The host ASIC drives DQ[71:0] with the data to be written to the data array, mask array, or register location of the selected device.
- · Cycle 3: Idle cycle.

At the termination of cycle 3, another operation can begin.<sup>[13]</sup>

Table 10-7. Write Address Format for Data Array, Mask Array, or SRAM (Single Write)

DQ [71:30]	DQ [29]	DQ [28:26]	DQ [25:22]	DQ [21]	DQ [20:19]	DQ [18:16]	DQ [15:0]
Reserved	0: Direct 1: Indirect	SSR (applicable if DQ[29] is indirect)	ID	Bank 0 or 1	00: Data Array	Reserved	If DQ[29] is 0, this field carries the address of the data array location.  If DQ[29] is 1, the SSR specified on DQ[28:26] is used to generate the address of data array location: {SSR[15:2], SSR[1]   DQ[1], SSR[0]   DQ[0]}.
Reserved	0: Direct 1: Indirect	SSR (applicable if DQ[29] is indirect)	ID	Bank 0 or 1	01: Mask Array	Reserved	If DQ[29] is 0, this field carries the address of the mask array location.  If DQ[29] is 1, the SSR specified on DQ[28:26] is used to generate the address of the mask array location: {SSR[15:2], SSR[1]   DQ[1], SSR[0]   DQ[0]}.
Reserved	1: Indirect	SSR (applicable if DQ[29] is indirect)	ID	Bank 0 or 1	10: External SRAM	Reserved	If DQ[29] is 0, this field carries the address of the SRAM location. If DQ[29] is 1, the SSR specified on DQ[28:26] is used to generate the address of SRAM location: {SSR[15:2], SSR[1]   DQ[1], SSR[0]   DQ[0]}.

Table 10-8. Write Address Format for Internal Registers

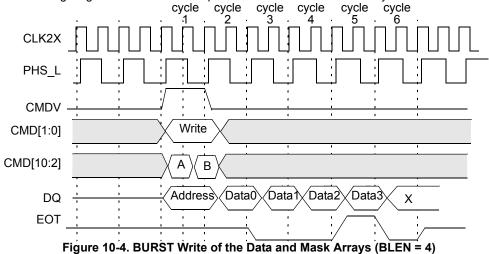
DQ[71:26]	DQ[25:22]	DQ[21]	DQ[20:19]	DQ[18:7]	DQ[6:0]
Reserved	ID	Bank 0 or 1	11: Register	Reserved	Register address

## Notes:

The latency of the SRAM Write will be different than the one described above (see Subsection 12.1, "SRAM PIO Access," on page 86). "|" stands for logical OR operation. "()" stands for concatenation operator.



Figure 10-4 shows the timing diagram of a burst Write operation of the data or mask array.



The burst Write operation lasts for (n + 2) CLK cycles. n signifies the number of accesses in the burst as specified in the BLEN field of the WBURREG register. The following is the block Write operation sequence. This operation assumes that the host ASIC has programmed the WBURREG of the appropriate bank with the starting ADR and BLEN before initiating a burst Write command.

- Cycle 1A: The host ASIC applies the Write instruction to CMD[1:0] (CMD[2] = 1) using CMDV = 1, and the address supplied on the DQ bus as shown in *Table 10-9*. The host ASIC also supplies the GMR index to mask the Write to the data or mask array locations in {CMD[10], CMD[5:3]}. The host ASIC sets CMD[9] to 0 for the normal Write.
- Cycle 1B: The host ASIC continues to apply the Write instruction on CMD[1:0] (CMD[2] = 1) using CMDV = 1, and the address supplied on the DQ bus. The host ASIC continues to supply the GMR index to mask the Write to the data or mask array locations in {CMD[10], CMD[5:3]}. The host ASIC selects the device for which ID[4:1] matches the DQ[25:22] lines and the bank of the device using DQ[21] lines. It selects all devices when DQ[25:21] = 11111.
- Cycle 2: The host ASIC drives the DQ[71:0] with the data to be written to the data or mask array location of the selected device. The CYNSE70256 device writes the data from the DQ[71:0] bus only to the subfield with the corresponding mask bit set to 1 in the GMR that is specified by the index {CMD[10],CMD[5:3]} supplied in cycle 1.
- Cycles 3 to n + 1: The host ASIC drives the DQ[71:0] with the data to be written to the next data or mask array location of the selected device (addressed by the auto-increment ADR field of the WBURREG register).

The CYNSE70256 device writes the data on the DQ[71:0] bus only to the subfield that has the corresponding mask bit set to 1 in the GMR specified by the index supplied in cycle 1 {CMD[10],CMD[5:3]}. The CYNSE70256 device drives the EOT signal LOW from cycle 3 to cycle n; the CYNSE70256 device drives the EOT signal HIGH in cycle n + 1 (n is specified in the BLEN field of the WBURREG).

• Cycle n + 2: The CYNSE70256 device drives the EOT signal LOW.

At the termination of cycle n + 2, the CYNSE70256 device floats the EOT signal to a three-state operation, and a new instruction can begin.

Table 10-9. Write Address Format for Data and Mask Array (Burst Write)

DQ [71:26]	DQ [25:22]	DQ [21]	DQ [20:19]	DQ [18:16]	DQ [15:0]
Reserved	ID	Bank 0 or 1	00: Data array		Do not care. These seventeen bits come from WBURADR, which increments with each access.
Reserved	ID	Bank 0 or 1	01: Mask array	Reserved Do not care. These seventeen bits co WBURADR, which increments with e	

### 10.5 Parallel Write

In order to write the data and mask arraysof both banks faster for initialization, testing, or diagnostics, many locations can be written simultaneously in the CYNSE70256 device. When CMD[9] is set in cycles A and B of the Write command during a Write to the data or mask arrays, the address present on DQ[10:1] that specifies 128 locations in the device is used, and 64 72-bit locations are simultaneously written in either the data or mask array. Setting DQ[21] to 0 will cause a write to the addresses, specified by DQ[10:1], closer to address 0, while setting DQ[21] to 1 will cause a write to the addresses, also specified by DQ[10:1], further from address 0.



## 10.6 Search Command

This subsection describes the following.

- 72-bit search on tables configured as ×72 using up to four devices
- 72-bit search on tables configured as ×72 using up to fifteen devices
- 144-bit search on tables configured as ×144 using up to four devices
- 144-bit search on tables configured as ×144 using up to fifteen device
- 288-bit search on tables configured as ×288 using up to four devices
- 288-bit search on tables configured as ×288 using up to fifteen devices
- Mixed-size search on tables configured with different widths using an CYNSE70256 with CFG L LOW
- Mixed-size searches on tables configured with different widths using an CYNSE70256 with CFG\_L HIGH.

# 10.6.1 72-bit Search on Tables Configured as ×72 using up to Four CYNSE70256 Devices

The hardware diagram of the Search subsystem of four devices is shown in *Figure 10-5*. The following are the parameters programmed into each bank of the the four devices.

- First three devices (devices 0-2, both banks): CFG = 000000000000000, TLSZ = 01, HLAT = 010, LRAM = 0, and LDEV = 0.
- Fourth device (device 3, Bank 0): CFG = 000000000000000, TLSZ = 01, HLAT = 010, LRAM = 0, and LDEV = 0
- Fourth device (device 3, Bank 1): CFG = 000000000000000, TLSZ = 01, HLAT = 010, LRAM = 1, and LDEV = 1.[15]

Figure 10-6 shows the timing diagram for a Search command in the 72-bit-configured table of four devices for device number 0. Figure 10-7 shows the timing diagram for a Search command in the 72-bit-configured table of four devices for device number 1. Figure 10-8 shows the timing diagram for a Search command in the 72-bit-configured table of four devices for device number 7 (the last device in this specific table). For these timing diagrams four 72-bit searches are performed sequentially. HIT/MISS assumptions were made as shown below in Table 10-10.

Table 10-10. Hit/Miss Assumptions

Search Number	1	2	3	4
Device 0	Hit	Miss	Hit	Miss
Device 1	Miss	Hit	Hit	Miss
Device 2	Miss	Miss	Miss	Miss
Device 3	Miss	Miss	Hit	Hit

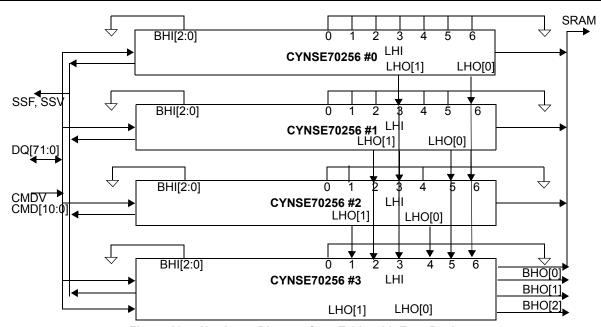


Figure 10-5. Hardware Diagram for a Table with Four Devices

#### Note:

<sup>15.</sup> Each bank of the four devices must be programmed with the same values for TLSZ and HLAT. Only the the last bank of the last device in the table (device number 3 in this case) must be programmed with LRAM = 1 and LDEV = 1.



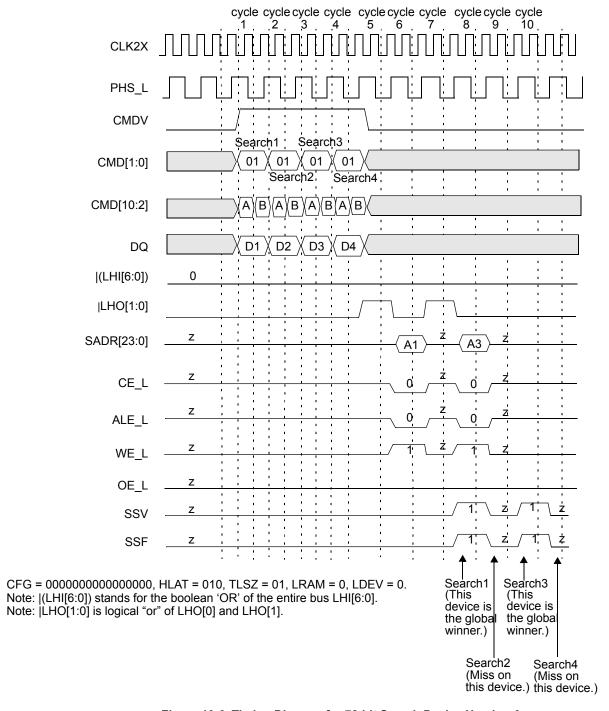


Figure 10-6. Timing Diagram for 72-bit Search Device Number 0



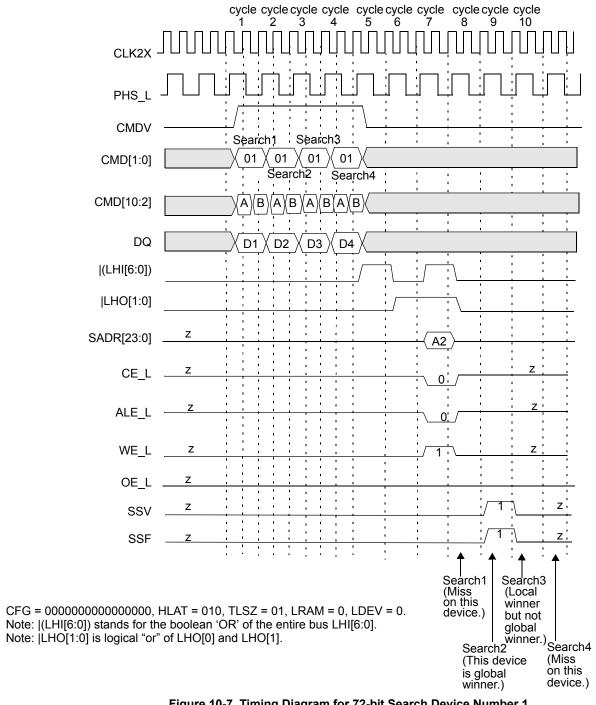


Figure 10-7. Timing Diagram for 72-bit Search Device Number 1



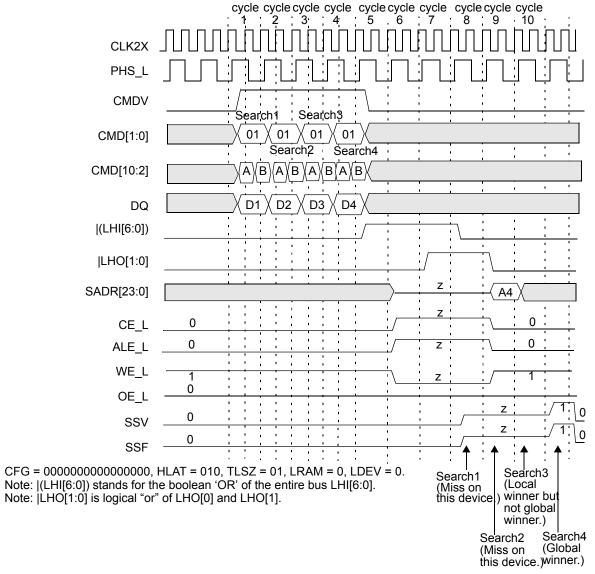


Figure 10-8. Timing Diagram for 72-bit Search Device Number 3 (Last Device)

The following is the sequence of operation for a single 72-bit Search command (also refer to Subsection 10.2, "Commands and Command Parameters," on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) to CMD[1:0] signals. {CMD[10],CMD[5:3]} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the same bits that will be driven on SADR[23:21] by this device if it has a hit. DQ[71:0] must be driven with the 72-bit data to be compared. The CMD[2] signal must be driven to logic 0.
- Cycle B: The host ASIC continues to drive CMDV HIGH and to apply Search command code (10) on CMD[1:0]. CMD[5:2] must be driven by the index of the comparand register pair for storing the 144-bit word presented on the DQ bus during cycles A and B. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and hit flag (see page 14 for a description of SSR[0:7]). The DQ[71:0] continues to carry the 72-bit data to be compared.

**Note.** For 72-bit searches, the host ASIC must supply the same 72-bit data on DQ[71:0] during both cycles A and B. Also, the even and odd pairs of GMRs selected for the comparison must be programmed with the same value.

The logical 72-bit Search operation is shown in *Figure 10-9*. The entire table of 72-bit entries (four devices) is compared to a 72-bit word K (presented on the DQ bus in both cycles A and B of the command) using the GMR and local mask bits. The effective GMR is the 72-bit word specified by the identical value in both even and odd GMR pairs, in each of two banks of the four devices, and selected by the GMR Index in the command's cycle A. The 72-bit word K (presented on the DQ bus in both cycles A and B of the command) is also stored in both even and odd comparand register pairs (selected by the comparand register index in



command cycle B) in each of the four devices. In the ×72 configuration, only the even comparand register can subsequently be used by the Learn command in one of the devices (the first non-full device only). The word K (presented on the DQ bus in both cycles A and B of the command) is compared with each entry in the table, starting at location 0. The first matching entry's location address L is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (see Section 12.0, "SRAM Addressing," on page 86). The global winning device will drive the bus in a specific cycle. On a global miss cycle, the device with LRAM = 1 (default driving device for the SRAM bus) and LDEV = 1 (default driving device for SSF and SSV signals) will be the default driver for such missed cycles.

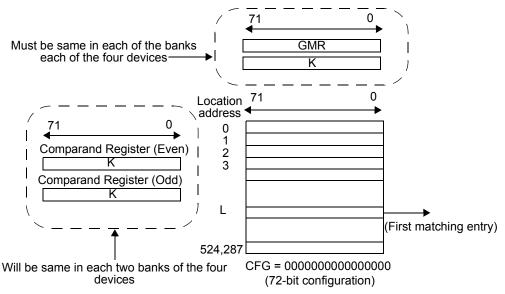


Figure 10-9. ×72 Table with Four Devices

The Search command is a pipelined operation and executes a Search at half the rate of the frequency of CLK2X for 72-bit searches in ×72-configured tables. The latency of SADR, CE\_L, ALE\_L, WE\_L, SSV, and SSF from the 72-bit Search command cycle (two CLK2X cycles) is shown in *Table 10-11*.

Table 10-11. Search Latency from Instruction to SRAM Access Cycle

	Number of Devices	Max Table Size	Latency in CLK Cycles
	1–4 (TLSZ = 01)	512K × 72 bits	5
Ī	1–15 (TLSZ = 10)	1920K × 72 bits	6

The latency of the Search from command to SRAM access cycle is 5 for up to four devices in the table (TLSZ = 01). SSV and SSF also shift further to the right for different values of HLAT, as specified in *Table 10-12*.

Table 10-12. Shift of SSF and SSV from SADR

HLAT	Number of CLK Cycles
000	0
001	1
010	2
011	3
100	4
101	5
110	6
111	7

# 10.6.2 72-bit Search on Tables Configured as ×72 using up to Fifteen CYNSE70256 Devices

The hardware diagram of the Search subsystem of fifteen devices is shown in *Figure 10-10*. Each of the four blocks in the diagram represents four CYNSE70256 devices (except the last, which has three devices). The diagram for a block of four devices is shown in *Figure 10-11*. The following are the parameters programmed into the fifteen devices.



- First thirty devices (devices 0–13, both banks): CFG = 00000000000000, TLSZ = 10, HLAT = 001, LRAM = 0, and LDEV = 0.
- First thirty devices (device 14, Bank 0): CFG = 000000000000000, TLSZ = 10, HLAT = 001, LRAM = 0, and LDEV = 0.
- Thirty-first device (device 14, Bank 1): CFG = 00000000000000, TLSZ = 10, HLAT = 001, LRAM = 1, and LDEV = 1.

**Note.** All fifteen devices must be programmed with the same values for TLSZ and HLAT. Only the last bank of the last device in the table must be programmed with LRAM = 1 and LDEV = 1 (device number 14 in this case).

The timing diagrams referred to in this paragraph reference the Hit/Miss assumptions defined in *Table 10-13*. For the purpose of illustrating the timings, it is further assumed that the there is only one device with a matching entry in each of the blocks. *Figure 10-12* shows the timing diagram for a Search command in the 72-bit-configured table of fifteen devices for each of the four devices in block number 0. *Figure 10-13* shows the timing diagram for a Search command in the 72-bit-configured table of fifteen devices for the all the devices in block number 1 (above the winning device in that block). *Figure 10-14* shows the timing diagram for the globally winning device (defined as the final winner within its own and all blocks) in block number 1. *Figure 10-15* shows the timing diagram for all the devices below the globally winning device in block number 1. *Figure 10-16*, *Figure 10-17*, and *Figure 10-18* show the timing diagrams of the devices above the globally winning device, the globally winning device, and the devices below the globally winning device, respectively, for block number 2. *Figure 10-20*, *Figure 10-21*, and *Figure 10-22* show the timing diagrams of the devices above globally winning device, the globally winning device, and the devices below the globally winning device except the last device (device 14), respectively, for block number 3.

The 72-bit Search operation is pipelined and executes as follows. Four cycles from the Search command, each of the devices knows the outcome internal to it for that operation. In the fifth cycle after the Search command, the devices in a block arbitrate for a winner among them (a "block" being defined as less than or equal to four devices resolving the winner within them using the LHI[6:0] and LHO[1:0] signalling mechanism). In the sixth cycle after the Search command, the blocks (of devices) resolve the winning block through the BHI[2:0] and BHO[2:0] signalling mechanism. The winning device within the winning block is the global winning device for a Search operation.

Table 10-13. Hit/Miss Assumptions

Search Number	1	2	3	4
Block 0	Miss	Miss	Miss	Miss
Block 1	Miss	Miss	Hit	Miss
Block 2	Miss	Hit	Hit	Miss
Block 3	Hit	Hit	Miss	Miss

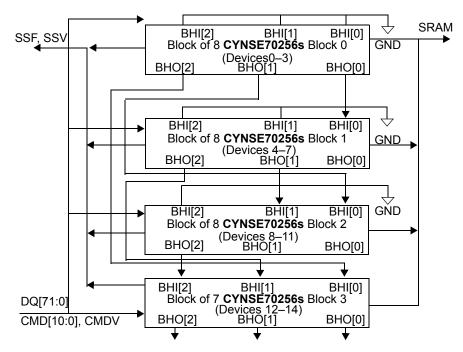


Figure 10-10. Hardware Diagram for a Table with Fifteen Devices



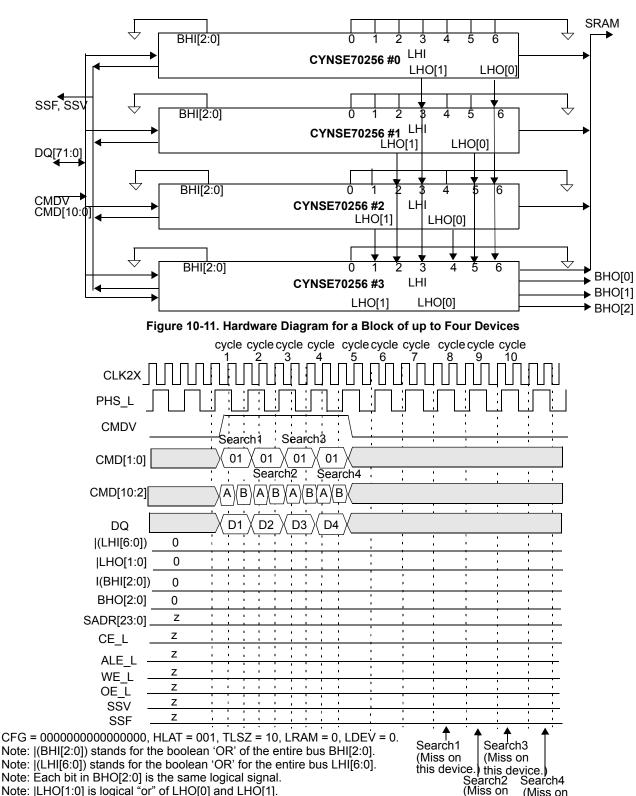


Figure 10-12. Timing Diagram for Each Device in Block Number 0 (Miss on Each Device)

Note: |LHO[1:0] is logical "or" of LHO[0] and LHO[1].

(Miss on this device. his device.)



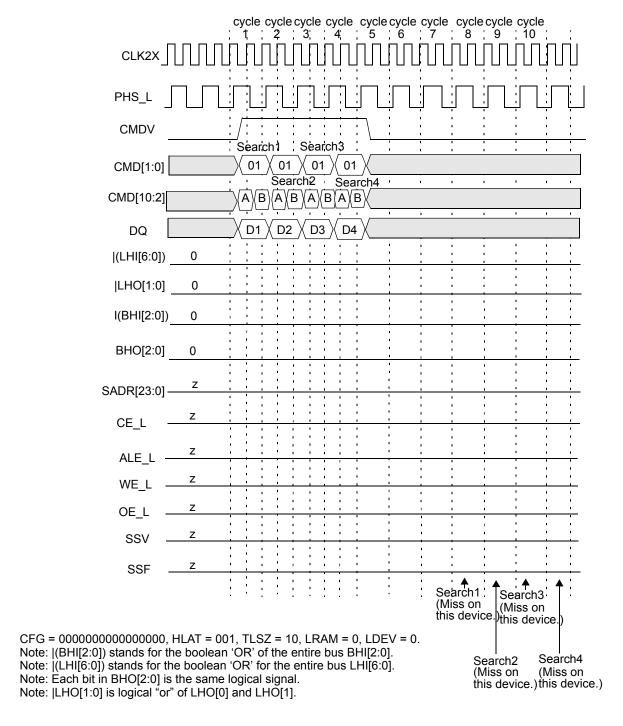


Figure 10-13. Timing Diagram for Each Device Above the Winning Device in Block Number 1



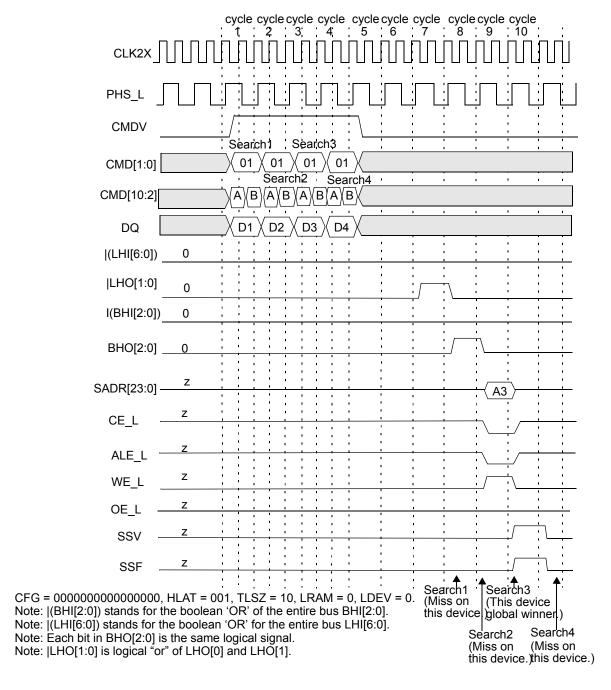


Figure 10-14. Timing Diagram for Globally Winning Device in Block Number 1



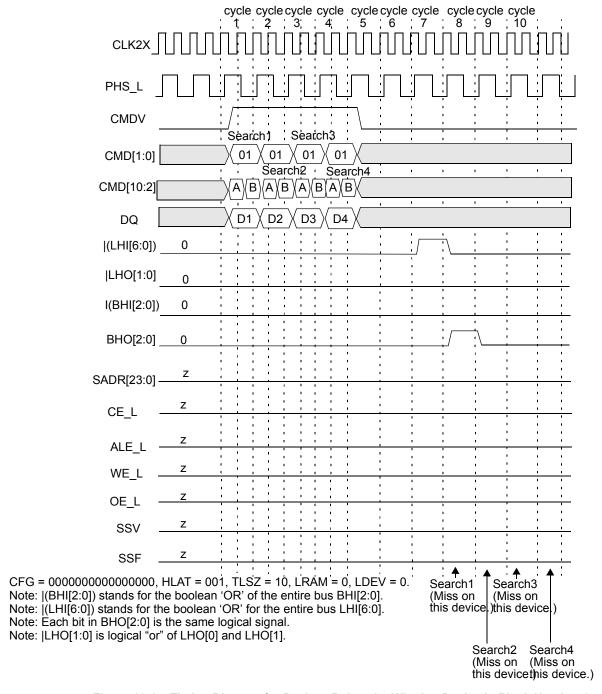


Figure 10-15. Timing Diagram for Devices Below the Winning Device in Block Number 1



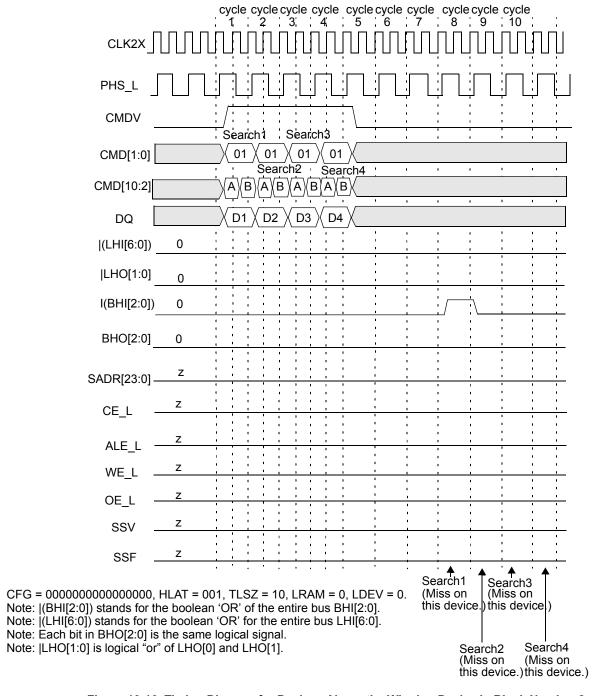


Figure 10-16. Timing Diagram for Devices Above the Winning Device in Block Number 2



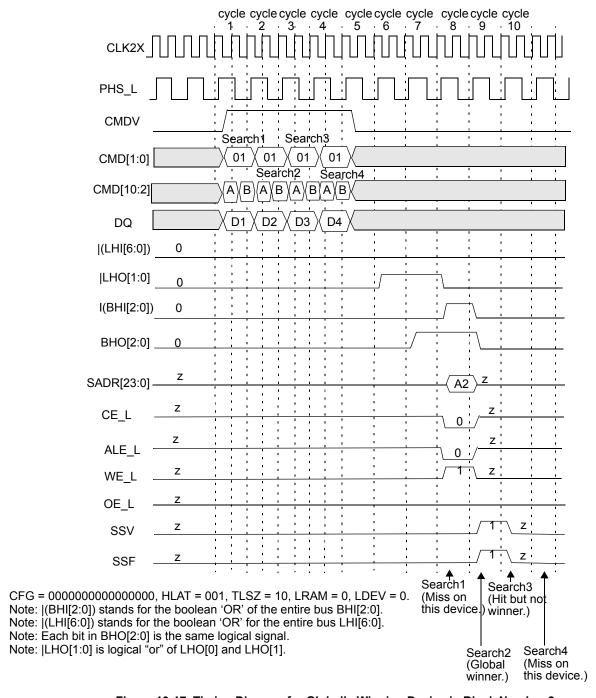


Figure 10-17. Timing Diagram for Globally Winning Device in Block Number 2



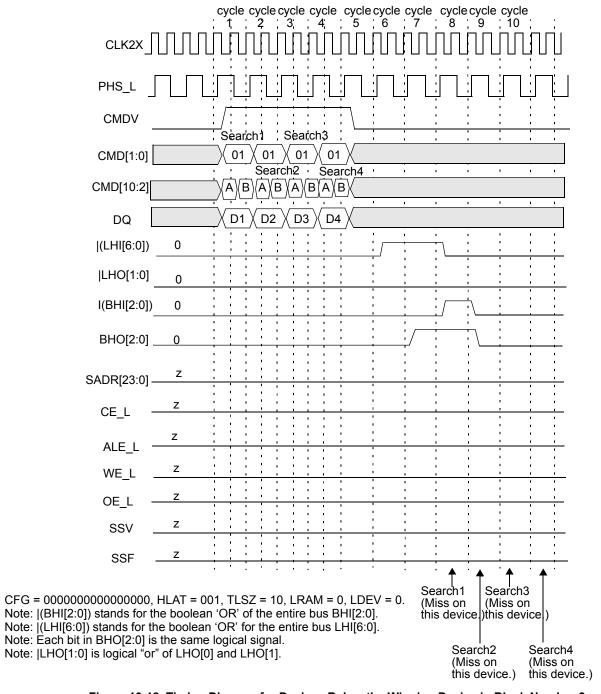


Figure 10-18. Timing Diagram for Devices Below the Winning Device in Block Number 2



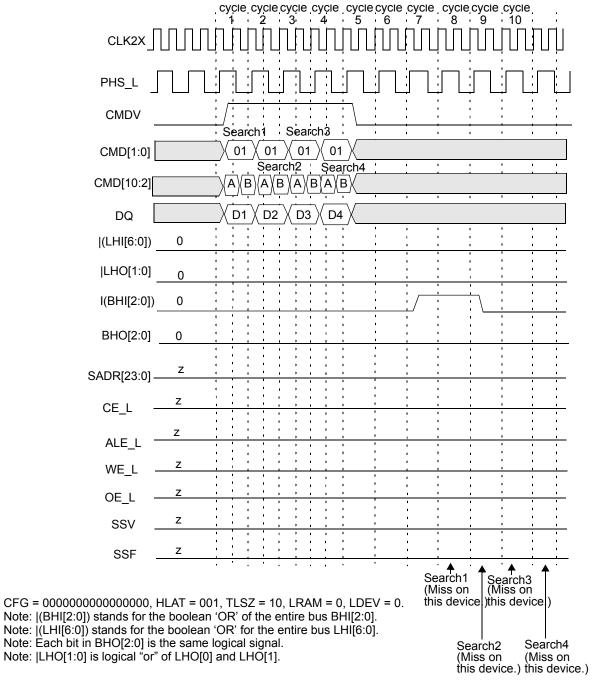


Figure 10-19. Timing Diagram for Devices Above the Winning Device in Block Number 3



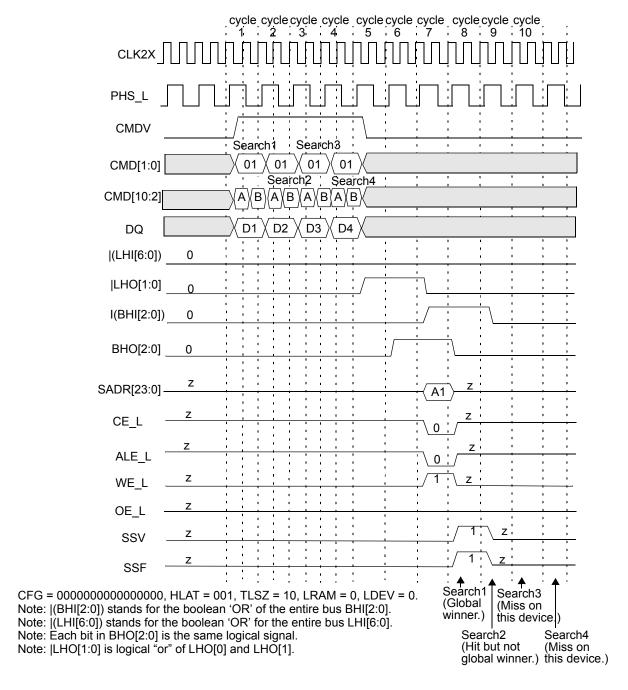


Figure 10-20. Timing Diagram for Globally Winning Device in Block Number 3



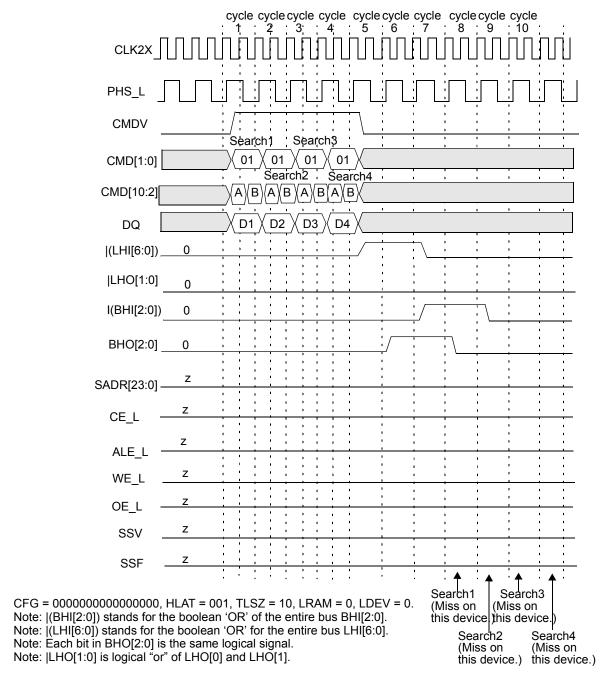


Figure 10-21. Timing Diagram for Devices Below the Winning Device in Block Number 3 (Except the Last Device [Device 14])



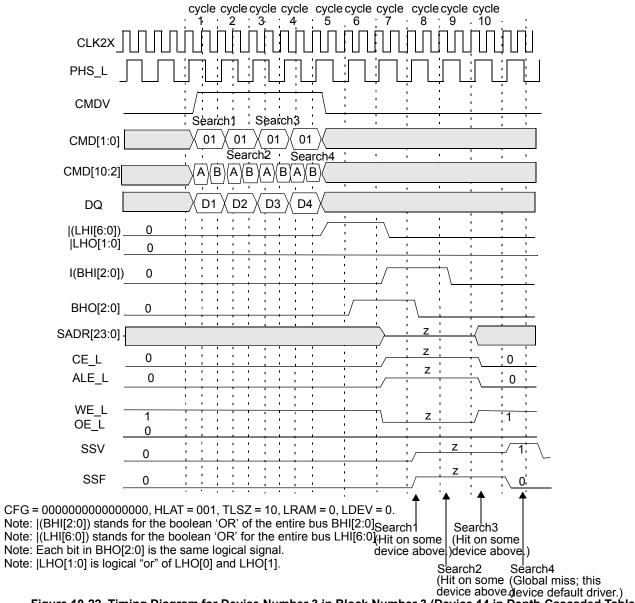


Figure 10-22. Timing Diagram for Device Number 3 in Block Number 3 (Device 14 in Depth-Cascaded Table)

The following is the sequence of operation for a single 72-bit Search command (also refer to Subsection 10.2, "Commands and Command Parameters," on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. {CMD[10], CMD[5:3]} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the same bits that will be driven on SADR[23:21] by this device if it has a hit. DQ[71:0] must be driven with the 72-bit data to be compared. The CMD[2] signal must be driven to a logic 0.
- Cycle B: The host ASIC continues to drive CMDV HIGH and applies Search command (10) on CMD[1:0]. CMD[5:2] must be driven by the index of the comparand register pair for storing the 144-bit word presented on the DQ bus during cycles A and B. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and the hit flag (see page 14 for the description of SSR[0:7]). The DQ[71:0] continues to carry the 72-bit data to be compared. Note. For 72-bit searches, the host ASIC must supply the same 72-bit data on DQ[71:0] during both cycles A and B. The even and odd pair of GMRs selected for the compare must be programmed with the same value.

The logical 72-bit Search operation is shown in Figure 10-23. The entire table (fifteen devices of 72-bit entries) is compared to a 72-bit word K (presented on the DQ bus in both cycles A and B of the command) using the GMR and local mask bits. The effective GMR is the 72-bit word specified by the identical value, in both even and odd GMR pairs in each of banks in each of the fifteen devices, and selected by the GMR index in the command's cycle A. The 72-bit word K (presented on the DQ bus in both



cycles A and B of the command) is also stored in both even and odd comparand register pairs, in each of the banks of all fifteen devices, and selected by the comparand register index in command cycle B. In the ×72 configuration, the even comparand register can be subsequently used by the Learn command only in the first non-full device. The word K (presented on the DQ bus in both cycles A and B of the command) is compared with each entry in the table, starting at location 0. The first matching entry's location address L is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (see "SRAM Addressing" on page 86). The global winning device will drive the bus in a specific cycle. On global miss cycles, the device with LRAM = 1 and LDEV = 1 will be the default driver for such missed cycles.

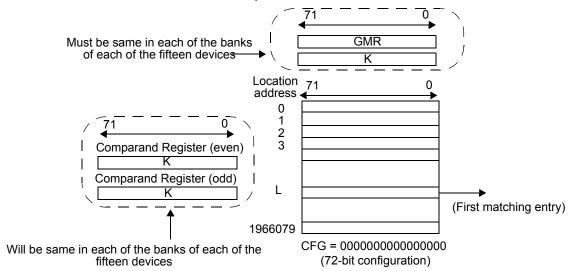


Figure 10-23. ×72 Table with Fifteen Devices

The Search command is a pipelined operation and executes a Search at half the rate of the frequency of CLK2X for 72-bit searches in ×72-configured tables. The latency of SADR, CE\_L, ALE\_L, WE\_L, SSV, and SSF from the 72-bit Search command cycle (two CLK2X cycles) is shown in *Table 10-14*.

Table 10-14. Search Latency from Instruction to SRAM Access Cycle

Number of Devices	Max Table Size	Latency in CLK Cycles
1–4 (TLSZ = 01)	512K × 72 bits	5
1–15 (TLSZ = 10)	1920 K × 72 bits	6

For up to fifteen devices in the table (with TLSZ = 10), the latency of the Search from command to SRAM access cycle is 6. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in *Table 10-15*.

Table 10-15. Shift of SSF and SSV from SADR

HLAT	Number of CLK Cycles
000	0
001	1
010	2
011	3
100	4
101	5
110	6
111	7

10.6.3 144-bit Search on Tables Configured as ×144 using up to Four CYNSE70256 Devices

The hardware diagram of the Search subsystem of four devices is shown in *Figure 10-24*. The following are parameters that are programmed into the four devices.

- First three devices (devices 0-2, banks 0 and 1): CFG = 0101010101010101, TLSZ = 01, HLAT = 010, LRAM = 0, and LDEV = 0.
- First three devices (devices 3, Bank 0): CFG = 010101010101010101, TLSZ = 01, HLAT = 010, LRAM = 0, and LDEV = 0.
- Fourth device (device 3, Bank 1): CFG = 010101010101010101, TLSZ = 01, HLAT = 010, LRAM = 1, and LDEV = 1.



**Note.** All four devices must be programmed with the same value of TLSZ and HLAT. Only the last bank of the last device in the table must be programmed with LRAM = 1 and LDEV = 1 (device number 3 in this case).

Figure 10-25 shows the timing diagram for a Search command in the 144-bit-configured table of four devices for device number 0. Figure 10-26 shows the timing diagram for a Search command in the 144-bit-configured table consisting of four devices for device number 1. Figure 10-27 shows the timing diagram for a Search command in the 144-bit configured table consisting of four devices for device number 3 (the last device in this specific table). For these timing diagrams, the four 144-bit searches are performed sequentially, and the following Hit/Miss assumptions are made (see Table 10-16).

Table 10-16. Hit/Miss Assumptions

Search Number	1	2	3	4
Device 0	Hit	Miss	Hit	Miss
Device 1	Miss	Hit	Hit	Miss
Devices 2–6	Miss	Miss	Miss	Miss
Device 7	Miss	Miss	Hit	Hit

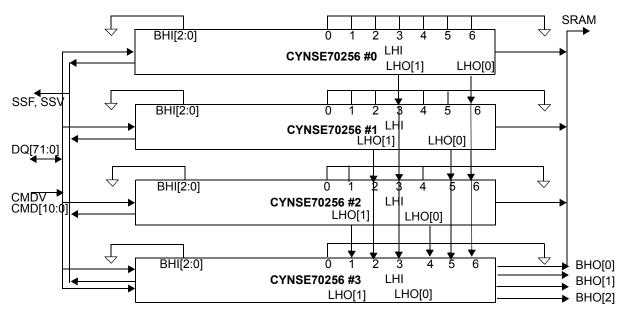


Figure 10-24. Hardware Diagram for a Table with Four Devices



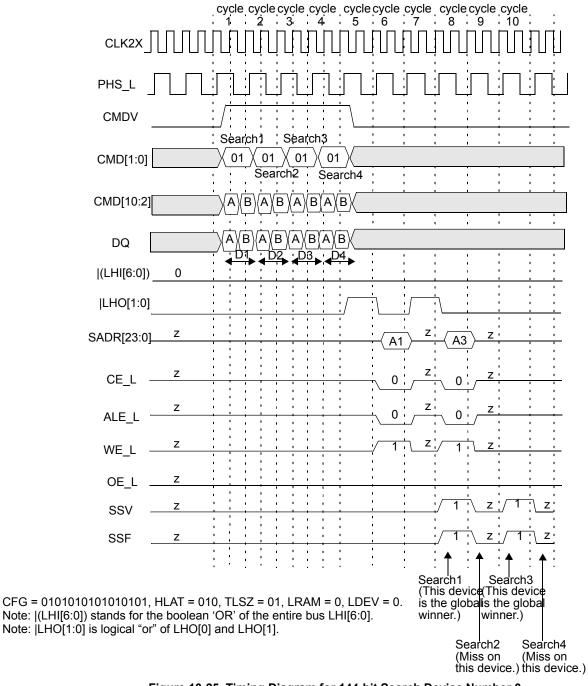


Figure 10-25. Timing Diagram for 144-bit Search Device Number 0



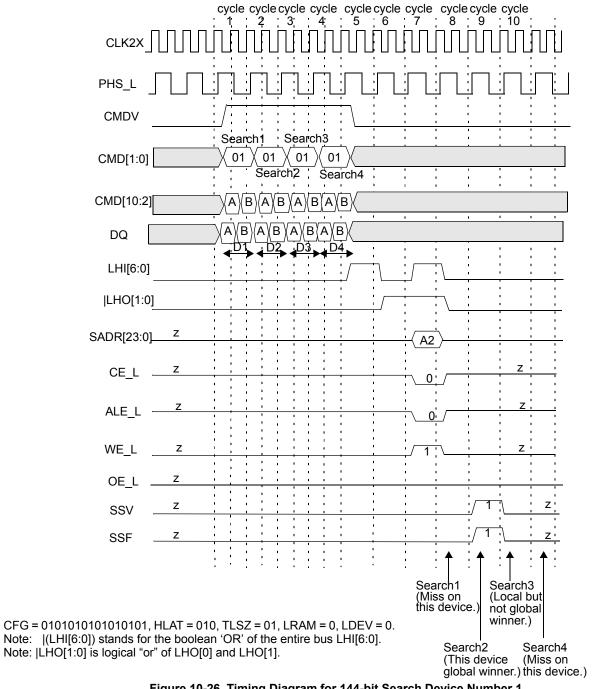


Figure 10-26. Timing Diagram for 144-bit Search Device Number 1



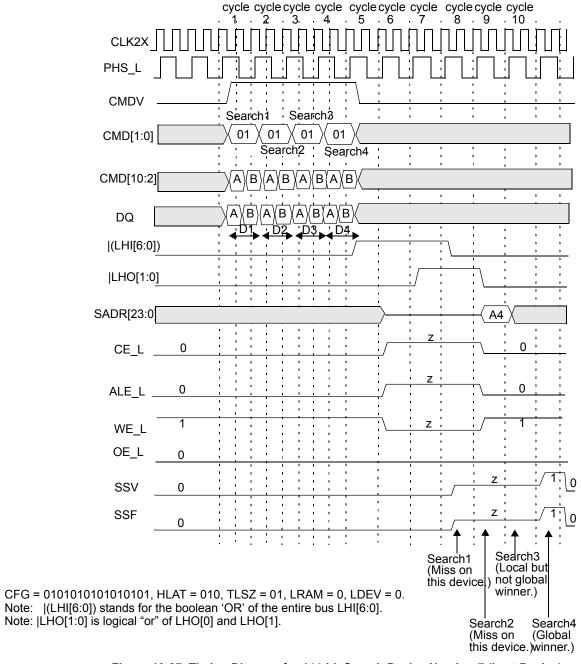


Figure 10-27. Timing Diagram for 144-bit Search Device Number 7 (Last Device)

The following is the sequence of operation for a single 144-bit Search command (also see Subsection 10.2, "Commands and Command Parameters," on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. {CMD[10], CMD[5:3]} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the same bits that will be driven by this device on SADR[23:21] if it has a hit. DQ[71:0] must be driven with the 72-bit data ([143:72]) in order to be compared against all even locations. The CMD[2] signal must be driven to logic 0.
- Cycle B: The host ASIC continues to drive CMDV HIGH and to apply Search command code (10) on CMD[1:0]. CMD[5:2] must be driven by the index of the comparand register pair for storing the 144-bit word presented on the DQ bus during cycles A and B. CMD[8:6] signals must be driven with the SSR index that will be used for storing the address of the matching entry and the hit flag (see page 14 for the description of SSR[0:7]). The DQ[71:0] is driven with 72-bit data ([71:0]) compared against all odd locations.



The logical 144-bit Search operation is shown in *Figure 10-28*. The entire table (four devices of 144-bit entries) is compared to a 144-bit word K (presented on the DQ bus in cycles A and B of the command) using the GMR and local mask bits. The GMR is the 144-bit word specified by the even and odd global mask pair selected by the GMR index in the command's cycle A. The 144-bit word K (presented on the DQ bus in cycles A and B of the command) is also stored in the even and odd comparand registers specified by the comparand register index in command cycle B. In ×144 configurations, the even and odd comparand registers can be subsequently used by the Learn command in only one of the devices (the first non-full device). The word K (presented on the DQ bus in cycles A and B of the command) is compared to each entry in the table starting at location 0. The first matching entry's location, address L, is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (see "SRAM Addressing" on page 86). The global winning device will drive the bus in a specific cycle. On global miss cycles the device with LRAM = 1 (the default driving device for the SRAM bus) and LDEV = 1 (the default driving device for SSF and SSV signals) will be the default driver for such missed cycles. [16]

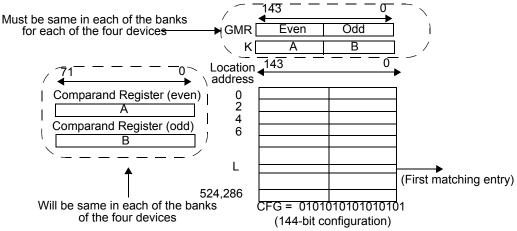


Figure 10-28. ×144 Table with Four Devices

The Search command is a pipelined operation and executes a Search at half the rate of the frequency of CLK2X for 144-bit searches in ×144-configured tables. The latency of SADR, CE\_L, ALE\_L, WE\_L, SSV, and SSF from the 144-bit Search command cycle (two CLK2X cycles) is shown in *Table 10-17*.

Table 10-17. Search Latency from Instruction to SRAM Access Cycle

Number of Devices	Max Table Size	Latency in CLK Cycles
1–4 (TLSZ = 01)	256K × 144 bits	5
1–15 (TLSZ = 10)	960K × 144 bits	6

For one to four devices in the table and TLSZ = 01, Search latency from command to SRAM access cycle is 5. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in *Table 10-18*.

Table 10-18. Shift of SSF and SSV from SADR

HLAT	Number of CLK Cycles
000	0
001	1
010	2
011	3
100	4
101	5
110	6
111	7

#### Note:

16. During 144-bit searches of 144-bit-configured tables, the Search hit will always be at an even address.



10.6.4 144-bit Search on Tables Configured as ×144 using up to Fifteen CYNSE70256 Devices

The hardware diagram of the Search subsystem of fifteen devices is shown in *Figure 10-29*. Each of the four blocks in the diagram represents a block of four CYNSE70256 devices (except the last, which has three devices). The diagram for a block of four devices is shown in *Figure 10-30*. The following are the parameters programmed into the fifteen devices.

- First fourteen devices (devices 0–13, Bank 0 and 1): CFG = 0101010101010101, TLSZ = 10, HLAT = 001, LRAM = 0, and LDEV = 0.
- Fifteenth device (device 14, Bank 0): CFG = 010101010101010101, TLSZ = 10, HLAT = 001, LRAM = 0, and LDEV = 0.
- Fifteenth device (device 14, Bank 1): CFG = 0101010101010101, TLSZ = 10, HLAT = 001, LRAM = 1, and LDEV = 1. [17]

The timing diagrams referred to in this paragraph reference the Hit/Miss assumptions defined in *Table 10-19*. For the purpose of illustrating the timings, it is further assumed that the there is only one device with a matching entry in each of the blocks. *Figure 10-31* shows the timing diagram for a Search command in the 144-bit-configured table (fifteen devices) for each of the four devices in block number 0. *Figure 10-32* shows the timing diagram for a Search command in the 144-bit-configured table (fifteen devices) for all the devices above the winning device in block number 1. *Figure 10-33* shows the timing diagram for the globally winning device (the final winner within its own and all blocks) in block number 1. *Figure 10-34* shows the timing diagram for all the devices below the globally winning device in block number 1. *Figure 10-35*, *Figure 10-36*, and *Figure 10-37* show the timing diagrams of the devices above globally winning device, the globally winning device, and the devices below the globally winning device, respectively, for block number 2. *Figure 10-38*, *Figure 10-39*, *Figure 10-40*, and *Figure 10-41* show, respectively, the timing diagrams of the devices above the globally winning device, the globally winning device, and devices below the globally winning device except the last device (device 14), and then the last device (device 14) for block number 3.

The 144-bit Search operation is pipelined and executes as follows. In the fifth cycle after the Search command, the devices in a block (being less than or equal to four devices resolving the winner within them using the LHI[6:0] and LHO[1:0] signalling mechanism) arbitrate for a winner among them. In the sixth cycle after the Search command, the blocks (of devices) resolve the winning block through the BHI[2:0] and BHO[2:0] signalling mechanism. The winning device in the winning block is the global winning device for a Search operation.

Table 10-19. Hit/Miss Assumptions

Search Number	1	2	3	4
Block 0	Miss	Miss	Miss	Miss
Block 1	Miss	Miss	Hit	Miss
Block 2	Miss	Hit	Hit	Miss
Block 3	Hit	Hit	Miss	Miss

#### Note:

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<sup>17.</sup> All fifteen devices must be programmed with the same value of TLSZ and HLAT. Only the last bank of the last device in the table must be programmed with LRAM = 1 and LDEV = 1 (device number 30 in this case).



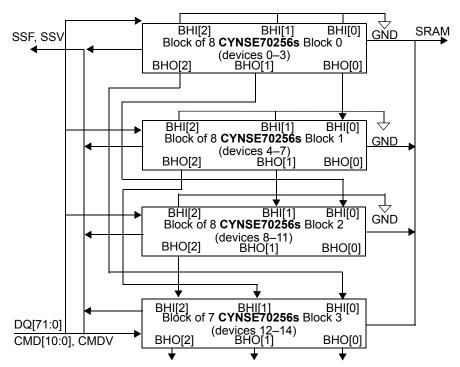


Figure 10-29. Hardware Diagram for a Table with Fifteen Devices

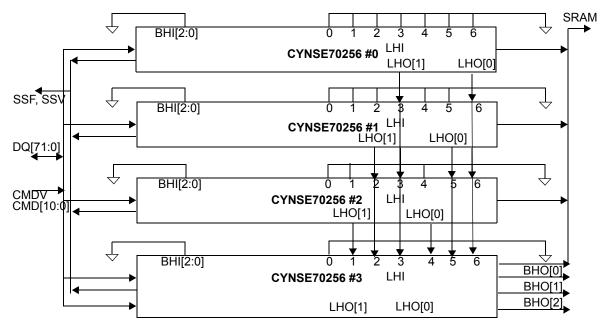


Figure 10-30. Hardware Diagram for a Table with Four Devices



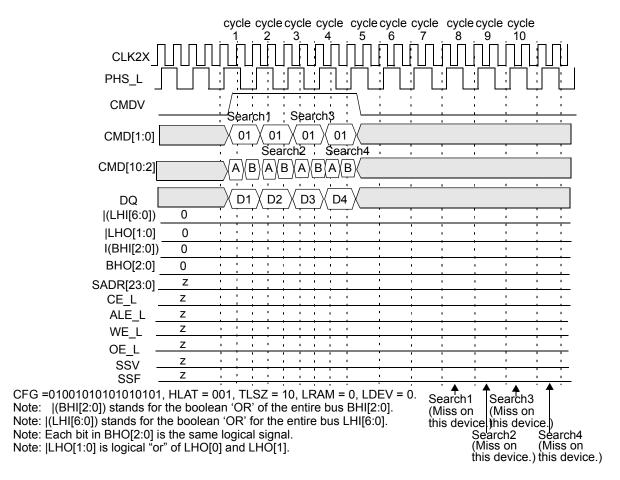


Figure 10-31. Timing Diagram for Each Device in Block Number 0 (Miss on Each Device)



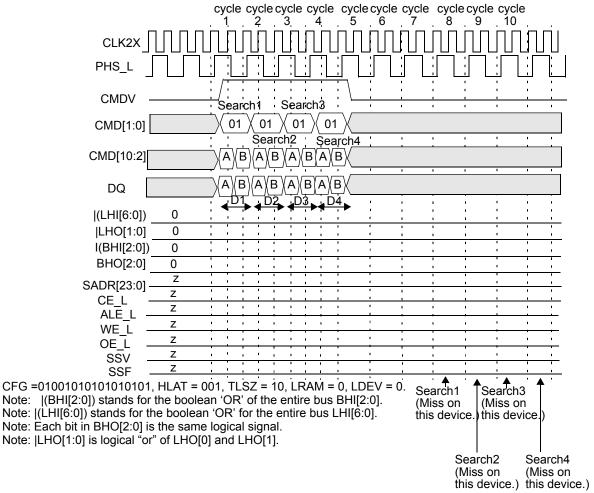


Figure 10-32. Timing Diagram for Each Device Above the Winning Device in Block Number 1



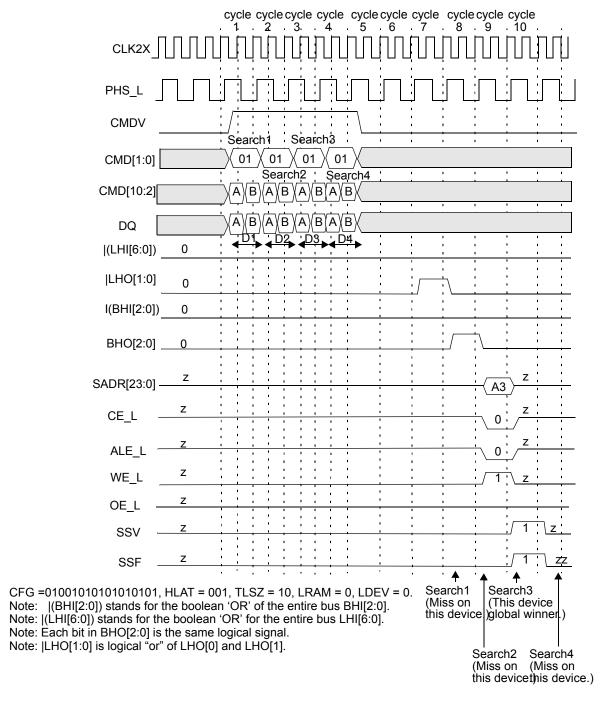


Figure 10-33. Timing Diagram for Globally Winning Device in Block Number 1



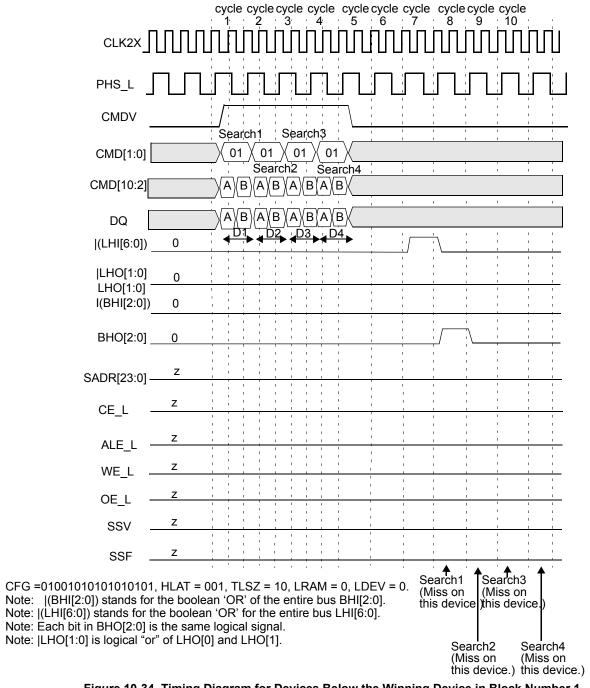


Figure 10-34. Timing Diagram for Devices Below the Winning Device in Block Number 1



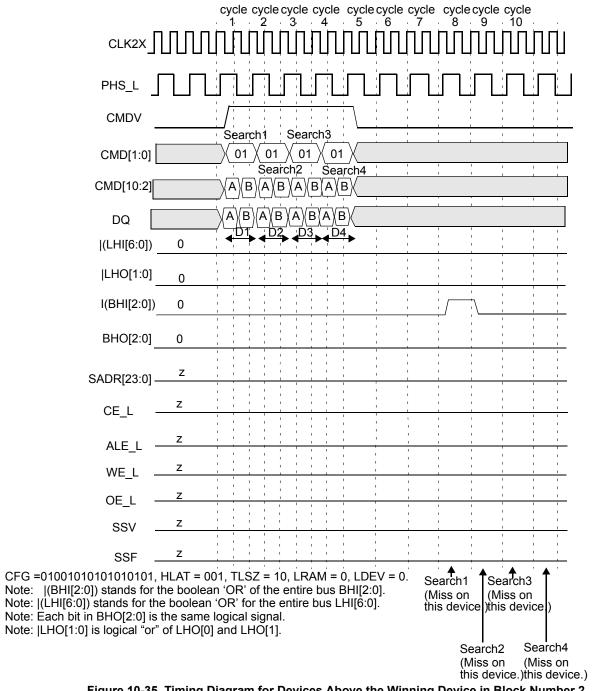


Figure 10-35. Timing Diagram for Devices Above the Winning Device in Block Number 2



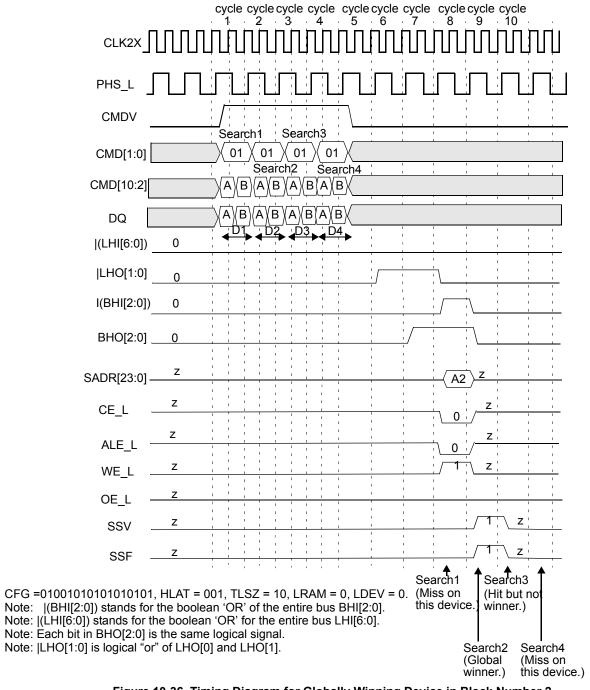


Figure 10-36. Timing Diagram for Globally Winning Device in Block Number 2



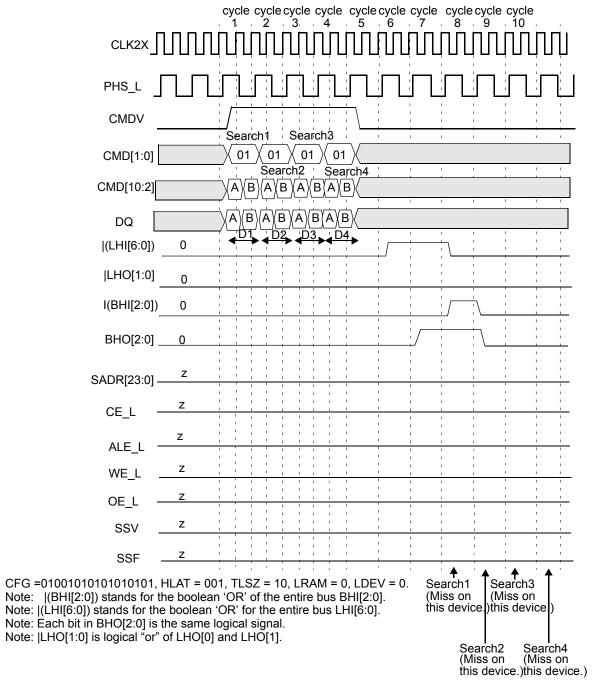


Figure 10-37. Timing Diagram for Devices Below the Winning Device in Block Number 2



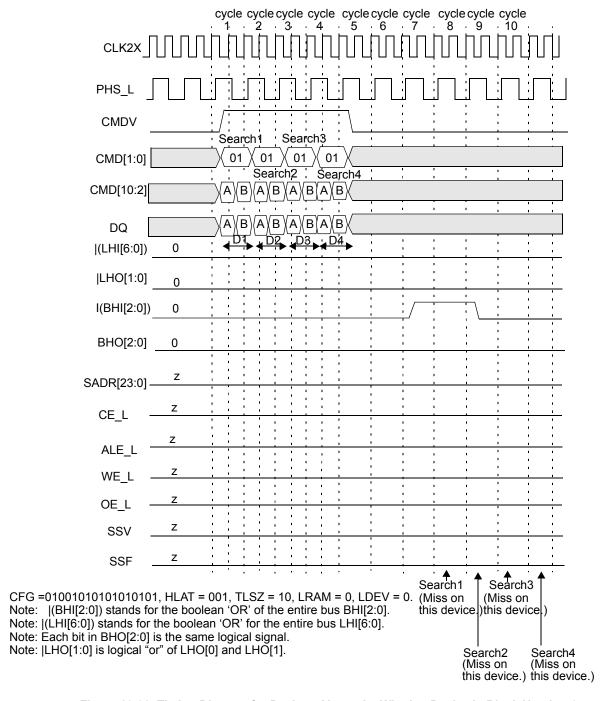


Figure 10-38. Timing Diagram for Devices Above the Winning Device in Block Number 3



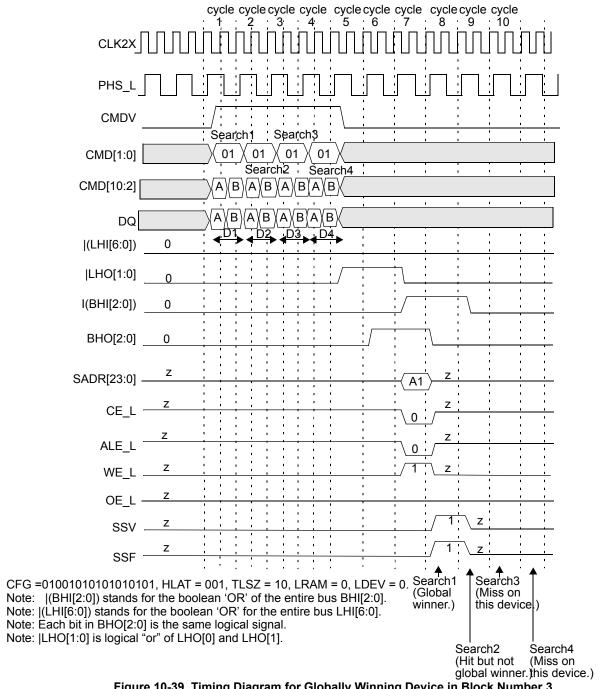


Figure 10-39. Timing Diagram for Globally Winning Device in Block Number 3



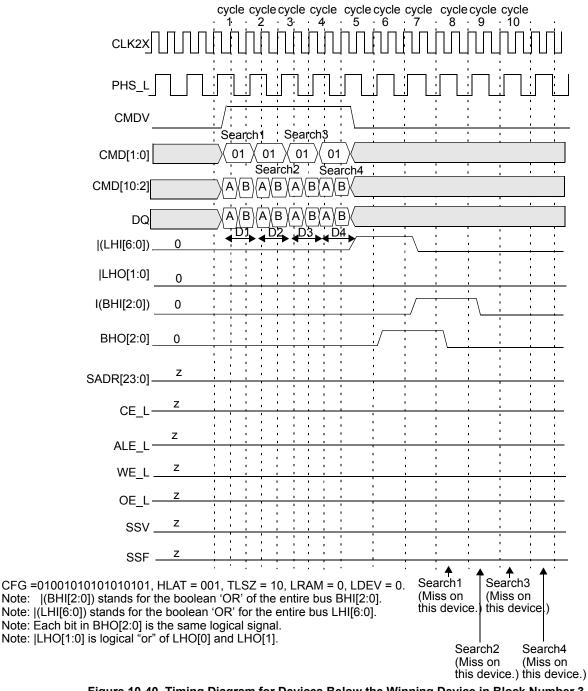


Figure 10-40. Timing Diagram for Devices Below the Winning Device in Block Number 3

Except Device 14 (the Last Device)



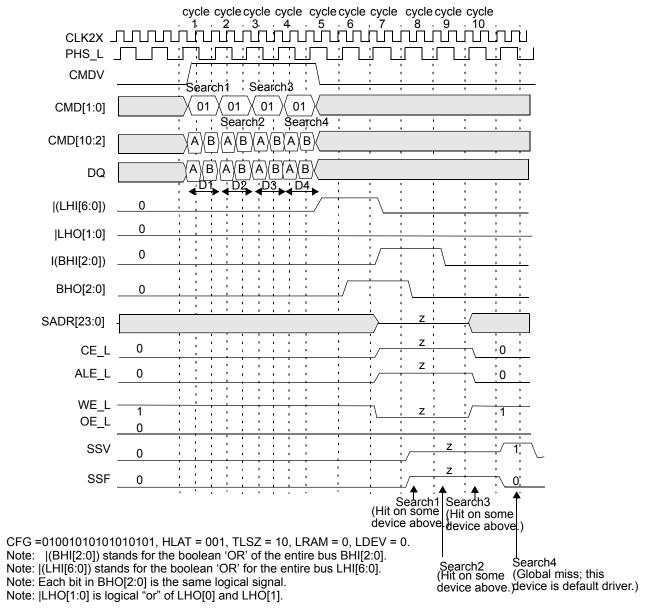


Figure 10-41. Timing Diagram for Device Number 2 in Block Number 3 (Device 14 in a Depth-Cascaded Table)

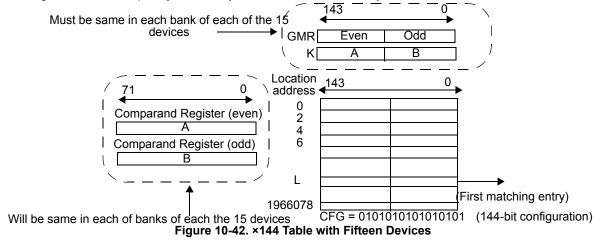
The following is the sequence of operation for a single 144-bit Search command (also refer to "Command and Command Parameters," Section 10.2 on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. {CMD[5:3]} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the bits that will be driven on SADR[23:21] by this device if it has a hit. DQ[71:0] must be driven with the 72-bit data ([143:72]) in order to be compared against all even locations. The CMD[2] signal must be driven to logic 0.
- Cycle B: The host ASIC continues to drive CMDV HIGH and to apply Search command code (10) on CMD[1:0]. CMD[5:2] must be driven by the index of the comparand register pair for storing the 144-bit word presented on the DQ bus during cycles A and B. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and the hit flag (see page 14 for the description of SSR[0:7]). The DQ[71:0] is driven with 72-bit data ([71:0]) to be compared against all odd locations.

The logical 144-bit Search operation is shown in *Figure 10-42*. The entire table of fifteen devices (consisting of 144-bit entries) is compared against a 144-bit word K that is presented on the DQ bus in cycles A and B of the command using the GMR and local mask bits. The GMR is the 144-bit word specified by the even and odd global mask pair selected by the GMR Index in the command's cycle A.



The 144-bit word K that is presented on the DQ bus in cycles A and B of the command is also stored in the even and odd comparand registers specified by the comparand register index in command cycle B. In ×144 configurations, the even and odd comparand registers can subsequently be used by the Learn command in only the first non-full device. [18, 19]



The Search command is a pipelined operation. It executes a Search at half the rate of the frequency of CLK2X for 144-bit searches in ×144-configured tables. The latency of SADR, CE\_L, ALE\_L, WE\_L, SSV, and SSF from the 144-bit Search command cycle (two CLK2X cycles) is shown in Table 10-20.

Table 10-20. Search Latency from Instruction to SRAM Access Cycle

Number of Devices	Max Table Size	Latency in CLK Cycles
1–4 (TLSZ = 01)	256K × 144 bits	5
1–14 (TLSZ = 10)	960K × 144 bits	6

Search latency from command to the SRAM access cycle is 6 for 1–15 devices in the table and TLSZ = 10. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in Table 10-21.

Table 10-21. Shift of SSF and SSV from SADR

HLAT	Number of CLK Cycles	
000	0	
001	1	
010	2	
011	3	
100	4	
101	5	
110	6	
111	7	

### 10.6.5 288-bit Search on ×288-configured Tables using up to Four CYNSE70256 Devices

The hardware diagram of the Search subsystem of four devices is shown in Figure 10-43. The following are the parameters programmed into the four devices.

- First seven devices (devices 0–6, Bank 0 and 1): CFG = 101010101010101010, TLSZ = 01, HLAT = 000, LRAM = 0, and LDEV = 0.
- Fourth device (device 7, Bank 0): CFG = 101010101010101010, TLSZ = 01, HLAT = 000, LRAM = 0, and LDEV = 0.
- Fourth device (device 7, Bank 1): CFG = 10101010101010101010, TLSZ = 01, HLAT = 000, LRAM = 1, and LDEV = 1. [20]

#### Notes:

- The Learn command is supported for only one of the blocks consisting of up to four devices in a depth-cascaded table of more than one block. The word K that is presented on the DQ bus in cycles A and B of the command is compared with each entry in the table, starting at location 0. The first matching entry's location address L is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (see Section 12.0, "SRAM Addressing," on page 86). The global winning device will drive the bus in a specific cycle. On global miss cycles the device with LRAM = 1 (the default driving device for the SRAM bus) and LDEV = 1 (the default driving device for SSF and SSV signals) will be the default driver for such missed cycles. During 144-bit searches of 144-bit-configured tables, the Search hit will always be at an even address. All four devices must be programmed with the same values of TLSZ and HLAT. Only the last bank of the last device in the table must be programmed with LRAM = 1 and LDEV = 1 (device number 3 in this case).



Figure 10-44 shows the timing diagram for a Search command in the 288-bit-configured table of four devices for device number 0. Figure 10-45 shows the timing diagram for a Search command in the 288-bit-configured table of four devices for device number 1. Figure 10-46 shows the timing diagram for a Search command in the 288-bit-configured table of four devices for device number 3 (the last device in this table). In these timing diagrams, three 288-bit searches are performed sequentially. The Hit/Miss assumptions were made as shown in Table 10-22.

Table 10-22. Hit/Miss Assumptions

Search Number	1	2	3
Device 0	Hit	Miss	Miss
Device 1	Miss	Hit	Miss
Devices 2–6	Miss	Miss	Miss
Device 7	Miss	Miss	Miss

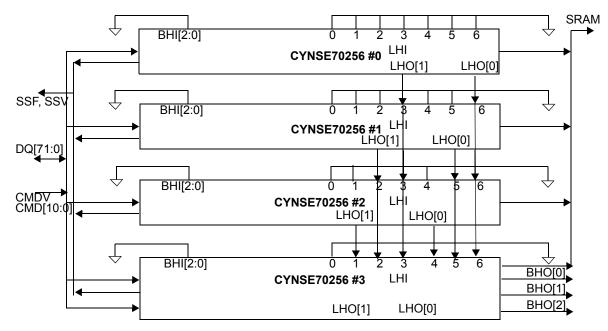


Figure 10-43. Hardware Diagram for a Table with Four Devices



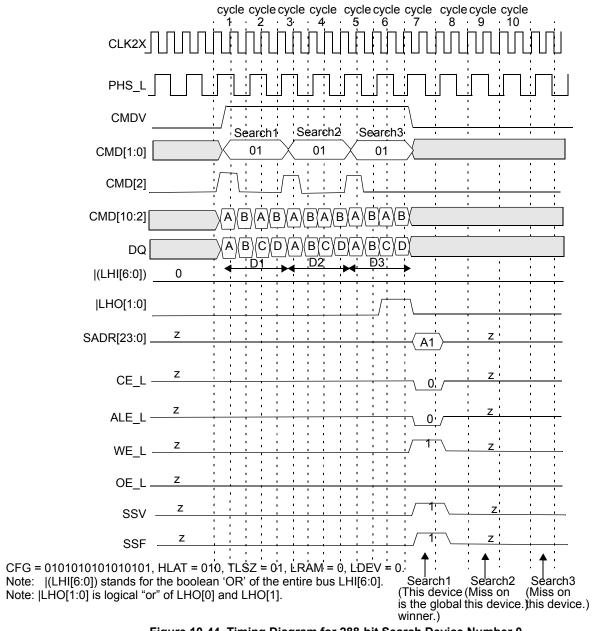


Figure 10-44. Timing Diagram for 288-bit Search Device Number 0



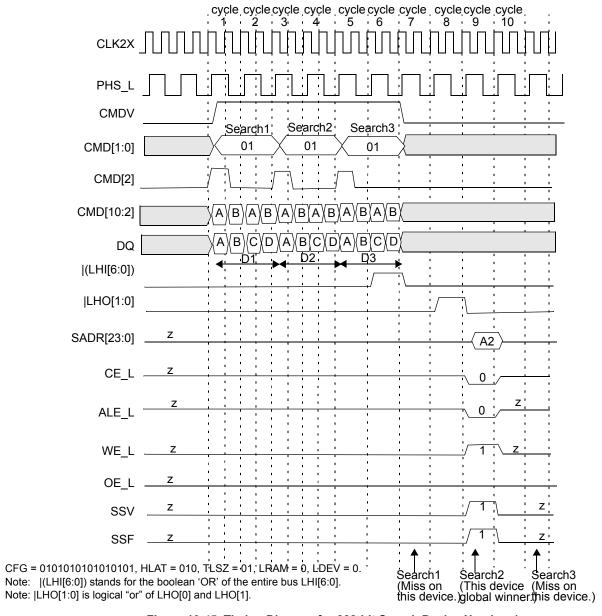


Figure 10-45. Timing Diagram for 288-bit Search Device Number 1



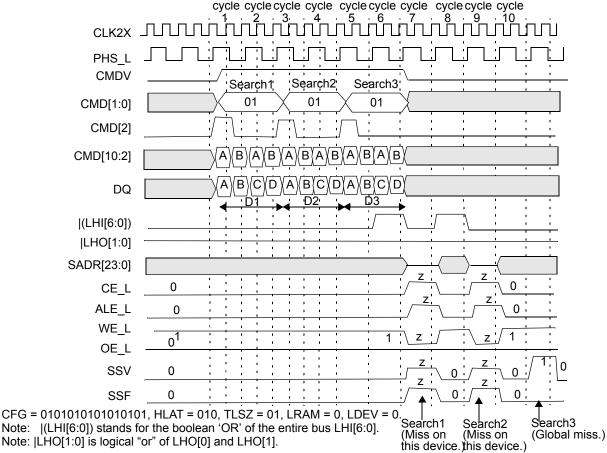


Figure 10-46. Timing Diagram for 288-bit Search Device Number 3 (Last Device)

The following is the sequence of operation for a single 288-bit Search command (see also Subsection 10.2, "Commands and Command Parameters," on page 19).

- Cvcle A: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMDI1:01 signals. (CMDI101.CMDI5:31) signals must be driven with the index to the GMR pair used for bits [287:144] of the data being searched in this operation. DQ[71:0] must be driven with the 72-bit data ([287:216]) to be compared against all locations 0 in the four-word 72-bit page. The CMD[2] signal must be driven to logic 1.[21]
- Cycle B: The host ASIC continues to drive CMDV HIGH and applies Search command code (10) on CMD[1:0]. The DQ[71:0] is driven with the 72-bit data ([215:144]) to be compared against all locations 1 in the four 72-bits-word pages.
- Cycle C: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. {CMD[10], CMD[5:3]} signals must be driven with the index to the GMR pair used for bits [143:0] of the data being searched. CMD[8:6] signals must be driven with the bits that will be driven on SADR[23:21] by this device if it has a hit. DQ[71:0] must be driven with the 72-bit data ([143:72]) to be compared against all locations 2 in the four 72-bits-word page. The CMD[2] signal must be driven to logic 0.
- Cycle D: The host ASIC continues to drive CMDV HIGH and applies Search command code (10) on CMD[1:0]. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and the hit flag (see page 14 for the description of SSR[0:7]). The DQ[71:0] is driven with the 72-bit data ([71:0]) to be compared to all locations 3 in the four 72-bits-word pages. CMD[5:2] is ignored because the Learn instruction is not supported for ×288 tables. [22]

The logical 288-bit Search operation is shown in Figure 10-47. The entire table of 288-bit entries is compared to a 288-bit word K that is presented on the DQ bus in cycles A, B, C, and D of the command using the GMR and the local mask bits. The GMR is the 288-bit word specified by the two pairs of GMRs selected by the GMR indexes in command cycles A and C in each of the four devices. The 288-bit word K that is presented on the DQ bus in cycles A, B, C, and D of the command is compared to each entry in the table starting at location 0. The first matching entry's location address L is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (Section 12.0, "SRAM Addressing," on page 86). [23]

## Notes:

CMD[2] = 1 signals that the Search is a 288-bit search. CMD[8:3] in this cycle is ignored.
For 288-bit searches, the host ASIC must supply four distinct 72-bit data words on DQ[71:0] during cycles A, B, C, and D. The GMR index in cycle A selects a pair of GMRs in each of the banks of each of the four devices that apply to DQ data in cycles A and B. The GMR index in cycle C selects a pair of GMRs in each of the four devices that apply to DQ data in cycles C and D.



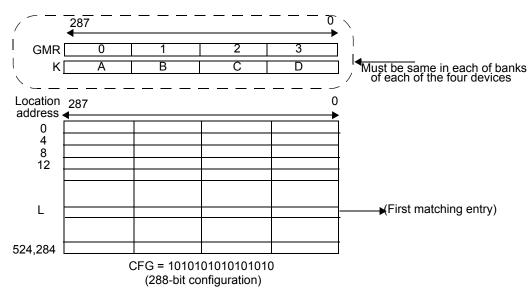


Figure 10-47. ×288 Table with Four Devices

The Search command is a pipelined operation and executes a Search at one-fourth the rate of the frequency of CLK2X for 288bit searches in ×288-configured tables. The latency of SADR, CE L, ALE L, WE L, SSV, and SSF from the 288-bit Search command (measured in CLK cycles) from the CLK2X cycle that contains the C and D cycles is shown in Table 10-23.

Table 10-23. Search Latency from Instruction to SRAM Access Cycle

Number of Devices	Max Table Size	Latency in CLK Dycles
1–4 (TLSZ = 01)	128K × 288 bits	5
1–15 (TLSZ = 10)	480K × 288 bits	6

Search latency from command to SRAM access cycle is 5 for only a single device in the table and TLSZ = 01. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in Table 10-24.

Table 10-24. Shift of SSF and SSV from SADR

HLAT	Number of CLK Cycles
000	0
001	1
010	2
011	3
100	4
101	5
110	6
111	7

# 10.6.6 288-bit Search on Tables Configured as ×288 Using up to Fifteen CYNSE70256 Devices

The hardware diagram of the Search subsystem of fifteen devices is shown in Figure 10-48. Each of the four blocks in the diagram represents a block of four CYNSE70256 devices except the last, which has three devices. The diagram for a block of four devices is shown in Figure 10-49. The following are the parameters programmed into the fifteen devices.

- First thirty devices (devices 0-13, banks 0 and 1): CFG = 10101010101010, TLSZ = 10, HLAT = 000, LRAM = 0, and LDEV = 0.
- Thirty-first device (device 30, Bank 0): CFG = 1010101010101010, TLSZ = 10, HLAT = 000, LRAM = 0, and LDEV = 0.
- Thirty-first device (device 30, Bank 1): CFG = 1010101010101010101010, TLSZ = 10, HLAT = 000, LRAM = 1, and LDEV = 1. [24]

#### Note:

The matching address is always going to be location 0 in a four-entry page for 288-bit Search (two LSBs of the matching index will be 00).

All fifteen devices must be programmed with the same value for TLSZ and HLAT. Only the last bank in the last device in the table must be programmed with LRAM = 1 and LDEV = 1 (device number 14 in this case).

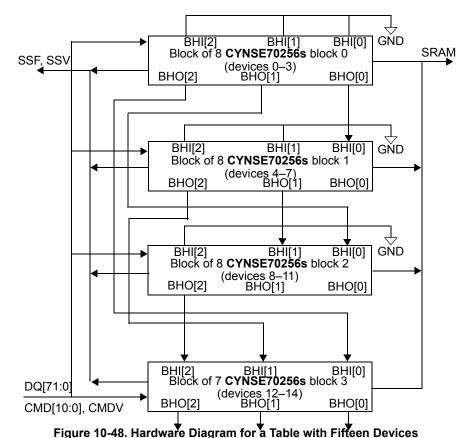


The timing diagrams referred to in this paragraph reference the Hit/Miss assumptions defined in *Table 10-25*. For the purpose of illustrating the timings, it is further assumed that there is only one device with the matching entry in each block. *Figure 10-50* shows the timing diagram for a Search command in the 288-bit-configured table of fifteen devices for each of the four devices in block number 0. *Figure 10-51* shows the timing diagram for a Search command in the 288-bit-configured table of fifteen devices for all devices above the winning device in block number 1. *Figure 10-52* shows the timing diagram for the globally winning device (the final winner within its own and all blocks) in block number 1. *Figure 10-53* shows the timing diagram for the devices below the globally winning device in block number 1. *Figure 10-54*, *Figure 10-55*, and *Figure 10-56*, respectively, show the timing diagrams of the devices above the globally winning device, the globally winning device, and the devices below the globally winning device in block number 2. *Figure 10-57*, *Figure 10-58*, *Figure 10-59*, and *Figure 10-60*, respectively, show the timing diagrams of the devices above the globally winning device, the globally winning device, the devices below the globally winning device (except device 14), and the last device (device 14) in block number 3.

The 288-bit Search operation is pipelined and executes as follows. Four cycles from the last cycle of the Search command, each of the devices knows the outcome internal to it for that operation. In the fifth cycle from the Search command, the devices in a block (less than or equal to four devices resolving the winner among them using LHI[6:0] and LHO[1:0] signalling mechanisms) arbitrate for a winner. In the sixth cycle after the Search command, the blocks of devices resolve the winning block through BHI[2:0] and BHO[2:0] signalling mechanisms. The winning device within the winning block is the global winning device for the Search operation.

Table 10-25. Hit/Miss Assumptions

Search Number	1	2	3
Block 0	Miss	Miss	Miss
Block 1	Miss	Miss	Hit
Block 2	Miss	Hit	Hit
Block 3	Hit	Hit	Miss





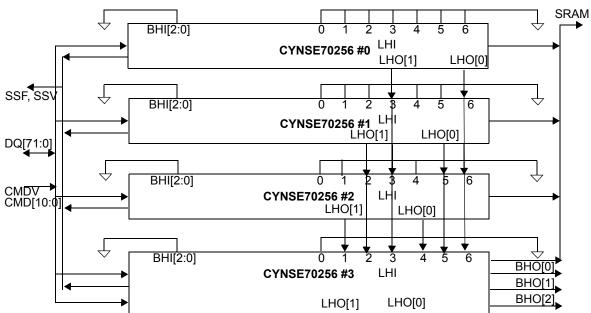


Figure 10-49. Hardware Diagram for a Block of up to Four Devices



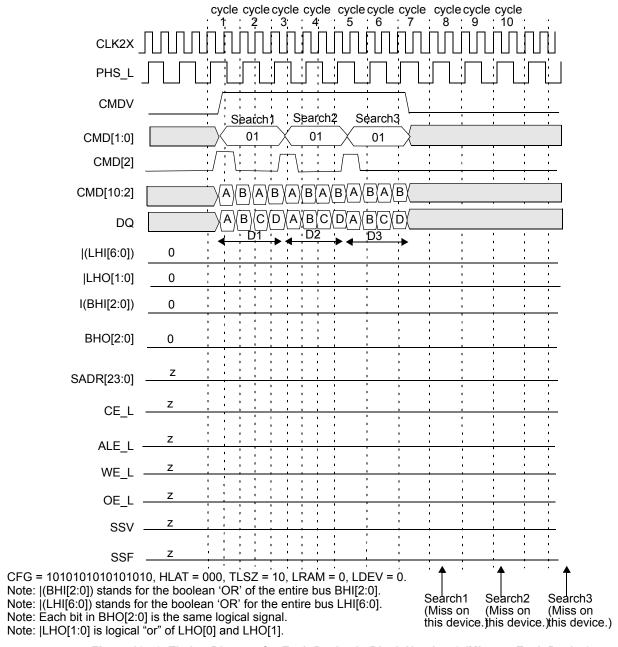


Figure 10-50. Timing Diagram for Each Device in Block Number 0 (Miss on Each Device)



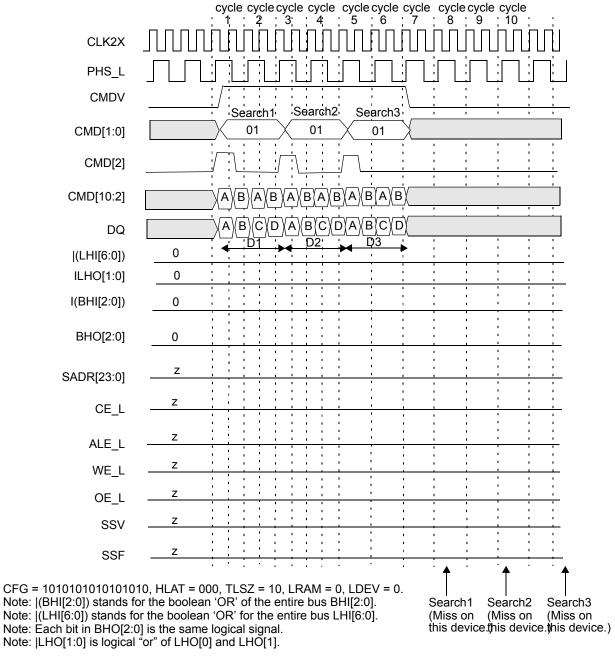


Figure 10-51. Timing Diagram for Each Device Above the Winning Device in Block Number 1



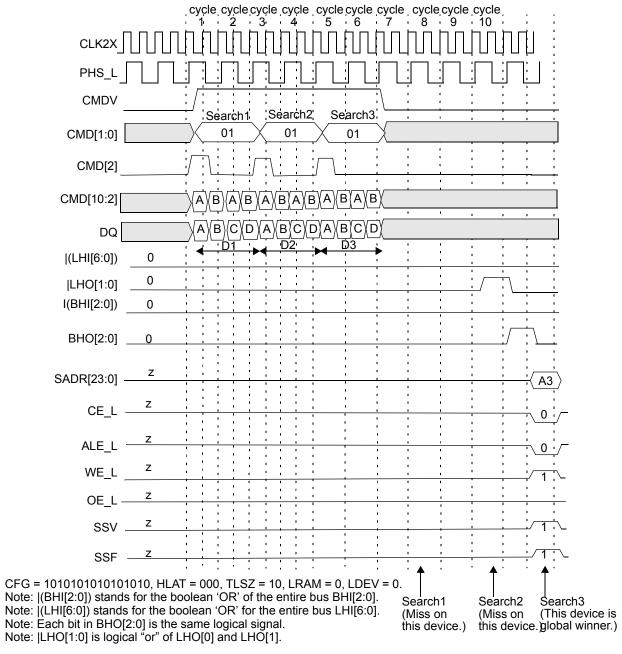


Figure 10-52. Timing Diagram for Globally Winning Device in Block Number 1



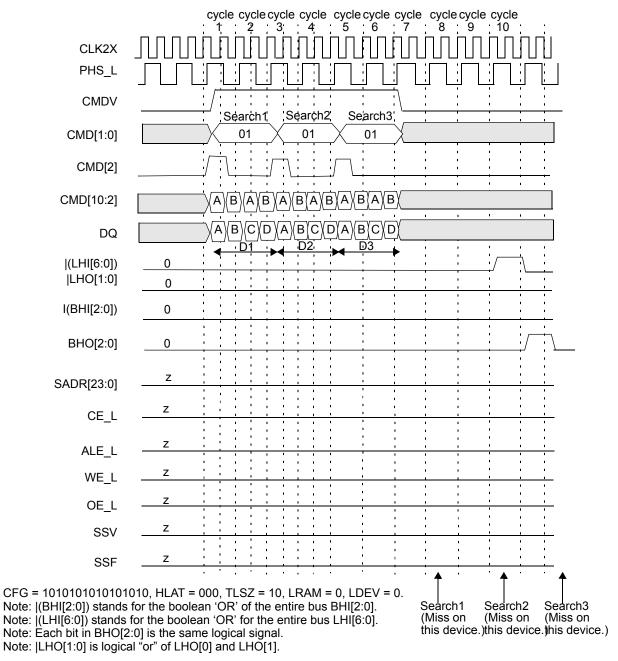


Figure 10-53. Timing Diagram for Devices Below the Winning Device in Block Number 1



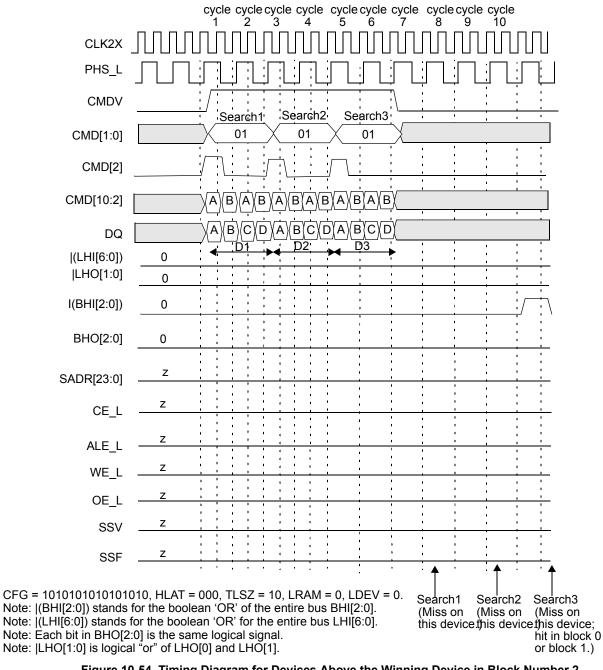


Figure 10-54. Timing Diagram for Devices Above the Winning Device in Block Number 2



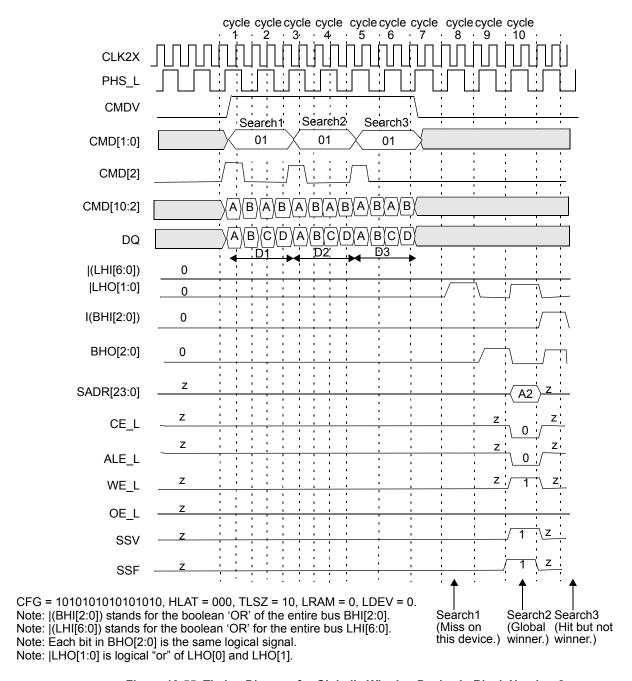


Figure 10-55. Timing Diagram for Globally Winning Device in Block Number 2



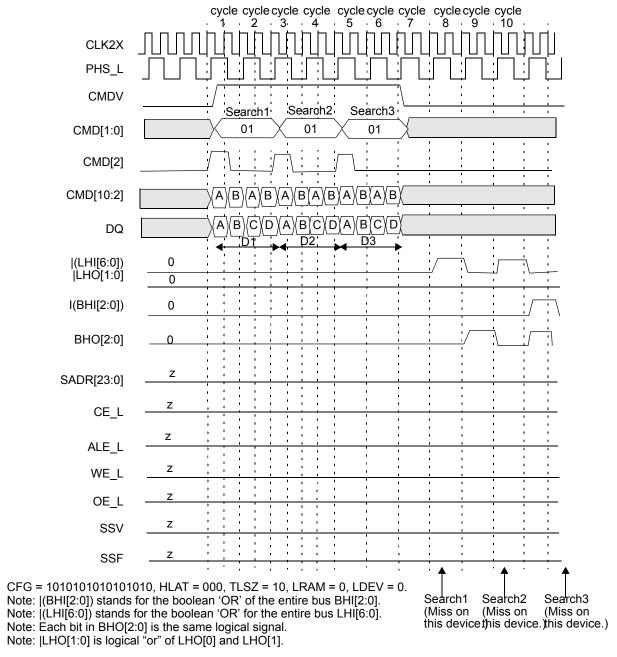


Figure 10-56. Timing Diagram for Devices Below the Winning Device in Block Number 2



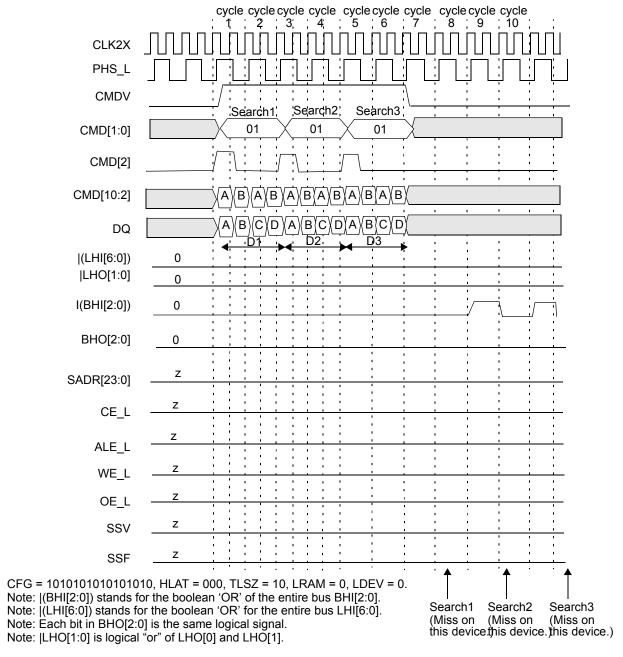


Figure 10-57. Timing Diagram for Devices Above the Winning Device in Block Number 3



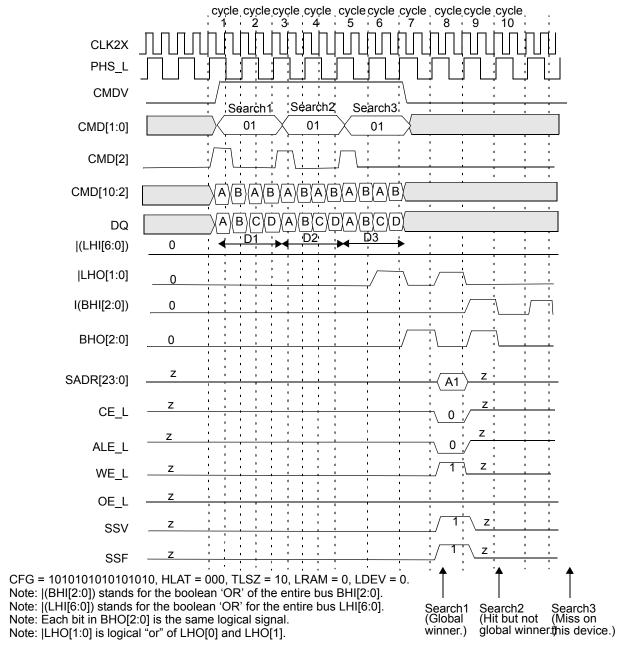


Figure 10-58. Timing Diagram for Globally Winning Device in Block Number 3



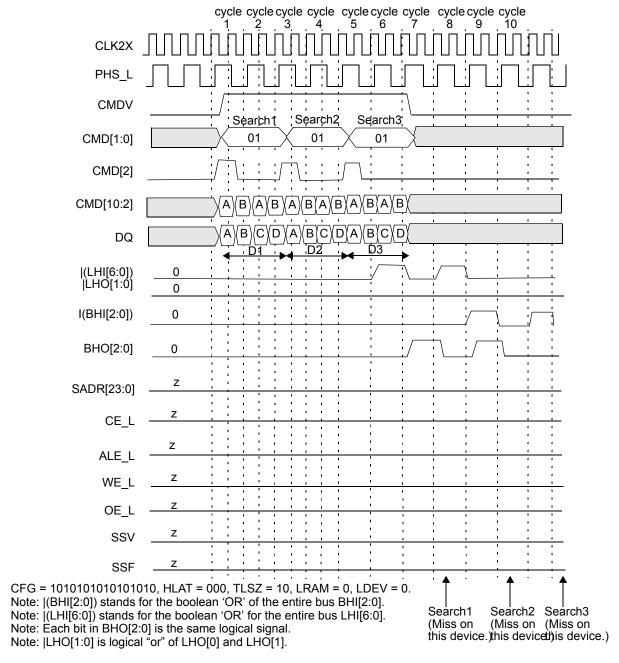


Figure 10-59. Timing Diagram for Devices Below the Winning Device in Block Number 3

Except Device 14 (Last Device)



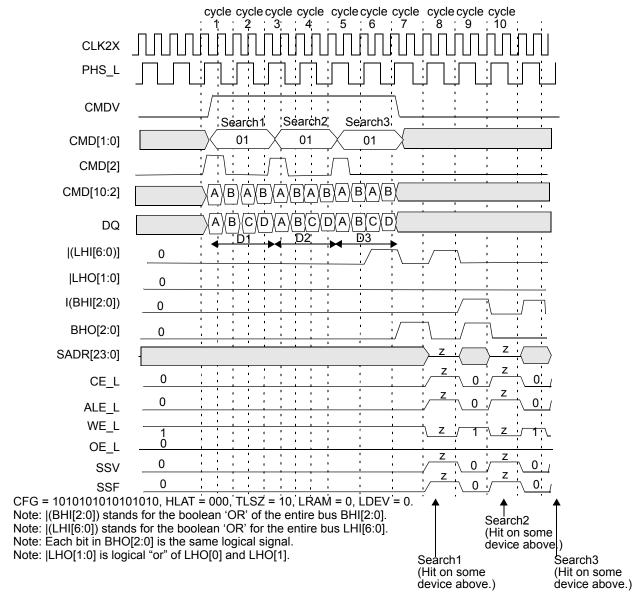


Figure 10-60. Timing Diagram of the Last Device in Block Number 3 (Device 14 in the Table)

The following is the sequence of operation for a single 288-bit Search command (see also Subsection 10.2, "Commands and Command Parameters," on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. {CMD[10],CMD[5:3]} signals must be driven with the index to the GMR pair used for bits [287:144] of the data being searched. DQ[71:0] must be driven with the 72-bit data ([287:216]) to be compared to all locations 0 in the four 72-bits-word page. The CMD[2] signal must be driven to logic 1. [25]
- Cycle B: The host ASIC continues to drive CMDV HIGH and to apply Search command (10) on CMD[1:0]. The DQ[71:0] is driven with the 72-bit data ([215:144]) to be compared to all locations 1 in the four 72-bits-word pages.
- Cycle C: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. {CMD[10],CMD[5:3]} signals must be driven with the index to the GMR pair used for the bits [143:0] of the data being searched. CMD[8:6] signals must be driven with the bits that will be driven by this device on SADR[23:21] if it has a hit. DQ[71:0] must be driven with the 72-bit data ([143:72]) to be compared to all locations 2 in the four 72-bits-word pages. The CMD[2] signal must be driven to logic 0.

### Note:

25. CMD[2] = 1 signals that the Search is a ×288-bit Search. CMD[8:6] is ignored in this cycle.



• Cycle D: The host ASIC continues to drive CMDV HIGH and to apply Search command code (10) on CMD[1:0]. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and the hit flag (see page 14 for a description of SSR[0:7]). The DQ[71:0] is driven with the 72-bit data ([71:0]) to be compared to all locations 3 in the four 72-bits-word pages. CMD[5:2] is ignored because the Learn instruction is not supported for ×288 tables. [26]

The logical 288-bit Search operation is shown in *Figure 10-61*. The entire table of 288-bit entries is compared to a 288-bit word K that is presented on the DQ bus in cycles A, B, C, and D of the command using the GMR and local mask bits. The GMR is the 288-bit word specified by the two pairs of GMRs selected by the GMR Indexes in command cycles A and C in each of the fifteen devices. The 288-bit word K that is presented on the DQ bus in cycles A, B, C, and D of the command is compared to each entry in the table starting at location 0. The first matching entry's location address L is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (see "SRAM Addressing" on page 86).<sup>[27]</sup>

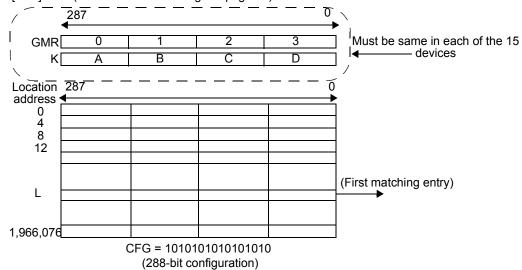


Figure 10-61. ×288 Table with Fifteen Devices

The Search command is a pipelined operation and executes a Search at one-fourth the rate of the frequency of CLK2X for 288-bit searches in ×288-configured tables. The latency of SADR, CE\_L, ALE\_L, WE\_L, SSV, and SSF from the 288-bit Search command (measured in CLK cycles) from the CLK2X cycle that contains the C and D cycles is shown in *Table 10-26*.

Table 10-26. Search Latency from Instruction to SRAM Access Cycle

Number of Devices	Max Table Size	Latency in CLK Cycles
1–8 (TLSZ = 01)	128K × 288 bits	5
1–15 (TLSZ = 10)	480K × 288 bits	6

Search latency from command to SRAM access cycle is 6 for a single device in the table and TLSZ = 10. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in *Table 10-27*.

Table 10-27. Shift of SSF and SSV from SADR

HLAT	Number of CLK Cycles
000	0
001	1
010	2
011	3
100	4
101	5
110	6
111	7

## Notes:

<sup>26.</sup> For 288-bit searches, the host ASIC must supply four distinct 72-bit data words on DQ[71:0] during cycles A, B, C, and D. The GMR index in cycle A selects a pair of GMRs in each of the fifteen devices that apply to DQ data in cycles A and B. The GMR index in cycle C selects a pair of GMRs in each of the fifteen devices that apply to DQ data in cycles C and D.

<sup>27.</sup> The matching address is always going to be location 0 in a four-entry page for 288-bit search (two LSBs of the matching index will be 00).



## 10.6.7 Mixed-size Searches on Tables Configured with Different Widths Using a CYNSE70256 With CFG\_L LOW

This subsection will cover mixed searches (×72, ×144, and ×288) with tables of different widths (×72, ×144, ×288). The sample operation shown is for a single device with CFG = 1010010100000000 that contains three tables of ×72, ×144, and ×288 widths. The operation can be generalized to a block of 4–15 devices using four blocks; the timing and pipeline operation is the same as described previously for fixed searches on a table of one-width size.

Figure 10-62 shows three sequential searches: first, a 72-bit Search on a ×72-configured table; a 144-bit Search on a ×144-configured table; finally, a 288-bit Search on a ×288-configured table that each results in a hit.<sup>[28]</sup>

Figure 10-63 shows the sample table. Two bits in each 72-bit entry will need to designated as the table number bits. One choice can be the 00 values for the table configured as ×72, 01 values for tables configured as ×144, and 10 values for tables configured as ×288. For the above explanation, it is further assumed that bits [71:70] for each entry will be designed as such table designation bits.

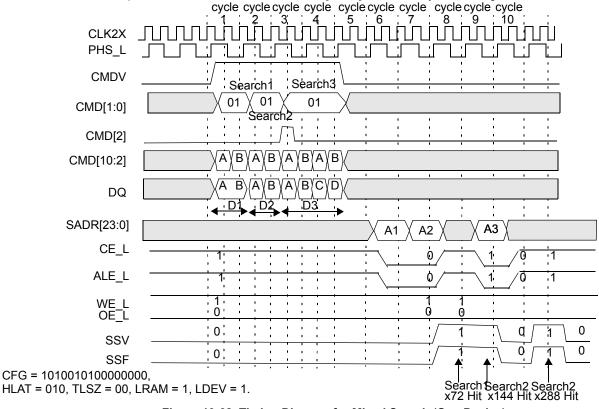


Figure 10-62. Timing Diagram for Mixed Search (One Device)

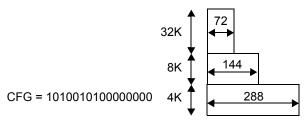


Figure 10-63. Multiwidth Configurations Example

#### Note:

28. The DQ[71:70] will be 00 in each of the two A and B cycles of the ×72-bit Search (Search1). DQ[71:70] is 01 in each of the A and B cycles of the ×144-bit Search (Search2). DQ[71:70] is 10 in each of the A, B, C, and D cycles of the ×288-bit Search (Search3). By having table designation bits, the CYNSE70256 device enables the creation of many tables in a bank of NSEs of different widths.

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10.6.8 Mixed-size Searches on Tables Configured to Different Widths using a CYNSE70256 Device with CFG L HIGH

This subsection will cover the mixed-size searches (×72, ×144 and ×288) with tables of different widths (×72, ×144, ×288) with CFG\_L set HIGH. The previous subsection described searches on tables of different widths using table designation bits in the data array, which can be wasteful. In order to avoid the waste of these bits and yet support up to three tables of ×72, ×144 and ×288 widths, CMD[2] and CMD[9] (in CFG\_L HIGH mode) in cycle A of the command can be used as shown in *Table 10-28*.

Table 10-28. Searches with CFG L Set HIGH

CMD[9]	CMD[2]	Search
0	0	Search 72-bit-configured partitions only.
1	0	Search 144-bit-configured partitions only.
X	1	Cycles A and B for searching 288-bit-configured partitions.
X	0	Cycles C and D for searching 288-bit-configured partitions.

## 10.7 LRAM and LDEV Description

When NSEs are cascaded using multiple CYNSE70256 devices, the SADR, CE\_L, and WE\_L (three-state signals) are all tied together. In order to eliminate external pull-up and pull downs, one device in a bank is designated the default driver. For nonSearch or nonLearn cycles (see Subsection 10.8, "Learn Command") or Search cycles with a global miss, the SADR, CE\_L, and WE\_L signals are driven by the device with the LRAM bit set. It is important that only one device in a bank of cascaded NSEs have this bit set. Failure to do so will cause contention on the SADR, CE\_L, and WE\_L, and can potentially cause damage to the device(s).

Similarly, when NSEs using multiple CYNSE70256 devices are cascaded, SSF and SSV (also three-state signals) are tied together. In order to eliminate external pull-up and pull downs, one device in a bank is designated as the default driver. For nonSearch cycles or Search cycles with a global miss, the SSF and SSV signals are driven by the device with the LDEV bit set in Bank 1. It is important that only one device in a bank of cascaded NSEs have this bit set. Failure to do so will cause contention on the SSV and SSF, and can potentially cause damage to the device(s).

## 10.8 Learn Command

Bit[0] of each 72-bit data location specifies whether an entry in the database is occupied. If all the entries in a device are occupied, the device asserts a FULO signal to inform the downstream devices that it is full. The result of this communication between depth-cascaded devices determines the global FULL signal for the entire table. The FULL signal in the last device determines the fullness of the depth-cascaded table.

The device contains sixteen pairs of internal 72-bit-wide comparand registers that store the comparands as the device executes searches. On a miss by the Search that is signalled to the ASIC through the SSV and SSF signals (SSV = 1, SSF = 0), the host ASIC can apply the Learn command to learn the entry from a comparand register as to the next-free location (see Subsection 7.8, "NFA Register," on page 17).

The NFA register is updated with the new next-free location information following each Write or Learn command. In a depth-cascaded table, only a single device will Learn the entry through the application of a Learn instruction. The determination as to which device will Learn is based on the FULI and FULO signals between the devices. The first non-full device learns the entry by storing the contents of the specified comparand registers to the location(s) pointed to by the NFA register. In a ×72-configured table, the Learn command writes a single 72-bit location. In a ×144-configured table, the Learn command writes the next even and odd 72-bit locations. In 144-bit mode, bit[0] of the even and odd 72-bit locations is 0, indicating that they are cascaded empty, or 1, indicating that they are occupied. The global FULL signal indicates to the table controller (the host ASIC) that all entries within a block are occupied and that no more entries can be learned. The CYNSE70256 device updates the signal after each Write or Learn command to a data array. The Learn command generates a Write cycle to the external SRAM, also using the NFA register as part of the SRAM address (see Section 12.0, "SRAM Addressing," on page 86).

The Learn command is supported on a single block containing up to four devices if the table is configured either as  $\times$ 72 or  $\times$ 144; it is not supported for  $\times$ 288-configured tables. The Learn command is a pipelined operation and lasts for two CLK cycles, as shown in *Figure 10-64* and *Figure 10-65* where TLSZ = 01. *Figure 10-64* and *Figure 10-65* assume that the device performing the Learn operation is not the last device in the table and will therefore have its LRAM bit set to 0.

#### Note:

29. The OE\_L for the device with the LRAM bit set goes HIGH for two cycles for each Learn (one during the SRAM Write cycle and one during the cycle before). The SRAM Write cycle latency from the second cycle of the instruction is shown in *Table 10-29*.

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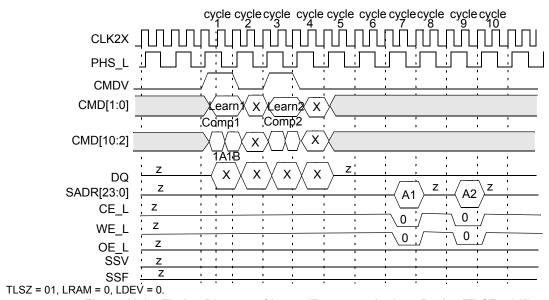


Figure 10-64. Timing Diagram of Learn (Except on the Last Device [TLSZ = 01])

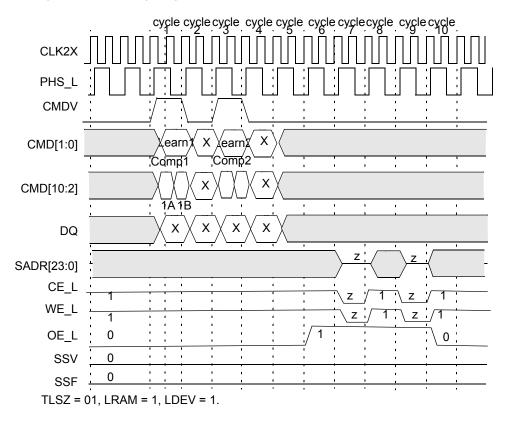


Figure 10-65. Timing Diagram of Learn on Device Number 3 (TLSZ = 01)

Table 10-29. SRAM Write Cycle Latency from Second Cycle of Learn Instruction

Number of Devices	Latency in CLK Cycles
1–4 (TLSZ = 01)	5
1–15 (TLSZ = 10)	6



The Learn operation lasts two CLK cycles. The sequence of operation is as follows.

- Cycle 1A: The host ASIC applies the Learn instruction on CMD[1:0] using CMDV = 1. The CMD[5:2] field specifies the index of the comparand register pair that will be written in the data array in the 144-bit-configured table. For a Learn in a 72-bit-configured table, the even-numbered comparand specified by this index will be written. CMD[8:6] carries the bits that will be driven on SADR[23:21] in the SRAM Write cycle.
- Cycle 1B: The host ASIC continues to drive CMDV to 1, CMD[1:0] to 11, and CMD[5:2] with the comparand pair index. CMD[6] must be set to 0 if the Learn is being performed on a 72-bit-configured table, and to 1 if the Learn is being performed on a 144-bit-configured table.
- Cycle 2: The host ASIC drives CMDV to 0.

At the end of cycle 2, a new instruction can begin. SRAM Write latency is the same as the Search to the SRAM Read cycle. It is measured from the second cycle of the Learn instruction.

## 11.0 Depth Cascading

The NSE application can depth-cascade the devices to various table sizes of different widths (72 bits, 144 bits, or 288 bits). The devices perform all the necessary arbitration to decide which device will drive the SRAM bus. Search latency increases as the table size increases; the Search rate itself remains constant.

## 11.1 Depth Cascading up to Four Devices (One Block)

Figure 11-1 shows that up to four devices can be cascaded to form 512K × 72, 256K × 144, or 128K × 288 tables. It also shows the interconnection between devices for depth cascading. Each NSE asserts the LHO[1] and LHO[0] signals to inform downstream devices of its result. LHI[6:0] signals for a device are connected to LHO signals of the upstream devices. The host ASIC must program the TLSZ to 01 for each of up to four devices in a block. Only a single device drives the SRAM bus in any single cycle.

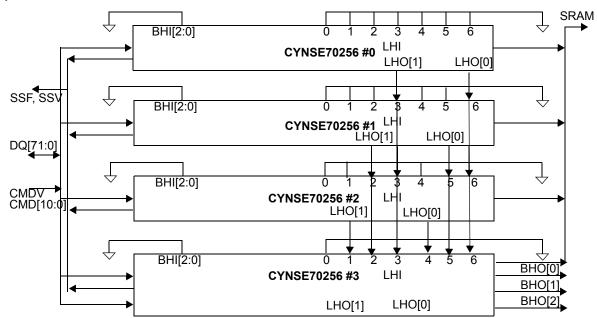


Figure 11-1. Hardware Diagram for a Block of up to Four Devices

### 11.2 Depth Cascading up to Fifteen Devices (Four Blocks)

*Figure 11-2* shows the cascading of up to four blocks. Each block except the last contains up to four CYNSE70256 devices, and the interconnection within each with the cascading of up to four devices in a block was shown in the previous subsection.<sup>[30]</sup>

#### Note

<sup>30.</sup> The interconnection between blocks for depth cascading is important. For each Search, a block asserts BHO[2], BHO[1], and BHO[0]. The BHO[2:0] signals for a block are taken only from the last device in that block. For all other devices within that block, these signals stay open and floating. The host ASIC must program TLSZ to 10 in each of the devices for cascading up to fifteen devices (in up to four blocks).



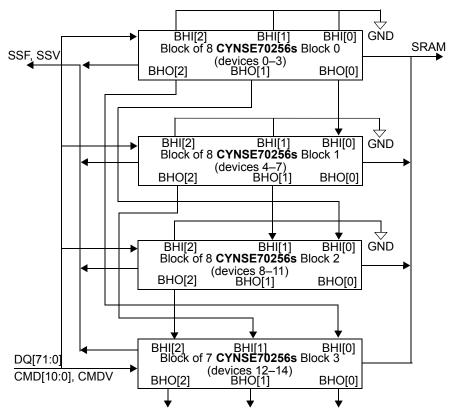


Figure 11-2. Depth Cascading four Blocks

## 11.3 Depth Cascading for a FULL Signal

Bit[0] of each of the 72-bit entries is designated as a special bit (1 = occupied, 0 = empty). For each Learn or PIO Write to the data array, each device asserts FULO[1] or FULO[0] depending on whether or not it has any empty locations within it (see *Figure 11-3*). Each device combines the FULO signals from the devices above it with its own full status to generate a FULL signal that gives the full status of the table up to the device asserting the FULL signal. *Figure 11-3* shows the hardware connection diagram for generating the FULL signal that goes back to the ASIC. In a depth-cascaded block of up to four devices, the FULL signal from the last device should be fed back to the ASIC controller to indicate the fullness of the table. The FULL signal of the other devices should be left open.<sup>[31]</sup>

#### Note:

31. The Learn instruction is supported for only up to four devices, whereas FULL cascading is allowed only for one block in tables containing more than four devices. In tables for which a Learn instruction is not going to be used, the bit[0] of each 72-bit entry should always be set to 1.



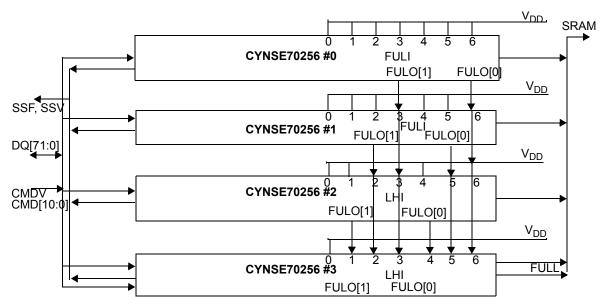


Figure 11-3. FULL Signal Generation in a Cascaded Table

## 12.0 SRAM Addressing

Table 12-1 describes the commands used to generate addresses on the SRAM address bus. The index [15:0] field contains the address of a 72-bit entry that results in a hit in 72-bit-configured quadrant. It is the address of the 72-bit entry that lies at the 144-bit page, and the 288-bit page boundaries in 144-bit- and 288-bit-configured quadrants, respectively.

Section 7.0, "Registers," on page 13 of this datasheet describes the NFA and SSR registers. ADR[15:0] contains the address supplied on the DQ bus during PIO access to the CYNSE70256. Command bits 8, 7, and 6 {CMD[8:6]} are passed from the command to the SRAM address bus. See Section 10.0, "Commands," on page 19, for more information. ID[4:0] is the ID of the device driving the SRAM bus (see Section 18.0, "Pinout Descriptions and Package Diagrams," on page 103, for more information).

Table 12-1.	SRAM Bus	Address
-------------	----------	---------

Command	SRAM Operation	23	22	21	[20:17]	16	[15:0]
Search	Read	C8	C7	C6	ID[4:1]	bank	Index[15:0]
Learn	Write	C8	C7	C6	ID[4:1]	bank	NFA[15:0]
PIO Read	Read	C8	C7	C6	ID[4:1]	bank	ADR16:0]
PIO Write	Write	C8	C7	C6	ID[4:1]	bank	ADR[15:0]
Indirect Access	Write/Read	C8	C7	C6	ID[4:1]	bank	SSR[15:0]

### 12.1 SRAM PIO Access

The remainder of Section 12.0 describes SRAM Read and Write operations.

SRAM Read enables Read access to the off-chip SRAM containing associative data. The latency from the issuance of the Read instruction to the appearance of the address on the SRAM bus is the same as the Search instruction latency, and will depend on the value programmed for the TLSZ parameter in the device configuration register. The latency of the ACK from the Read instruction is the same as that from the Search instruction to the SRAM address latency, plus the HLAT programmed in the configuration register.<sup>[32, 33]</sup>

#### Notes

33. SRAM Write is a pipelined operation.

<sup>32.</sup> SRAM Read is a blocking operation—no new instruction can begin until the ACK is returned by the selected device performing the access. SRAM Write enables Write access to the off-chip SRAM containing associative data. The latency from the second cycle of the Write instruction to the appearance of the address on the SRAM bus is the same as the Search instruction latency, and will depend on the TLSZ value parameter programmed in the device configuration register.



## 12.2 SRAM Read with a Table of up to Four Devices

The following explains the SRAM Read operation completed through a table of up to four devices using the following parameter: TLSZ = 01. Figure 12-1 diagrams a block of four devices. The following assumes that SRAM access is successfully achieved through CYNSE70256 device number 0. Figure 12-2 and Figure 12-3 show timing diagrams for device number 0 and device number 3, respectively.

- Cycle 1A: The host ASIC applies the Read instruction on CMD[1:0] using CMDV = 1. The DQ bus supplies the address, with DQ[20:19] set to 10, to select the SRAM address. The host ASIC selects the device for which ID[4:0] matches the DQ[25:21] lines. During this cycle the host ASIC also supplies SADR[23:21] on CMD[8:6].
- Cycle 1B: The host ASIC continues to apply the Read instruction on CMD[1:0], using CMDV = 1. The DQ bus supplies the address, with DQ[20:19] set to 10, to select the SRAM address.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive DQ[71:0].
- Cycle 5: The selected device continues to drive DQ[71:0] and drives ACK from High-Z to LOW.
- Cycle 6: The selected device drives the Read address on SADR[23:0], and drives ACK HIGH, CE\_L LOW, WE\_L HIGH, and ALE L LOW.
- Cycle 7: The selected device drives CE\_L , ALE\_L, WE\_L, and the DQ bus in a three-state condition. It continues to drive ACK LOW.

At the end of cycle 7, the selected device floats ACK in a three-state condition. A new command can begin.

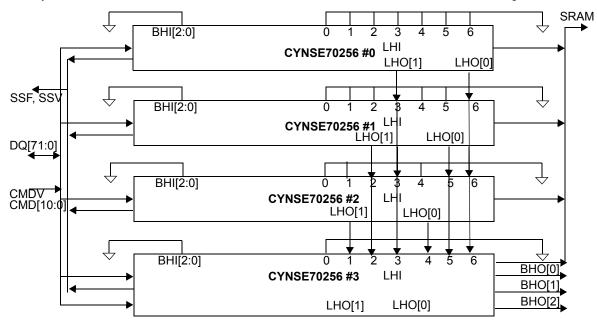


Figure 12-1. Hardware Diagram of a Block of Four Devices



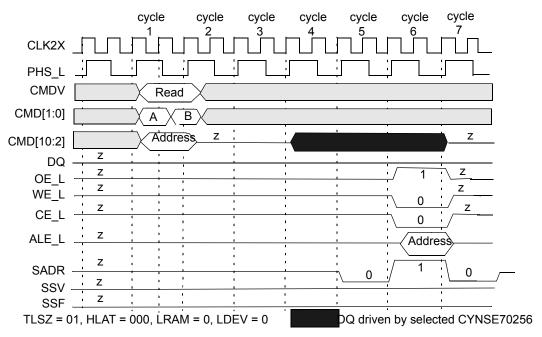


Figure 12-2. SRAM Read Through Device Number 0 in a Block of Four Devices

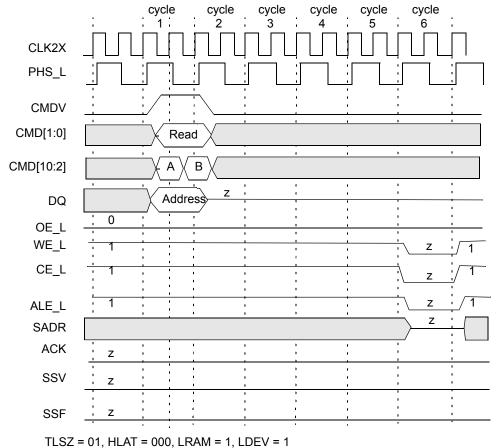


Figure 12-3. SRAM Read Timing for Device Number 7 in a Block of Four Devices



## 12.3 SRAM Read with a Table of up to Fifteen Devices

The following explains the SRAM Read operation accomplished through a table of up to fifteen devices, using the following parameter: TLSZ = 10. The hardware diagram is shown in *Figure 12-4*. The following assumes that SRAM access is being accomplished through CYNSE70256 device number 0, and that device number 0 is the selected device. *Figure 12-5* and *Figure 12-6* show the timing diagrams for device number 0 and device number 14, respectively.

- Cycle 1A: The host ASIC applies the Read instruction to CMD[1:0] using CMDV = 1. The DQ bus supplies the address, with DQ[20:19] set to 10, to select the SRAM address. The host ASIC selects the device for which the ID[4:0] matches the DQ[25:21] lines. During this cycle, the host ASIC also supplies SADR[23:21] on CMD[8:6].
- Cycle 1B: The host ASIC continues to apply the Read instruction to CMD[1:0], using CMDV = 1. The DQ bus supplies the address, with DQ[20:19] set to 10, to select the SRAM address.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive DQ[71:0].
- Cycles 5 to 6: The selected device continues to drive DQ[71:0].
- Cycle 7: The selected device continues to drive DQ[71:0], and drives an SRAM Read cycle.
- Cycle 8: The selected device drives ACL from Z to LOW.
- Cycle 9: The selected device drives ACK to HIGH.
- Cycle 10: The selected device drives ACK from HIGH to LOW.

At the end of cycle 10, the selected device floats ACK in a three-state condition.

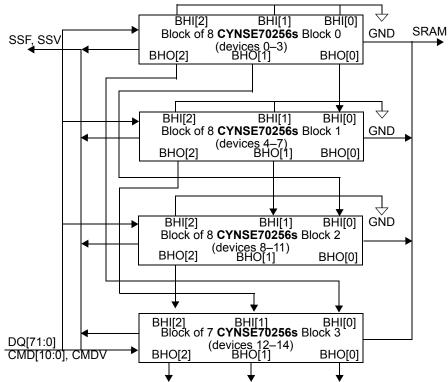


Figure 12-4. Hardware Diagram of Fifteen Devices Using Four Blocks

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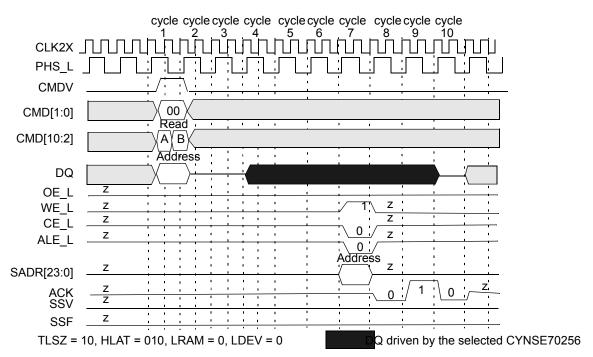


Figure 12-5. SRAM Read Through Device Number 0 in a Bank of Fifteen Devices (Device Number 0 Timing)

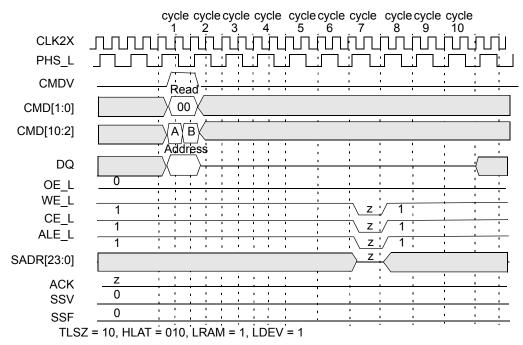


Figure 12-6. SRAM Read Through Device Number 0 in a Bank of Fifteen Devices (Device Number 14 Timing)

## 12.4 SRAM Write with a Table of up to Four Devices

The following explains the SRAM Write operation accomplished through a table(s) of up to four devices with the following parameters (TLSZ = 01). The hardware diagram for this table is shown in *Figure 12-7*. The following assumes that SRAM access is achieved through CYNSE70256 device number 0. *Figure 12-8* and *Figure 12-9* show the timing diagram for device number 0 and device number 3, respectively.



- Cycle 1A: The host ASIC applies the Write instruction on CMD[1:0] using CMDV = 1. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. The host ASIC selects the device for which the ID[4:0] matches the DQ[25:21] lines. The host ASIC also supplies SADR[23:21] on CMD[8:6] in this cycle. [34]
- Cycle 1B: The host ASIC continues to apply the Write instruction on CMD[1:0] using CMDV = 1. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address.<sup>[34]</sup>
- Cycle 2: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the CYNSE70256 device.
- Cycle 3: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the CYNSE70256 device.

At the end of cycle 3, a new command can begin. Write is a pipelined operation, but the Write cycle appears at the SRAM bus with the same latency as that of a Search instruction, as measured from the second cycle of the Write command.

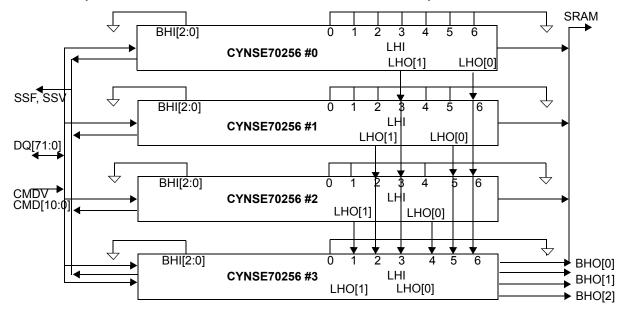


Figure 12-7. Hardware Diagram of a Block of Four Devices

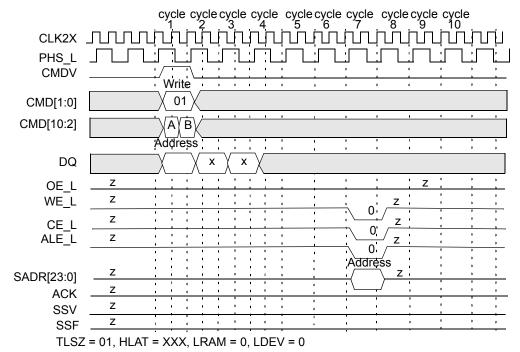


Figure 12-8. SRAM Write Through Device Number 0 in a Block of Four Devices

Note:

34. CMD[2] must be set to 0 for SRAM Write because burst Writes into the SRAM are not supported.



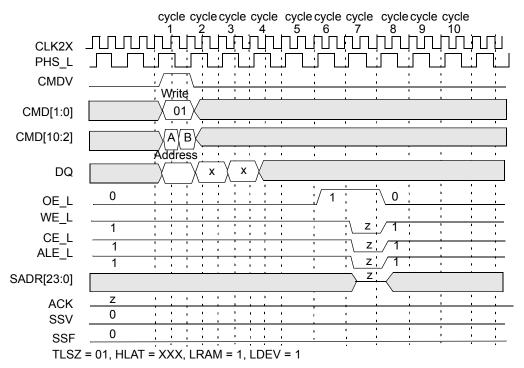


Figure 12-9. SRAM Write Timing for Device Number 3 in Block of Four Devices

## 12.5 SRAM Write with Table(s) Consisting of up to Fifteen Devices

The following explains the SRAM Write operation accomplished through a table of up to fifteen devices with the following parameter: TLSZ = 10. The hardware diagram is shown in *Figure 12-10*. The following assumes that SRAM access is accomplished through the selected device: CYNSE70256 device number 0. *Figure 12-11* and *Figure 12-12* show timing diagrams for device number 0 and device number 14, respectively.

- Cycle 1A: The host ASIC applies the Write instruction on CMD[1:0] using CMDV = 1. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. The host ASIC selects the device for which the ID[4:0] matches the DQ[25:21] lines. The host ASIC also supplies SADR[23:21] on CMD[8:6] in this cycle. [34]
- Cycle 1B: The host ASIC continues to apply the Write instruction on CMD[1:0] using CMDV = 1. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. [34]
- Cycle 2: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the CYNSE70256 device.
- Cycle 3: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the CYNSE70256 device.

At the end of cycle 3, a new command can begin. The Write is a pipelined operation, but the Write cycle appears at the SRAM bus with the same latency as that of a Search instruction, as measured from the second cycle of the Write command.



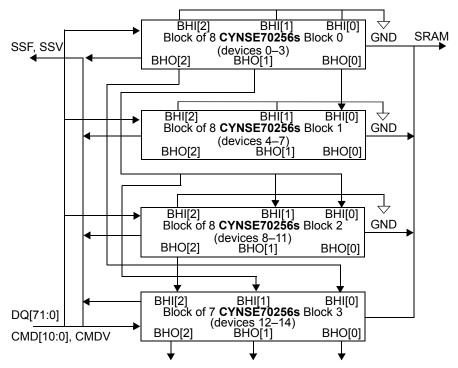


Figure 12-10. Table of Fifteen Devices (Four Blocks)

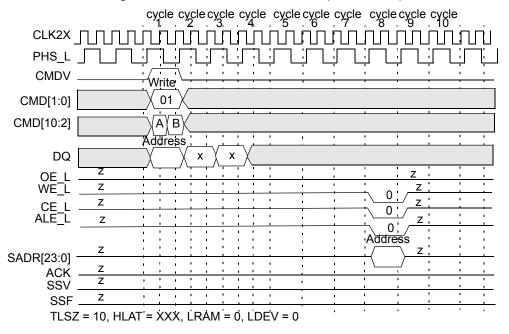


Figure 12-11. SRAM Write Through Device Number 0 in Bank of Fifteen Devices (Device 0 Timing)



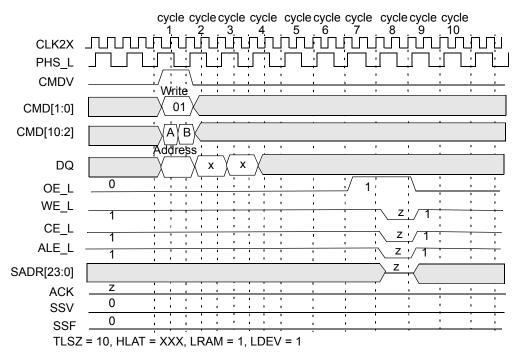


Figure 12-12. SRAM Write Through Device Number 0 in Bank of Fifteen CYNSE70256 Devices (Device Number 14 Timing)

## 13.0 Power

CYNSE70256 has two separate power supplies, one for the core (V<sub>DD</sub>) and another for the IOs (V<sub>DDQ</sub>).

## 13.1 Power-up Sequence

Proper power-up sequence is required to correctly initialize the Cypress Network Search Engines before functional access to the device can begin. RST\_L and TRST\_L should be held low before the power supplies ramp up. RST\_L must be set low for a duration of time afterward and then set high. The following steps describe the proper power-up sequence.

- 1. Set RST\_L and TRST\_L low.
- 2. Power up  $V_{DD}$ ,  $V_{DDQ}$  and start running CLK1X when operating in CLK1X mode or CLK2X and PHS\_L when operating in CLK2X mode. The order in which these signals (including  $V_{DD}$  and  $V_{DDQ}$ ) are applied is not critical.
- 3. RST\_L should be held low for 0.5 ms (PLL lock time requirement). In CLK1X mode, the counting starts on the first rising edge of CLK1X after both V<sub>DD</sub> and V<sub>DDQ</sub> have reached their steady state voltages. In CLK2X mode, the counting starts on the first rising edge of CLK2X when PHS\_L is high, after both V<sub>DD</sub> and V<sub>DDQ</sub> have reached their steady state voltages.
- 4. Continue to hold RST\_L low for a minimum of 32 CLK1X cycles (when operating in CLK1X mode) or 64 CLK2X cycles (when operating in CLK2X mode). Set RST\_L to high afterward to complete the power-up sequence. For JTAG reset, TRST\_L can be brought high after V<sub>DD</sub> and V<sub>DDQ</sub> have both reached their steady state voltages.



Figure 13-1 and Figure 13-2 illustrate the proper sequences of the power-up operation.

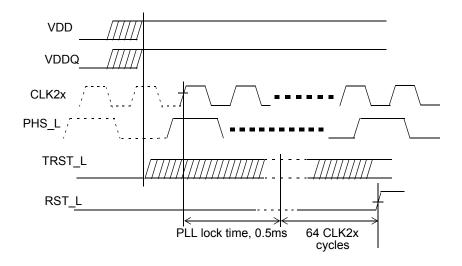


Figure 13-1. Power-up Sequence (CLK2x)

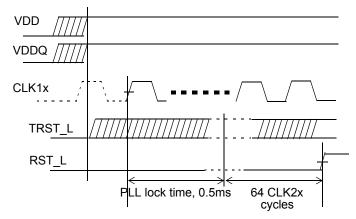
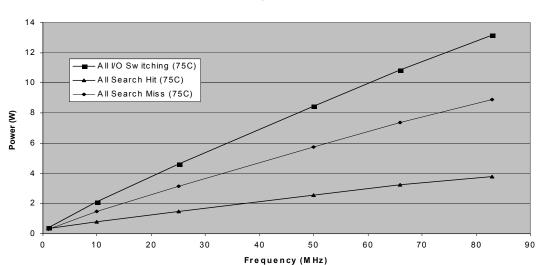


Figure 13-2. Power-up Sequence (CLK1x)

## 13.2 Power Consumption

The following figure depicts expected power consumption over a range of frequencies. The calculations assume 100% of the operations will be SEARCH operations. If an application includes other operations such as READ or WRITE, then power consumption will be lower. The worst case line indicates power consumption when the I/Os switch 100 percent of the time. The other lines (All Search Hit and All Search Miss) assume the I/Os switch 50% of the time.





### Power Consumption of CYNSE70256

Figure 13-3. Power Consumption of CYNSE70256

## 14.0 Application

Figure 14-1 shows how an NSE subsystem can be formed using a host ASIC and a CYNSE70256 bank. It also shows how this NSE subsystem is integrated in a switch or router. The CYNSE70256 device can access synchronous and asynchronous SRAMs by allowing the host ASIC to set the same HLAT parameter in all NSEs within a bank of NSEs.

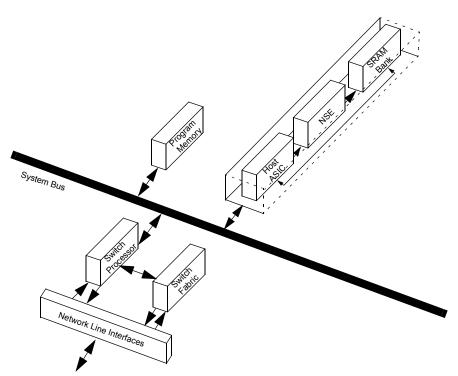


Figure 14-1. Sample Switch/Router Using the CYNSE70256 Device

## 15.0 JTAG (1149.1) Testing

The CYNSE70256 device supports the Test Access Port and Boundary Scan Architecture as specified in IEEE JTAG Standard Number 1149.1. The pin interface to the chip consists of five signals with the standard definitions: TCK, TMS, TDI, TDO, and TRST\_L. *Table 15-1* describes the operations that the test access port controller supports, and *Table 15-2* describes the TAP



Device ID Register.<sup>[35]</sup> **Note**: To disable JTAG functionality, connect the TCK, TMS and TDI pins to VDDQ through a pull-up, and TRST\_L to ground through a pull-down **Table 15-1. Supported Operations** 

Instruction	Type	Description
SAMPLE/PRELOA D		Sample/Preload. This operation loads the values of signals going to and from I/O pins into the boundary scan shift register to provide a snapshot of the normal functional operation.
EXTEST	Mandatory	<b>External Test</b> . This operation uses boundary scan values shifted in from the TAP to test connectivity external to the device.
BYPASS	Mandatory	<b>Bypass</b> . This operation bypasses the device in a JTAG chain by loading a single bit shift register between TDI and TDO and provides a minimum-length serial path when no test operation is required.
IDCODE	Optional	<b>Device JTAG ID Code</b> . This operation selects the JTAG Identification register and output the IDCODE field serially through TDO.
CLAMP	Optional	Output Clamp. This operation drives preset values onto the outputs of the device.
HIGHZ	Optional	High-Z Output. This operation sets the device output signals in high impedance state.

Table 15-2. TAP Device ID Register

Field	Range	Initial Value	Description
Revision	[31:28]	0001	<b>Revision Number.</b> This is the current device revision number. Numbers start from one and increment by one for each revision of the device.
Part Number	[27:12]	0000 0000 0000 0100	This is the part number for the device.
MFID	[11:1]	000_1101_1100	<b>Manufacturer ID</b> . This field is the same as the manufacturer ID used in the TAP controller.
LSB	[0]	1	Least significant bit.

#### 16.0 **Electrical Specifications**

This section describes the electrical specifications, capacitance, operating conditions, DC characteristics, and AC timing parameters for the CYNSE70256 device (see *Table 16-1* and *Table 16-2*).

Table 16-1. DC Electrical Characteristics for CYNSE70256

Parameter	Description	Conditions	Min.	Max.	Unit
ILI	Input leakage current	$V_{DDQ} = V_{DDQ} Max, V_{IN} = 0 \text{ to } V_{DDQ} Max$	-20	20	μА
I <sub>LO</sub>	Output leakage current	$V_{DDQ} = V_{DDQ} Max, V_{IN}=0 \text{ to } V_{DDQ} Max$	-20	20	μА
V <sub>IL</sub>	Input LOW voltage (2.5V)		-0.3	0.7	V
V <sub>IH</sub>	Input HIGH voltage (2.5V)		1.7	V <sub>DDQ</sub> + 0.3	V
V <sub>OL</sub>	Output LOW voltage (2.5V)	V <sub>DDQ</sub> = V <sub>DDQ</sub> Min, I <sub>OL</sub> = 8mA		0.4	V
V <sub>OH</sub>	Output HIGH voltage (2.5V)	V <sub>DDQ</sub> = V <sub>DDQ</sub> Min, I <sub>OH</sub> = 8mA	2.0		V
V <sub>IL</sub>	Input LOW voltage (3.3V)		-0.3	0.8	
V <sub>IH</sub>	Input HIGH voltage (3.3V)		2.0	V <sub>DDQ</sub> + 0.3	
$V_{OL}$	Output LOW voltage (3.3V)	V <sub>DDQ</sub> = V <sub>DDQ</sub> Min, I <sub>OL</sub> = 8mA		0.4	V
V <sub>OH</sub>	Output HIGH voltage (3.3V)	V <sub>DDQ</sub> = V <sub>DDQ</sub> Min, I <sub>OH</sub> = 8mA	2.4		V
I <sub>DD2</sub>	3.3V/2.5V supply current at V <sub>DD</sub> Max	83-MHz search rate, I <sub>OUT</sub> = 0mA		600	mA
I <sub>DD2</sub>	3.3V/2.5V supply current at V <sub>DD</sub> Max	66-MHz search rate, I <sub>OUT</sub> = 0mA		480	mA
I <sub>DDI</sub>	1.5V supply current at V <sub>DD</sub> (Typical)	83-MHz search rate		7.4 <sup>[41]</sup>	Α
I <sub>DDI</sub>	1.5V supply current at V <sub>DD</sub> (Typical)	66-MHz search rate		5.9 <sup>[41]</sup>	Α

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<sup>35.</sup> To disable JTAG functionality, connect the TCK, TMS, and TDI pins toVDDQ through a pull-up, and TRST\_L to ground through a pull-down.



Parameter	Description	Max.	Unit
C <sub>IN</sub>	Input capacitance	12	pF <sup>[36]</sup>
C <sub>OUT</sub>	Output capacitance	12	pF <sup>[37]</sup>

## Table 16-2. Operating Conditions for CYNSE70256

Parameter	Description	Min.	Max.	Unit
V <sub>DDQ</sub> = 3.3V	Operating voltage for I/O	3.135	3.465	V
V <sub>DDQ</sub> = 2.5V	Operating voltage for I/O	2.375	2.625	V
V <sub>DD</sub>	Operating supply voltage	1.425	1.575	V
V <sub>IH</sub>	Input HIGH voltage <sup>[39]</sup> (2.5V)	1.7	V <sub>DDQ</sub> + 0.3	V
V <sub>IL</sub>	Input LOW voltage <sup>[38]</sup> (2.5V)	-0.3	0.7	V
V <sub>IH</sub>	Input HIGH voltage <sup>[40]</sup> (3.3V)	2.0	V <sub>DDQ</sub> + 0.3	V
V <sub>IL</sub>	Input LOW voltage <sup>[38]</sup> (3.3V)	-0.3	0.8	V
T <sub>A</sub>	Ambient operating temperature	0	70	°C
T <sub>J</sub>	Junction temperature	0	150 <sup>[42]</sup>	°C
	Supply voltage tolerance	-5%	+5%	

## Notes:

- 36. f = 1 MHz, V<sub>IN</sub> = 0 V.
  37. f = 1 MHz, V<sub>OUT</sub> = 0 V.
  38. Minimum allowable applies to undershoot only.
  39. Maximum allowable applies to overshoot only (V<sub>DDQ</sub> is 2.5V supply).
  40. Maximum allowable applies to overshoot only (V<sub>DDQ</sub> is 3.3V supply).
  41. Typical. 80% Compare utilization.
  42. Please refer to "CYNSE70256 Airflow and Heat Sink Requirements" application note.



#### 17.0 **AC Timing Waveforms**

Table 17-1 and Table 17-2 show the AC timing parameters for the CYNSE70256 device. Table 17-3 shows the AC test conditions for the CYNSE70256 device. Figure 17-1 shows the input wave form for the CYNSE70256 device. Figure 17-2 and Figure 17-3 show the output load and output load equivalent of the CYNSE70256 device. Figure 17-4 shows timing wave form diagrams for CLK2X. Figure 17-5 details timing wave form diagrams for CLK1X.

Table 17-1. AC Timing Parameters with CLK2X

		CYNSE70256-066		CYNSE7	0256-083	
Parameter	Description	Min.	Max.	Min.	Max.	Unit
f <sub>CLOCK</sub>	CLK2X frequency.	40	133	40	166	MHz
t <sub>CLOK</sub>	PLL lock time.		0.5		0.5	ms
t <sub>CKHI</sub>	CLK2X HIGH pulse. <sup>[43]</sup>	3.0		2.4		ns
t <sub>CKLO</sub>	CLK2X LOW pulse. <sup>[43]</sup>	3.0		2.4		ns
t <sub>ISCH</sub>	Input setup time to CLK2X rising edge.[43]	2.5		1.8		ns
t <sub>IHCH</sub>	Input hold time to CLK2X rising edge.[43]	0.6		0.6		ns
t <sub>ICSCH</sub>	C]ascaded input setup time to CLK2X rising edge. <sup>[43]</sup>	4.2		3.5		ns
t <sub>ICHCH</sub>	Cascaded input hold time to CLK2X rising edge. [43]	2.0		2.0		ns
t <sub>CKHOV</sub>	Rising edge of CLK2X to LHO, FULO, BHO, FULL valid. [44]		8.5		7.0	ns
t <sub>CKHDV</sub>	Rising edge of CLK2X to DQ valid. [44]		9.0		7.5	ns
t <sub>CKHDZ</sub>	Rising edge of CLK2X to DQ HIGH-Z. <sup>[45]</sup>	0.5	8.5	0.5	7.0	ns
t <sub>CKHSV</sub>	Rising edge of CLK2X to SRAM bus valid. [44]		9.0		7.5	ns
t <sub>CKHSHZ</sub>	Rising edge of CLK2X to SRAM bus HIGH-Z. <sup>[45]</sup>	0.5	6.5	0.5	6.0	ns
t <sub>CKHSLZ</sub>	Rising edge of CLK2X to SRAM bus LOW-Z.[45]	7.0		6.5		ns

Table 17-2. AC Timing Parameters with CLK1X

		CYNSE7	0256-066	CYNSE7		
Parameter	Description	Min	Max	Min	Max	Unit
f <sub>CLOCK</sub>	CLK1X frequency.	20	66	20	83	MHz
t <sub>CLOK</sub>	PLL lock time.		0.5		0.5	ms
t <sub>CKHI</sub>	CLK1X HIGH pulse; worst-case duty cycle. <sup>[46]</sup>	6.75		5.4		ns
t <sub>CKLO</sub>	CLK1X LOW pulse; worst-case duty cycle. <sup>[46]</sup>	6.75		5.4		ns
t <sub>ISCH</sub>	Input setup time to CLK1X edge. <sup>[46]</sup>	2.5		1.8		ns
t <sub>IHCH</sub>	Input hold time to CLK1X edge. <sup>[46]</sup>	0.6		0.6		ns
t <sub>ICSCH</sub>	Cascaded input setup time to CLK1X rising edge. [46]	4.2		3.5		ns
t <sub>ICHCH</sub>	Cascaded input hold time to CLK1X rising edge.[46]	2.0		2.0		ns
t <sub>CKHOV</sub>	Rising edge of CLK1X to LHO, FULO, BHO, FULL valid.[47]		8.5		7.0	ns
t <sub>CKHDV</sub>	Rising edge of CLK1X to DQ valid. [47]		9.0		7.5	ns
t <sub>CKHDZ</sub>	Rising edge of CLK1X to DQ HIGH-Z. <sup>[48]</sup>	0.5	8.5	0.5	7.0	ns
t <sub>CKHSV</sub>	Rising edge of CLK1X to SRAM bus valid. [47]		9.0		7.5	ns
t <sub>CKHSHZ</sub>	Rising edge of CLK1X to SRAM bus HIGH-Z.[48]	0.5	6.5	0.5	6.0	ns
t <sub>CKHSLZ</sub>	Rising edge of CLK1X to SRAM bus LOW-Z.[48]	7.0		6.5		ns

#### Notes:

- 43. Values are based on 50% signal levels. 44. Based on an AC load of  $C_L$  = 30 pF (see Figure 17-1, Figure 17-2, and Figure 17-3).

- 45. These parameters are sampled but not 100% tested, and are based on an AC load of 5 pF.
  46. Values are based on 50% signal levels and a 50%/50% duty cycle of CLK1X.
  47. Based on an AC load of C<sub>L</sub> = 30 pF (see *Figure 17-1*, *Figure 17-2*, and *Figure 17-3*).
  48. These parameters are sampled but not 100% tested, and are based on an AC load of 5 pF.



Table 17-3. AC Table for Test Condition of CYNSE70256

Conditions	Results
Input pulse levels (V <sub>DDQ</sub> =3.3V)	GND to 3.3V
Input pulse levels (V <sub>DDQ</sub> =2.5V)	GND to 2.5V
Input rise and fall times measured at 0.3V and 2.7V (V <sub>DDQ</sub> =3.3V)	≤ 2 ns (see Figure 17-1)
Input rise and fall times measured at 0.25V and 2.25V (V <sub>DDQ</sub> =2.5V)	≤ 2 ns (see Figure 17-1)
Input timing reference levels (V <sub>DDQ</sub> =3.3V)	1.65V
Input timing reference levels (V <sub>DDQ</sub> =2.5V)	1.25V
Output reference levels (V <sub>DDQ</sub> =3.3V)	1.65V
Output reference levels (V <sub>DDQ</sub> =2.5V)	1.25V
Output load	See Figure 17-2 and Figure 17-3

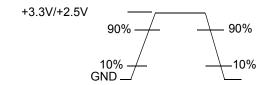


Figure 17-1. Input Wave Form for CYNSE70256

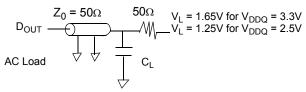


Figure 17-2. Output Load for CYNSE70256

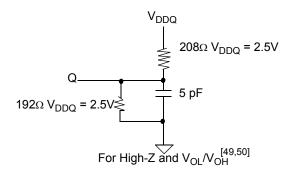


Figure 17-3. 2.5 I/O Output Load Equivalent for CYNSE70256

#### Notes:

49. Output loading is specified with CL = 5pF as in Figure 17-3. Transition is measured at  $\pm$  200 mV from steady state voltage. 50. The load used for  $V_{OH}$ ,  $V_{OL}$  testing is shown in Figure 17-3.



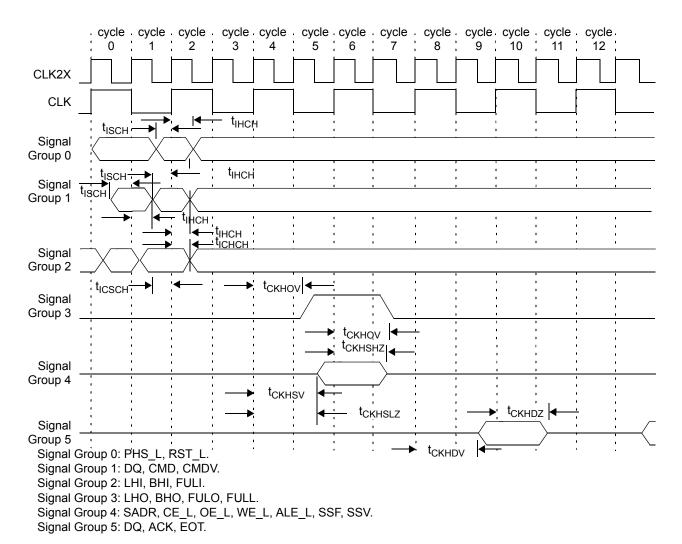


Figure 17-4. AC Timing Wave Forms with CLK2X



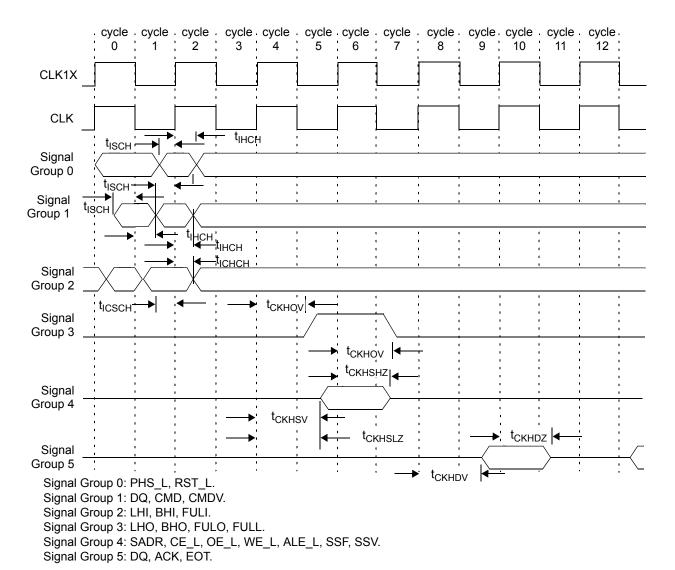


Figure 17-5. AC Timing Wave Forms with CLK1X



# 18.0 Pinout Descriptions and Package Diagrams

	AF	AE	AD	AC	AB	AA	Υ	W	٧	U	т	R	Р	N	М	L	K	J	н	G	F	E	D	С	В	Α	
1	NC	V <sub>SS</sub>	RST_L	V <sub>SS</sub>	FULL	FULO1	FULI6	$V_{DDQ}$	FULI2	FULI0	BHO2	$V_{DDQ}$	BHO0	BHI1	$V_{DDQ}$	LHO0	LHI6	LHI2	LHI0	ID3	ID1	ID0	TRST_L	TCK	TDI	$V_{DDQ}$	1
2	V <sub>SS</sub>	V <sub>SS</sub>	$V_{DDQ}$	EOT	ACK	$V_{DDQ}$	FULO0	FULI5	FULI3	$V_{DDQ}$	V <sub>SS</sub>	BHO1	NC	BHI2	BHI0	LHO1	LHI4	LHI3	LHI1	ID4	ID2	$V_{DDQ}$	TDO	TMS	V <sub>SS</sub>	DQ71	2
3	DQ68	DQ70	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	NC2	FULI4	FULI1	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	V <sub>DD</sub>	LHI5	$V_{DDQ}$	NC1	$V_{DD}$	V <sub>DD</sub>	$V_{DD}$	$V_{DD}$	V <sub>DD</sub>	DQ69	$V_{DDQ}$	3
4	DQ66	$V_{DDQ}$	V <sub>DD</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	$V_{DD}$	V <sub>DD</sub>	$V_{DD}$	V <sub>DD</sub>	$V_{DD}$	V <sub>DD</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>DD</sub>	DQ65	DQ67	4
5	DQ62	DQ64	V <sub>DD</sub>	V <sub>SS</sub>																			V <sub>SS</sub>	V <sub>DD</sub>	DQ61	DQ63	5
6	V <sub>DDQ</sub>	DQ60	V <sub>DD</sub>	V <sub>SS</sub>																			V <sub>SS</sub>	V <sub>DD</sub>	DQ59	$V_{DDQ}$	6
7	DQ56	DQ58	V <sub>DD</sub>	V <sub>SS</sub>																			V <sub>SS</sub>	V <sub>DD</sub>	DQ55	DQ57	7
8	DQ52	DQ54	NC3	V <sub>SS</sub>																			V <sub>SS</sub>	NC8	$V_{DDQ}$	DQ53	8
9	DQ48	DQ50	$V_{DDQ}$	V <sub>SS</sub>																			V <sub>SS</sub>	DQ49	DQ47	DQ51	9
10	$V_{DDQ}$	DQ44	DQ46	V <sub>SS</sub>																			V <sub>SS</sub>	$V_{\rm DDQ}$	DQ45	DQ43	10
11	DQ40	DQ42	$V_{DD}$	$V_{DD}$							V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>							$V_{DD}$	$V_{DD}$	DQ39	DQ41	11
12	DQ36	DQ38	$V_{DD}$	$V_{DD}$							V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>							$V_{DD}$	$V_{DD}$	$V_{\rm DDQ}$	DQ37	12
13	DQ34	$V_{\rm DDQ}$	$V_{DD}$	$V_{DD}$							V <sub>SS</sub>	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>							$V_{DD}$	V <sub>DD</sub>	DQ33	DQ35	13
14	DQ30	DQ32	$V_{DD}$	$V_{DD}$							V <sub>SS</sub>	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>							$V_{DD}$	V <sub>DD</sub>	DQ29	DQ31	14
15	$V_{DDQ}$	DQ28	$V_{DD}$	$V_{DD}$							V <sub>SS</sub>	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub>							$V_{DD}$	V <sub>DD</sub>	DQ27	$V_{DDQ}$	15
16	DQ24	DQ26	$V_{DD}$	$V_{DD}$							V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>							$V_{DD}$	$V_{DD}$	DQ23	DQ25	16
17	DQ22	$V_{\rm DDQ}$	DQ20	V <sub>SS</sub>																			V <sub>SS</sub>	DQ19	$V_{\rm DDQ}$	DQ21	17
18	DQ14	DQ18	DQ16	V <sub>SS</sub>																			V <sub>SS</sub>	DQ13	DQ15	DQ17	18
19	$V_{DDQ}$	DQ12	NC4	V <sub>SS</sub>																			V <sub>SS</sub>	NC7	DQ11	$V_{DDQ}$	19
20	DQ08	DQ10	$V_{DD}$	V <sub>SS</sub>																			V <sub>SS</sub>	$V_{DD}$	DQ07	DQ09	20
21	DQ04	DQ06	$V_{DD}$	V <sub>SS</sub>																			V <sub>SS</sub>	$V_{DD}$	$V_{DDQ}$	DQ05	21
22	DQ02	$V_{DDQ}$	$V_{DD}$	V <sub>SS</sub>																			V <sub>SS</sub>	$V_{DD}$	DQ01	DQ03	22
23	SSV	DQ00	$V_{DD}$	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub>	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	V <sub>DD</sub>	V <sub>SS</sub>	V <sub>DD</sub>	V <sub>SS</sub>	V <sub>SS</sub>	23						
24	SSF	$V_{DDQ}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	NC	CE_L	OE_L	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	V <sub>DD</sub>	SADR13	SADR11	NC6	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	$V_{DD}$	CFG_L	$V_{DDQ}$	24
25	CMD10	V <sub>SS</sub>	CMD8	CMD6	CMD5	CMD3	CMD1	CMDV	$V_{\rm DDQ}$	PHS_L	CLK_M ODE	SADR22	SADR21	SADR19	$V_{\rm DDQ}$	SADR15	$V_{DDQ}$	SADR12	$V_{DDQ}$	SADR08	SADR06	SADR05	SADR03	SADR01	$V_{SS}$	HIGH_S PEED	25
26	CMD9	V <sub>SS</sub>	CMD7	V <sub>DDQ</sub>	CMD4	CMD2	CMD0	ALE_L	WE_L	CLK1X/ CLK2X	SADR23	V <sub>DDQ</sub>	SADR20	SADR18	SADR17	SADR16	SADR14	SADR10	SADR09	SADR07	$V_{DDQ}$	SADR04	SADR02	V <sub>DDQ</sub>	SADR00	$V_{DDQ}$	26
	AF	AE	AD	AC	AB	AA	Y	w	V	U	Т	R	Р	N	М	L	K	J	Н	G	F	Е	D	С	В	A	

Figure 18-1. Pinout Diagram

Table 18-1. Pinout Descriptions for Pinout Diagram

Package Ball Number	Signal Name	Signal Type	Package Ball Number	Signal Name	Signal Type	
A1	V <sub>DDQ</sub> <sup>[51]</sup>	2.5V/3.3V	AA26	CMD[2]	Input	
A10	DQ[43]	I/O	AA3	$V_{DD}$	1.5V	
A11	DQ[41]	I/O	AA4	$V_{SS}$	Ground	
A12	DQ[37]	I/O	AB1	FULL	Output	
A13	DQ[35]	I/O	AB2	ACK	Output-T	
A14	DQ[31]	I/O	AB23	VSS	Ground	
A15	$V_{\mathrm{DDQ}}$	2.5V/3.3V	AB24	$V_{DD}$	1.5V	
A16	DQ[25]	I/O	AB25	CMD[5]	Input	
A17	DQ[21]	I/O	AB26	CMD[4]	Input	
A18	DQ[17]	I/O	AB3	$V_{DD}$	1.5V	
A19	$V_{\mathrm{DDQ}}$	2.5V/3.3V	AB4	$V_{SS}$	Ground	
A2	DQ[71]	I/O	AC1	$V_{SS}$	Ground	
A20	DQ[09]	I/O	AC10	$V_{SS}$	Ground	
A21	DQ[05]	I/O	AC11	$V_{DD}$	1.5V	
A22	DQ[03]	I/O	AC12	$V_{DD}$	1.5V	
A23	V <sub>SS</sub>	Ground	AC13	$V_{DD}$	1.5V	
A24	$V_{\mathrm{DDQ}}$	2.5V/3.3V	AC14	$V_{DD}$	1.5V	
A25	HIGH_SPEED	Ground	AC15	$V_{DD}$	1.5V	
A26	$V_{\mathrm{DDQ}}$	2.5V/3.3V	AC16	$V_{DD}$	1.5V	

Note:

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<sup>51.</sup> All  $V_{\text{DDQ}}$  pins should be set to either 2.5V or 3.3V.



 Table 18-1. Pinout Descriptions for Pinout Diagram (continued)

Package Ball Number	Signal Name	Signal Type	Package Ball Number	Signal Name	Signal Type	
A3	$V_{\mathrm{DDQ}}$	2.5V/3.3V	AC17	$V_{SS}$	Ground	
A4	DQ[67]	I/O	AC18	$V_{SS}$	Ground	
A5	DQ[63]	I/O	AC19	$V_{SS}$	Ground	
A6	$V_{\mathrm{DDQ}}$	2.5V3.3V	AC2	EOT	Output-T	
A7	DQ[57]	I/O	AC20	$V_{SS}$	Ground	
A8	DQ[53]	I/O	AC21	$V_{SS}$	Ground	
A9	DQ[51]	I/O	AC22	$V_{SS}$	Ground	
AA1	FULO[1]	Output	AC23	$V_{SS}$	Ground	
AA2	$V_{\mathrm{DDQ}}$	2.5V/3.3V	AC24	$V_{DD}$	1.5V	
AA23	VSS	Ground	AC25	CMD[6]	Input	
AA24	$V_{DD}$	1.5V	AC26	$V_{\mathrm{DDQ}}$	2.5V/3.3V	
AA25	CMD[3]	Input	AC3	$V_{DD}$	1.5V	
AC4	V <sub>SS</sub>	Ground	AE10	DQ[44]	I/O	
AC5	V <sub>SS</sub>	Ground	AE11	DQ[42]	I/O	
AC6	V <sub>SS</sub>	Ground	AE12	DQ[38]	I/O	
AC7	V <sub>SS</sub>	Ground	AE13	$V_{\mathrm{DDQ}}$	2.5V/3.3V	
AC8	V <sub>SS</sub>	Ground	AE14	DQ[32]	I/O	
AC9	V <sub>SS</sub>	Ground	AE15	DQ[28]	I/O	
AD1	RST_L	Input	AE16	DQ[26]	I/O	
AD10	DQ[46]	I/O	AE17	$V_{DDQ}$	2.5V/3.3V	
AD11	V <sub>DD</sub> 1.5V		AE18	DQ[18]	I/O	
AD12	$V_{DD}$	1.5V	AE19	DQ[12]	I/O	
AD13	$V_{DD}$	1.5V	AE2	$V_{SS}$	Ground	
AD14	$V_{DD}$	1.5V	AE20	DQ[10]	I/O	
AD15	$V_{DD}$	1.5V	AE21	DQ[06]	I/O	
AD16	$V_{DD}$	1.5V	AE22	$V_{DDQ}$	2.5V/3.3V	
AD17	DQ[20]	I/O	AE23	DQ[00]	I/O	
AD18	DQ[16]	I/O	AE24	$V_{\mathrm{DDQ}}$	2.5V/3.3V	
AD19	NC	No Connect	AE25	$V_{SS}$	Ground	
AD2	$V_{\mathrm{DDQ}}$	2.5V/3.3V	AE26	$V_{SS}$	Ground	
AD20	$V_{\mathrm{DD}}$	1.5V	AE3	DQ[70]	I/O	
AD21	$V_{DD}$	1.5V	AE4	$V_{DDQ}$	2.5V/3.3V	
AD22	$V_{DD}$	1.5V	AE5	DQ[64]	I/O	
AD23	$V_{DD}$	1.5V	AE6	DQ[60]	I/O	
AD24	$V_{DD}$	1.5V	AE7	DQ[58]	I/O	
AD25	CMD[8]	Input	AE8	DQ[54]	I/O	
AD26	CMD[7]	Input	AE9	DQ[50]	I/O	
AD3	$V_{DD}$	1.5V	AF1	NC	No Connect	
AD4	$V_{DD}$	1.5V	AF10	$V_{DDQ}$	2.5V/3.3V	
AD5	V <sub>DD</sub>	1.5V	AF11	DQ[40]	I/O	
AD6	V <sub>DD</sub>	1.5V	AF12	DQ[36]	I/O	
AD7	V <sub>DD</sub>	1.5V	AF13	DQ[34]	I/O	
AD8	NC	No Connect	AF14	DQ[30]	I/O	



 Table 18-1. Pinout Descriptions for Pinout Diagram (continued)

Package Ball Number	Signal Name	Signal Type	Package Ball Number	Signal Name	Signal Type
AD9	$V_{\mathrm{DDQ}}$	2.5V/3.3V	AF15	$V_{\mathrm{DDQ}}$	2.5V/3.3V
AE1	V <sub>SS</sub>	Ground	AF16	DQ[24]	I/O
AF17	DQ[22]	I/O	B23	V <sub>SS</sub>	Ground
AF18	DQ[14]	I/O	B24	CFG_L	Input
AF19	$V_{\mathrm{DDQ}}$	2.5V/3.3V	B25	V <sub>SS</sub>	Ground
AF2	V <sub>SS</sub>	Ground	B26	SADR[0]	Output-T
AF20	DQ[08]	I/O	В3	DQ[69]	I/O
AF21	DQ[04]	I/O	B4	DQ[65]	I/O
AF22	DQ[02]	I/O	B5	DQ[61]	I/O
AF23	SSV	Output-T	B6	DQ[59]	I/O
AF24	SSF	Output-T	B7	DQ[55]	I/O
AF25	CMD[10]	Input	B8	$V_{DDQ}$	2.5V/3.3V
AF26	CMD[9]	Input	B9	DQ[47]	I/O
AF3	DQ[68]	I/O	C1	TCK	Input
AF4	DQ[66]	I/O	C10	V <sub>DDQ</sub>	2.5V/3.3V
AF5	DQ[62]	I/O	C11	V <sub>DD</sub>	1.5V
AF6	V <sub>DDQ</sub>	2.5V/3.3V	C12	V <sub>DD</sub>	1.5V
AF7	DQ[56]	I/O	C13	V <sub>DD</sub>	1.5V
AF8	DQ[52]	I/O	C14	V <sub>DD</sub>	1.5V
AF9	DQ[48]	I/O	C15	V <sub>DD</sub>	1.5V
B1	TDI	Input	C16	V <sub>DD</sub>	1.5V
B10	DQ[45]	I/O	C17	DQ[19]	I/O
B11	DQ[39]	I/O	C18	DQ[13]	I/O
B12	V <sub>DDQ</sub>	2.5V/3.3V	C19	NC NC	No Connect
B13	DQ[33]	1/0	C2	TMS	Input
B14	DQ[29]	I/O	C20	V <sub>DD</sub>	1.5V
B15	DQ[27]	I/O	C21	V <sub>DD</sub>	1.5V
B16	DQ[23]	I/O	C22	V <sub>DD</sub>	1.5V
B17	V <sub>DDQ</sub>	2.5V/3.3V	C23	V <sub>DD</sub>	1.5V
B18	DQ[15]	I/O	C24	V <sub>DD</sub>	1.5V
B19	DQ[13]	I/O	C25	SADR[1]	Output-T
B2	V <sub>SS</sub>	Ground	C26	V <sub>DDQ</sub>	2.5V/3.3V
B20	DQ[07]	I/O	C3	V <sub>DD</sub>	1.5V
B20 B21	= =	2.5V/3.3V	C4	V <sub>DD</sub>	1.5V
B21	V <sub>DDQ</sub> DQ[01]	I/O	C5		1.5V
C6	= =	1.5V	E24	V <sub>DD</sub>	1.5V
C7	V <sub>DD</sub>	1.5V	E24 E25	V <sub>DD</sub> SADR[5]	Output-T
C8	V <sub>DD</sub> NC	No Connect	E26	SADR[5] SADR[4]	Output-T
C8		I/O	E20		1.5V
	DQ[49]			V <sub>DD</sub>	
D1	TRST_L	Input	E4	V <sub>SS</sub>	Ground
D10			F1	ID[1]	Input
D11	V <sub>DD</sub>	1.5V	F2	ID[2]	Input
D12	V <sub>DD</sub>	1.5V	F23	V <sub>SS</sub>	Ground



 Table 18-1. Pinout Descriptions for Pinout Diagram (continued)

Package Ball Number	Signal Name	Signal Type	Package Ball Number	Signal Name	Signal Type	
D13	$V_{DD}$	1.5V	F24	$V_{DD}$	1.5V	
D14	$V_{DD}$	1.5V	F25	SADR[6]	Output-T	
D15	$V_{DD}$	1.5V	F26	$V_{DDQ}$	2.5V/3.3V	
D16	$V_{DD}$	1.5V	F3	$V_{DD}$	1.5V	
D17	V <sub>SS</sub>	Ground	F4	V <sub>SS</sub>	Ground	
D18	V <sub>SS</sub>	Ground	G1	ID[3]	Input	
D19	V <sub>SS</sub>	Ground	G2	ID[4]	Input	
D2	TDO	Output-T	G23	$V_{SS}$	Ground	
D20	V <sub>SS</sub>	Ground	G24	$V_{DD}$	1.5V	
D21	V <sub>SS</sub>	Ground	G25	SADR[8]	Output-T	
D22	V <sub>SS</sub>	Ground	G26	SADR[7]	Output-T	
D23	V <sub>SS</sub>	Ground	G3	$V_{\mathrm{DD}}$	1.5V	
D24	V <sub>DD</sub>	1.5V	G4	V <sub>SS</sub>	Ground	
D25	SADR[3]	Output-T	H1	LHI[0]	No Connect	
D26	SADR[2]	Output-T	H2	LHI[1]	Input	
D3	V <sub>DD</sub>	1.5V	H23	V <sub>SS</sub>	Ground	
D4	V <sub>SS</sub>	Ground	H24	NC NC	No Connect	
D5	V <sub>SS</sub>	Ground	H25	$V_{DDQ}$	2.5V/3.3V	
D6	V <sub>SS</sub>	Ground	H26	SADR[9]	Output-T	
D7	V <sub>SS</sub>	Ground	H3	NC	No Connect	
D8	V <sub>SS</sub>	Ground	H4	V <sub>SS</sub>	Ground	
D9	V <sub>SS</sub>	Ground	J1	LHI[2]	Input	
E1	ID[0]	No Connect	J2	LHI[3]	Input	
E2	V <sub>DDQ</sub>	2.5V/3.3V	J23	V <sub>SS</sub>	Ground	
E23	V <sub>SS</sub>	Ground	J24	SADR[11]	Output-T	
J25	SADR[12]	Output-T	M2	BHI[0]	Input	
J26	SADR[10]	Output-T	M23	V <sub>DD</sub>	1.5V	
J3	V <sub>DDQ</sub>	2.5V/3.3V	M24	V <sub>DD</sub>	1.5V	
J4	V <sub>SS</sub>	Ground	M25	V <sub>DDQ</sub>	2.5V/3.3V	
K1	LHI[6]	Input	M26	SADR[17]	Output-T	
K2	LHI[4]	Input	M3	V <sub>DD</sub>	1.5V	
K23	V <sub>SS</sub>	Ground	M4	V <sub>DD</sub>	1.5V	
K24	SADR[13]	Output-T	N1	BHI[1]	Input	
K25	V <sub>DDQ</sub>	2.5V/3.3V	N11	V <sub>SS</sub>	Ground	
K26	SADR[14]	Output-T	N12	V <sub>SS</sub>	Ground	
K3	LHI[5]	Input	N13	V <sub>SS</sub>	Ground	
K4	V <sub>SS</sub>	Ground	N14	V <sub>SS</sub>	Ground	
L1	LHO[0]	Output	N15	V <sub>SS</sub>	Ground	
L11 V <sub>SS</sub>		Ground	N16	V <sub>SS</sub>	Ground	
L12	V <sub>SS</sub> Ground		N2	BHI[2]		
L13		Ground	N23	V <sub>DD</sub>	Input 1.5V	
L13			N24	V <sub>DD</sub>	1.5V	
L14	V <sub>SS</sub>	Ground Ground	N24 N25	SADR[19]	Output-T	



**Table 18-1. Pinout Descriptions for Pinout Diagram** (continued)

Package Ball Number	Signal Name	Signal Type	Package Ball Number	Signal Name	Signal Type	
L16	$V_{SS}$	Ground	N26	SADR[18]	Output-T	
L2	LHO[1]	Output	N3	$V_{DD}$	1.5V	
L23	$V_{DD}$	1.5V	N4	$V_{DD}$	1.5V	
L24	$V_{DD}$	1.5V	P1	BHO[0]	Output	
L25	SADR[15]	Output-T	P11	V <sub>SS</sub>	Ground	
L26	SADR[16]	Output-T	P12	V <sub>SS</sub>	Ground	
L3	$V_{DD}$	1.5V	P13	V <sub>SS</sub>	Ground	
L4	$V_{DD}$	1.5V	P14	V <sub>SS</sub>	Ground	
M1	$V_{DDQ}$	2.5V/3.3V	P15	V <sub>SS</sub>	Ground	
M11	V <sub>SS</sub>	Ground	P16	V <sub>SS</sub>	Ground	
M12	V <sub>SS</sub>	Ground	P2	NC	No Connect	
M13	V <sub>SS</sub>	Ground	P23	V <sub>DD</sub>	1.5V	
M14	V <sub>SS</sub>	Ground	P24	$V_{DD}$	1.5V	
M15	V <sub>SS</sub>	Ground	P25	SADR[21]	Output-T	
M16	V <sub>SS</sub>	Ground	P26	SADR[20]	Output-T	
P3	V <sub>DD</sub>	1.5V	U24	OE_L	Output-T	
P4	$V_{DD}$	1.5V	U25	PHS_L	Input	
R1	$V_{\mathrm{DDQ}}$	2.5V/3.3V	U26	CLK1X/CLK2X	Input	
R11	$V_{SS}$	Ground	U3	FULI[1]	Input	
R12	$V_{SS}$	Ground	U4	V <sub>SS</sub>	Ground	
R13	V <sub>SS</sub>	Ground	V1	FULI[2]	Input	
R14	V <sub>SS</sub>	Ground	V2	FULI[3]	Input	
R15			V23	V <sub>SS</sub>	Ground	
R16	$V_{SS}$	Ground	V24	CE_L	Output-T	
R2	BHO[1]	Output	V25	$V_{\mathrm{DDQ}}$	2.5V/3.3V	
R23	$V_{DD}$	1.5V	V26	WE_L	Output-T	
R24	$V_{DD}$	1.5V	V3	FULI[4]	Input	
R25	SADR[22]	Output-T	V4	$V_{SS}$	Ground	
R26	$V_{\mathrm{DDQ}}$	2.5V/3.3V	W1	$V_{\mathrm{DDQ}}$	2.5V/3.3V	
R3	$V_{DD}$	1.5V	W2	FULI[5]	Input	
R4	$V_{DD}$	1.5V	W23	V <sub>SS</sub>	Ground	
T1	BHO[2]	Output	W24	NC	Output-T	
T11	$V_{SS}$	Ground	W25	CMDV	Input	
T12	$V_{SS}$	Ground	W26	ALE_L	Output-T	
T13	$V_{SS}$	Ground	W3	NC	No Connect	
T14	$V_{SS}$	Ground	W4	$V_{SS}$	Ground	
T15	$V_{SS}$	Ground	Y1	FULI[6]	Input	
T16			Y2	FULO[0]	Output	
T2	V <sub>SS</sub>	Ground	Y23	V <sub>SS</sub>	Ground	
T23	$V_{DD}$	1.5V	Y24	$V_{DD}$	1.5V	
T24	$V_{DD}$	1.5V	Y25	CMD[1]	Input	
T25	CLK_MODE	Input	Y26	CMD[0]	Input	
T26	SADR[23]	Output-T	Y3	$V_{DD}$	1.5V	



Table 18-1. Pinout Descriptions for Pinout Diagram (continued)

Package Ball Number	Signal Name	Signal Type	Package Ball Number	Signal Name	Signal Type
Т3	$V_{DD}$	1.5V	Y4	V <sub>SS</sub>	Ground
T4	$V_{DD}$	1.5V			
U1	FULI[0]	No Connect			
U2	$V_{\mathrm{DDQ}}$	2.5V/3.3V			
U23	V <sub>SS</sub>	Ground			

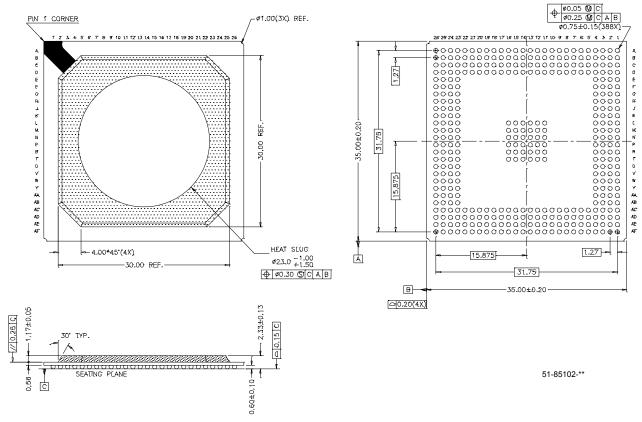
## 19.0 Ordering Information

Table 19-1. Ordering Information

Part Number	Description	I/O Voltage	Frequency	Temperature Range
CYNSE70256-066BHC	NSE	2.5V/3.3V	66 MHz	Commercial
CYNSE70256-083BHC	NSE	2.5V/3.3V	83 MHz	Commercial

## 20.0 Package Diagram

## 388-ball HSBGA (35 x 35 x 2.33 mm) BH388



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# **Document History Page**

	nt Title: CYN nt Number: 3		twork Searc	ch Engine
REV.	ECN NO.	Issue Date	Orig. of Change	Description of Change
**	110448	11/29/01	AFX	New Data Sheet
*A	112905	03/22/02	ED	Added 3.3V I/O specs. Added package diagrams.
*B	115995	08/27/02	KHS	Updated AC Timing, DC Characteristics, JTAG, Pinout Diagram and Pinout Description. Removed references to TEST signals from Pinout Diagram, Pinout Description and Signal Description. Added Power section covering power-up sequence and power consumption. Removed all references to 1.8V I/O. Removed all references to CLK_TUNE[3:0] and set it to 100% ("1001").
*C	119308	11/22/02	KHS	Changed package from BGC to BHC. Added min hold timing to 0.5 ns for t <sub>CKHDZ</sub> and t <sub>CKHSHZ</sub> . Added note to power-up sequence. Updated power-up figures. Remove alternative power-up sequence. Added note to JTAG testing.
*D	125508	05/08/03	DCU	Added note for LHI[0] and FULI[0] being unconnected. Added Ground in empty pin description. Clarified Parallel Write description. Changed diagram for power-up sequence. Changed from Preliminary to Final Data Sheet
*E	131896	12/12/03	FSG	Minor Change: Upload MPN to external website. No content change